

## LP2996A DDR Termination Regulator

### 1 Features

- 1.35V Minimum  $V_{DDQ}$
- Source and Sink Current
- Low Output Voltage Offset
- No External Resistors Required
- Linear Topology
- Suspend to Ram (STR) Functionality
- Low External Component Count
- Thermal Shutdown
- LP2998/8Q recommended for  $-40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$

### 2 Applications

- DDR1, DDR2, DDR3, and DDR3L Termination Voltage
- FPGA
- Industrial/Medical PC
- SSTL-2 and SSTL-3 Termination
- HSTL Termination

### 3 Description

The LP2996A linear regulator is designed to meet the JEDEC SSTL-2 specifications for termination of DDR-SDRAM. The device also supports DDR2, DDR3 and DDR3L VTT bus termination with  $V_{DDQ}$  min of 1.35V. The device contains a high-speed operational amplifier to provide excellent response to load transients. The output stage prevents shoot through while delivering 1.5A continuous current and transient peaks up to 3A in the application as required for DDR-SDRAM termination. The LP2996A also incorporates a  $V_{SENSE}$  pin to provide superior load regulation and a  $V_{REF}$  output as a reference for the chipset and DIMMs.

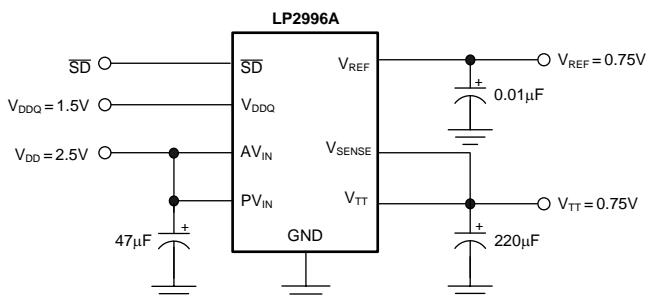
An additional feature found on the LP2996A is an active low shutdown ( $\overline{\text{SD}}$ ) pin that provides Suspend To RAM (STR) functionality. When  $\overline{\text{SD}}$  is pulled low the  $V_{TT}$  output will tri-state providing a high impedance output, but,  $V_{REF}$  will remain active. A power savings advantage can be obtained in this mode through lower quiescent current.

#### Device Information(1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LP2996A	SO PowerPAD (8)	4.89 mm x 3.90 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.

#### Simplified Schematic



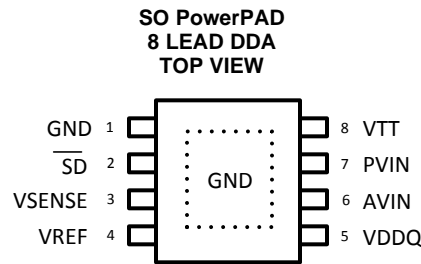
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## 4 Revision History

DATE	REVISION	NOTES
June 2014	*	Initial release.

## 5 Pin Configuration and Functions



### Pin Functions

PIN	TYPE	DESCRIPTION
1	GND	Ground
2	$\overline{\text{SD}}$	Shutdown
3	VSENSE	Feedback pin for regulating $V_{\text{TT}}$ .
4	VREF	Buffered internal reference voltage of $V_{\text{DDQ}}/2$
5	VDDQ	Input for internal reference equal to $V_{\text{DDQ}}/2$
6	AVIN	Analog input pin
7	PVIN	Power input pin
8	VTT	Output voltage for connection to termination resistors
	EP	Exposed pad thermal connection. Connect to Ground.

### 5.1 Pin Descriptions

<b>AVIN AND PVIN</b>	AVIN and PVIN are the input supply pins for the LP2996A. AVIN is used to supply all the internal control circuitry. PVIN, however, is used exclusively to provide the rail voltage for the output stage used to create VTT. These pins have the capability to work off separate supplies depending on the application. Higher voltages on PVIN will increase the maximum continuous output current because of output RDSON limitations at voltages close to VTT. The disadvantage of high values of PVIN is that the internal power loss will also increase, thermally limiting the design. For SSTL-2 applications, a good compromise would be to connect the AVIN and PVIN directly together at 2.5 V. This eliminates the need for bypassing the two supply pins separately. The only limitation on input voltage selection is that PVIN must be equal to or lower than AVIN. It is recommended to connect PVIN to voltage rails equal to or less than 3.3 V to prevent the thermal limit from tripping because of excessive internal power dissipation. If the junction temperature exceeds the thermal shutdown than the part will enter a shutdown state identical to the manual shutdown where VTT is tri-stated and VREF remains active.
<b>VDDQ</b>	VDDQ is the input used to create the internal reference voltage for regulating VTT. The reference voltage is generated from a resistor divider of two internal 50 kΩ resistors. This ensures that VTT will track VDDQ / 2 precisely. The optimal implementation of VDDQ is as a remote sense. This can be achieved by connecting VDDQ directly to the 2.5 V rail at the DIMM instead of AVIN and PVIN. This ensures that the reference voltage tracks the DDR memory rails precisely without a large voltage drop from the power lines. For SSTL-2 applications VDDQ will be a 2.5 V signal, which will create a 1.25 V termination voltage at VTT (See <a href="#">Electrical Characteristics</a> Table for exact values of VTT over temperature).
<b>VSENSE</b>	The purpose of the sense pin is to provide improved remote load regulation. In most motherboard applications the termination resistors will connect to VTT in a long plane. If the output voltage was regulated only at the output of the LP2996A then the long trace will cause a significant IR drop resulting in a termination voltage lower at one end of the bus than the other. The VSENSE pin can be used to improve this performance, by connecting it to the middle of the bus. This will provide a better distribution across the entire termination bus. If remote load regulation is not used then the VSENSE pin must still be connected to VTT. Care should be taken when a long VSENSE trace is implemented in close proximity to the memory. Noise pickup in the VSENSE trace can cause problems with precise regulation of VTT. A small 0.1 uF ceramic capacitor placed next to the VSENSE pin can help filter any high frequency signals and preventing errors.
<b>SHUTDOWN</b>	The LP2996A contains an active low shutdown pin that can be used to tri-state VTT. During shutdown VTT should not be exposed to voltages that exceed AVIN. With the shutdown pin asserted low the quiescent current of the LP2996A will drop, however, VDDQ will always maintain its constant impedance of 100 kΩ for generating the internal reference. Therefore to calculate the total power loss in shutdown both currents need to be considered. For more information refer to the <a href="#">Thermal Dissipation</a> section. The shutdown pin also has an internal pull-up current, therefore to turn the part on the shutdown pin can either be connected to AVIN or left open.

## Pin Descriptions (continued)

$V_{REF}$	$V_{REF}$ provides the buffered output of the internal reference voltage $V_{DDQ} / 2$ . This output should be used to provide the reference voltage for the Northbridge chipset and memory. Since these inputs are typically an extremely high impedance, there should be little current drawn from $V_{REF}$ . For improved performance, an output bypass capacitor can be used, located close to the pin, to help with noise. A ceramic capacitor in the range of 0.1 $\mu$ F to 0.01 $\mu$ F is recommended. This output remains active during the shutdown state and thermal shutdown events for the suspend to RAM functionality.
$V_{TT}$	$V_{TT}$ is the regulated output that is used to terminate the bus resistors. It is capable of sinking and sourcing current while regulating the output precisely to $V_{DDQ} / 2$ . The LP2996A is designed to handle peak transient currents of up to $\pm 3$ A with a fast transient response. The maximum continuous current is a function of $V_{IN}$ and can be viewed in the <a href="#">Typical Performance Characteristics</a> section. If a transient is expected to last above the maximum continuous current rating for a significant amount of time then the output capacitor should be sized large enough to prevent an excessive voltage drop. Despite the fact that the LP2996A is designed to handle large transient output currents it is not capable of handling these for long durations, under all conditions. The reason for this is the standard packages are not able to thermally dissipate the heat as a result of the internal power loss. If large currents are required for longer durations, then care should be taken to ensure that the maximum junction temperature is not exceeded. Proper thermal derating should always be used (please refer to the <a href="#">Thermal Dissipation</a> section). If the junction temperature exceeds the thermal shutdown point than $V_{TT}$ will tri-state until the part returns below the hysteretic trip-point.

## 6 Specifications

### 6.1 Absolute Maximum Ratings <sup>(1)(2)</sup>

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
	AVIN to GND	-0.3	6	V
	PVIN to GND	-0.3	AVIN	
	VDDQ <sup>(3)</sup>	-0.3	6	V
	Junction Temperature		150	°C
	Lead Temperature (Soldering, 10 sec)		260	°C

- (1) Absolute maximum ratings indicate limits beyond which damage to the device may occur. Operating range indicates conditions for which the device is intended to be functional, but does not ensure specific performance limits. For ensured specifications and test conditions see Electrical Characteristics. The ensured specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (3) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 Handling Ratings

		MIN	MAX	UNIT
$T_{stg}$	Storage temperature range	-65	150	°C
$V_{(ESD)}$	Electrostatic discharge		1	kV
	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins <sup>(1)</sup>			

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
Junction Temp. Range <sup>(1)</sup>	0		125	°C
AVIN to GND	2.2		5.5	V
PVIN Supply Voltage	0		AVIN	
SD Input Voltage	0		AVIN	

- (1) At elevated temperatures, devices must be derated based on thermal resistance.

## 6.4 Thermal Information

THERMAL METRIC <sup>(1)(2)(3)</sup>		SO PowerPAD-8 DDA	UNIT
		8 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	56.5	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	65.1	
R <sub>θJB</sub>	Junction-to-board thermal resistance	36.5	
Ψ <sub>JT</sub>	Junction-to-top characterization parameter	15.9	
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	36.5	
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	8.4	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

(2) The package thermal impedance is calculated in accordance with JESD 51-7

(3) Thermal Resistances were simulated on a 4 layer, JEDEC board.

## 6.5 Electrical Characteristics

Specifications are for T<sub>J</sub> = 25°C and apply over the full Operating Temperature Range (T<sub>J</sub> = 0°C to +125°C)<sup>(1)</sup>. Unless otherwise specified, AVIN = PVIN = 2.5V, VDDQ = 2.5V<sup>(2)</sup>.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>REF</sub>	V <sub>REF</sub> voltage (DDR I)	VIN = VDDQ = 2.3 V	1.135	1.158	1.185	V
		VIN = VDDQ = 2.5 V	1.235	1.258	1.285	
		VIN = VDDQ = 2.7 V	1.335	1.358	1.385	
	V <sub>REF</sub> voltage (DDR II)	PVIN = VDDQ = 1.7 V	0.837	0.860	0.887	
		PVIN = VDDQ = 1.8 V	0.887	0.910	0.937	
		PVIN = VDDQ = 1.9 V	0.936	0.959	0.986	
V <sub>REF</sub> Voltage (DDR III)	PVIN = VDDQ = 1.35V	0.669	0.684	0.699		
	PVIN = VDDQ = 1.5V	0.743	0.758	0.773		
	PVIN = VDDQ = 1.6V	0.793	0.808	0.823		
Z <sub>VREF</sub>	V <sub>REF</sub> Output Impedance	I <sub>REF</sub> = -30 to +30 μA	2.5		kΩ	
V <sub>TT</sub>	V <sub>TT</sub> Output Voltage (DDR I) <sup>(3)</sup>	I <sub>OUT</sub> = 0 A			V	
		VIN = VDDQ = 2.3 V	1.120	1.159		1.190
		VIN = VDDQ = 2.5 V	1.210	1.259		1.290
		VIN = VDDQ = 2.7 V	1.320	1.359		1.390
		I <sub>OUT</sub> = +/- 1.5 A				
		VIN = VDDQ = 2.3 V	1.125	1.159		1.190
		VIN = VDDQ = 2.5 V	1.225	1.259		1.290
V <sub>TT</sub> Output Voltage (DDR II) <sup>(3)</sup>	V <sub>TT</sub> Output Voltage (DDR II) <sup>(3)</sup>	I <sub>OUT</sub> = 0 A, AVIN = 2.5 V			V	
		PVIN = VDDQ = 1.7 V	0.822	0.856		0.887
		PVIN = VDDQ = 1.8 V	0.874	0.908		0.939
		PVIN = VDDQ = 1.9 V	0.923	0.957		0.988
		I <sub>OUT</sub> = +/- 0.5A, AVIN = 2.5 V				
		PVIN = VDDQ = 1.7 V	0.820	0.856		0.890
		PVIN = VDDQ = 1.8 V	0.870	0.908		0.940
PVIN = VDDQ = 1.9 V	0.920	0.957	0.990			

(1) Limits are 100% production tested at 25°C. Limits over the operating temperature range are specified through correlation using Statistical Quality Control (SQC) methods. The limits are used to calculate Texas Instruments' Average Outgoing Quality Level (AOQL).

(2) VIN is defined as VIN = AVIN = PVIN.

(3) V<sub>TT</sub> load regulation is tested by using a 10 ms current pulse and measuring V<sub>TT</sub>.

**Electrical Characteristics (continued)**

Specifications are for  $T_J = 25^\circ\text{C}$  and apply over the full Operating Temperature Range ( $T_J = 0^\circ\text{C}$  to  $+125^\circ\text{C}$ )<sup>(1)</sup>. Unless otherwise specified,  $AVIN = PVIN = 2.5\text{V}$ ,  $VDDQ = 2.5\text{V}$ <sup>(2)</sup>.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT			
$V_{TT}$ Output Voltage (DDR III) <sup>(3)</sup>	$I_{OUT} = 0\text{A}$ , $AVIN = 2.5\text{V}$				V			
	$PVIN = VDDQ = 1.35\text{V}$	0.656	0.677	0.698				
	$PVIN = VDDQ = 1.5\text{V}$	0.731	0.752	0.773				
	$PVIN = VDDQ = 1.6\text{V}$	0.781	0.802	0.823				
	$I_{OUT} = +0.2\text{A}$ , $AVIN = 2.5\text{V}$ $PVIN = VDDQ = 1.35\text{V}$	0.667	0.688	0.710				
	$I_{OUT} = -0.2\text{A}$ , $AVIN = 2.5\text{V}$ $PVIN = VDDQ = 1.35\text{V}$	0.641	0.673	0.694				
	$I_{OUT} = +0.4\text{A}$ , $AVIN = 2.5\text{V}$ $PVIN = VDDQ = 1.5\text{V}$	0.740	0.763	0.786				
	$I_{OUT} = -0.4\text{A}$ , $AVIN = 2.5\text{V}$ $PVIN = VDDQ = 1.5\text{V}$	0.731	0.752	0.773				
	$I_{OUT} = +0.5\text{A}$ , $AVIN = 2.5\text{V}$ $PVIN = VDDQ = 1.6\text{V}$	0.790	0.813	0.836				
	$I_{OUT} = -0.5\text{A}$ , $AVIN = 2.5\text{V}$ $PVIN = VDDQ = 1.6\text{V}$	0.781	0.802	0.823				
$VOS_{V_{TT}}$	$V_{TT}$ Output Voltage Offset ( $V_{REF} - V_{TT}$ ) for DDR I <sup>(3)</sup>	$I_{OUT} = 0\text{A}$	-30	0	30	mV		
		$I_{OUT} = -1.5\text{A}$	-30	0	30			
		$I_{OUT} = 1.5\text{A}$	-30	0	30			
	$V_{TT}$ Output Voltage Offset ( $V_{REF} - V_{TT}$ ) for DDR II <sup>(3)</sup>	$I_{OUT} = 0\text{A}$	-30	0	30			
		$I_{OUT} = -0.5\text{A}$	-30	0	30			
		$I_{OUT} = 0.5\text{A}$	-30	0	30			
	$V_{TT}$ Output Voltage Offset ( $V_{REF} - V_{TT}$ ) for DDR III <sup>(3)</sup>	$I_{OUT} = 0\text{A}$	-30	0	30			
		$I_{OUT} = \pm 0.2\text{A}$	-30	0	30			
		$I_{OUT} = \pm 0.4\text{A}$	-30	0	30			
		$I_{OUT} = \pm 0.5\text{A}$	-30	0	30			
	$I_Q$	Quiescent Current <sup>(4)</sup>	$I_{OUT} = 0\text{A}$		320		500	$\mu\text{A}$
	$Z_{VDDQ}$	VDDQ Input Impedance			100			$\text{k}\Omega$
$I_{SD}$	Quiescent current in shutdown <sup>(4)</sup>	$SD = 0\text{V}$		115	150	$\mu\text{A}$		
$I_{Q\_SD}$	Shutdown leakage current	$SD = 0\text{V}$		2	5	$\mu\text{A}$		
$V_{IH}$	Minimum Shutdown High Level			1.9		V		
$V_{IL}$	Maximum Shutdown Low Level				0.8			
$I_V$	$V_{TT}$ leakage current in shutdown	$SD = 0\text{V}$ $V_{TT} = 1.25\text{V}$		1	10	$\mu\text{A}$		
$I_{SENSE}$	$V_{SENSE}$ Input current			13		nA		
$T_{SD}$	Thermal Shutdown <sup>(5)</sup>			165		$^\circ\text{C}$		
$T_{SD\_HYS}$	Thermal Shutdown Hysteresis			10				

(4) Quiescent current defined as the current flow into AVIN.

(5) The maximum allowable power dissipation is a function of the maximum junction temperature,  $T_{J(MAX)}$ , the junction to ambient thermal resistance,  $\theta_{JA}$ , and the ambient temperature,  $T_A$ . Exceeding the maximum allowable power dissipation will cause excessive die temperature and the regulator will go into thermal shutdown.

## 6.6 Typical Performance Characteristics

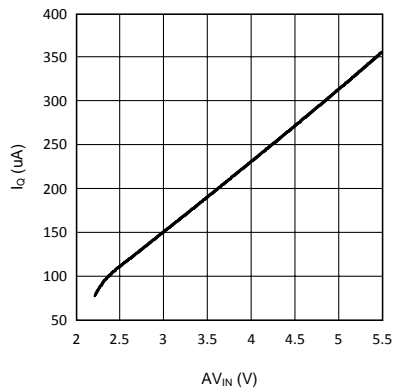


Figure 1. IQ vs AVIN In SD

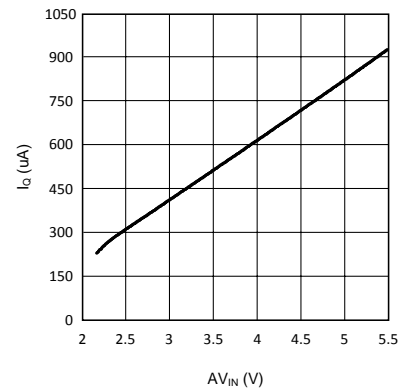


Figure 2. IQ vs AVIN

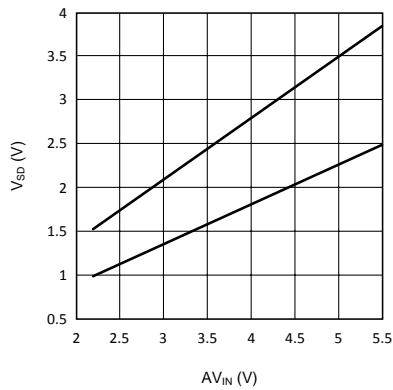


Figure 3. VIH and VIL

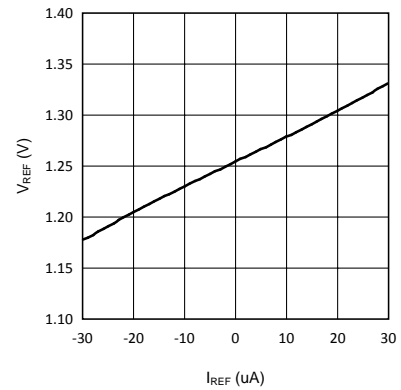


Figure 4. VREF vs IREF

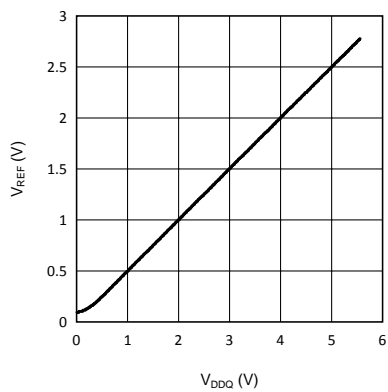


Figure 5. VREF vs VDDQ

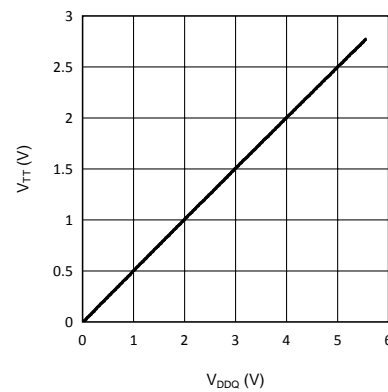


Figure 6. VTT vs VDDQ

Typical Performance Characteristics (continued)

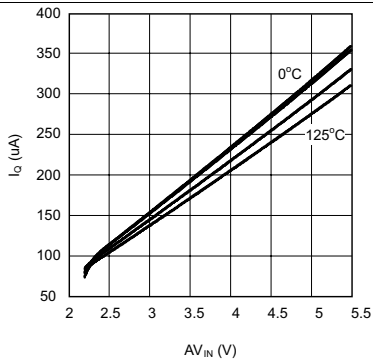


Figure 7. IQ vs AV<sub>IN</sub> in SD Temperature

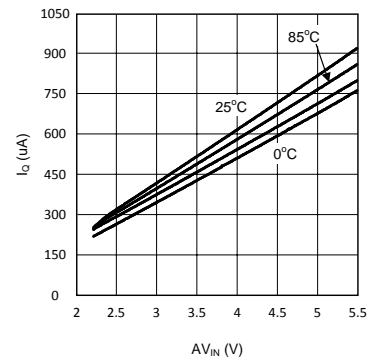


Figure 8. IQ vs AV<sub>IN</sub> Temperature

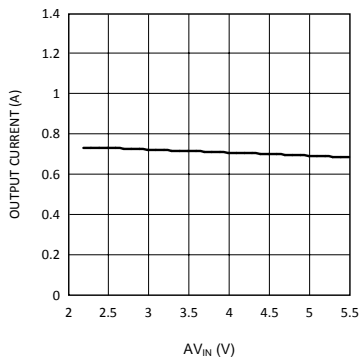


Figure 9. Maximum Sourcing Current vs AV<sub>IN</sub>  
(V<sub>DDQ</sub> = 2.5 V, PV<sub>IN</sub> = 1.8 V)

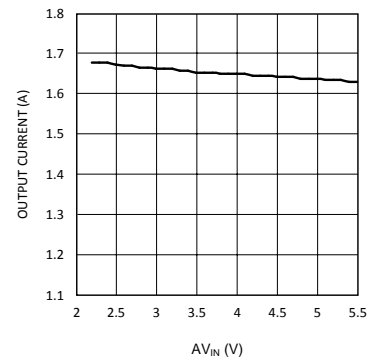


Figure 10. Maximum Sourcing Current vs AV<sub>IN</sub>  
(V<sub>DDQ</sub> = 2.5 V, PV<sub>IN</sub> = 2.5 V)

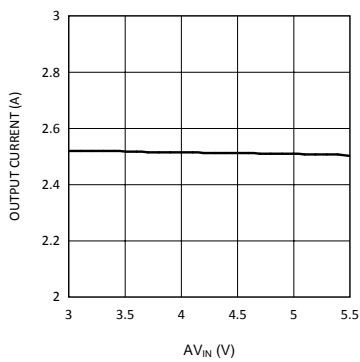


Figure 11. Maximum Sourcing Current vs AV<sub>IN</sub>  
(V<sub>DDQ</sub> = 2.5 V, PV<sub>IN</sub> = 3.3 V)

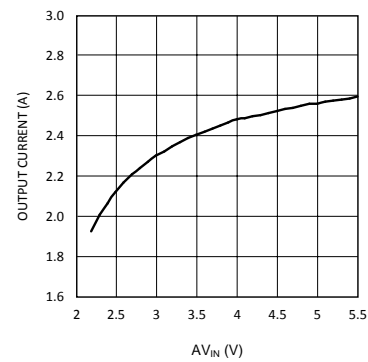


Figure 12. Maximum Sinking Current vs AV<sub>IN</sub>  
(V<sub>DDQ</sub> = 2.5 V)



Typical Performance Characteristics (continued)

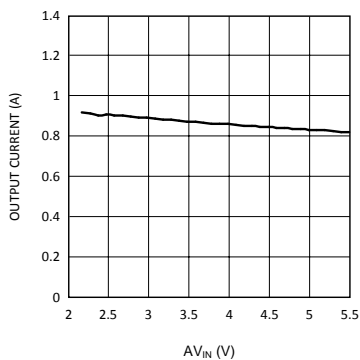


Figure 13. Maximum Sourcing Current vs AV<sub>IN</sub>  
(V<sub>DDQ</sub> = 1.8 V, PV<sub>IN</sub> = 1.8 V)

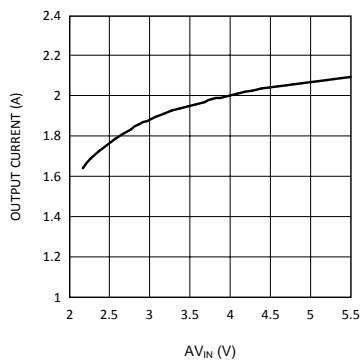


Figure 14. Maximum Sinking Current vs AV<sub>IN</sub>  
(V<sub>DDQ</sub> = 1.8 V)

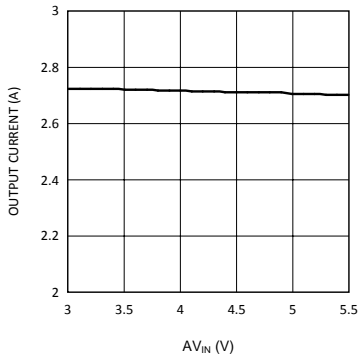


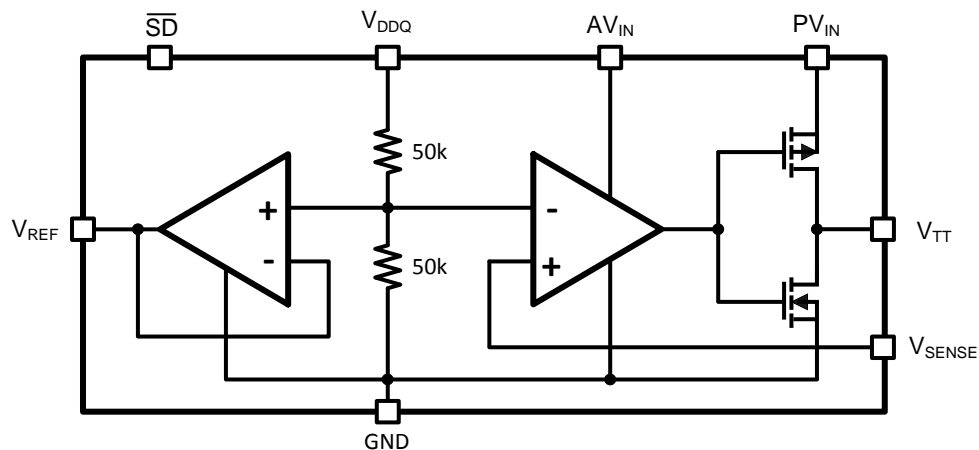
Figure 15. Maximum Sourcing Current vs AV<sub>IN</sub>  
(V<sub>DDQ</sub> = 1.8 V, PV<sub>IN</sub> = 3.3 V)

## 7 Detailed Description

### 7.1 Overview

The LP2996A linear regulator is designed to meet the JEDEC SSTL-2 specifications for termination of DDR-SDRAM. The device also supports DDR2, DDR3 and DDR3L VTT bus termination with  $V_{DDQ}$  min of 1.35V. The device contains a high-speed operational amplifier to provide excellent response to load transients. The output stage prevents shoot through while delivering 1.5A continuous current and transient peaks up to 3A in the application as required for DDR-SDRAM termination.

### 7.2 Functional Block Diagram

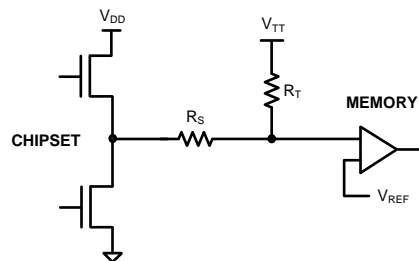


### 7.3 Feature Description

The LP2996A is a linear bus termination regulator designed to meet the JEDEC requirements of SSTL-2. The output,  $V_{TT}$  is capable of sinking and sourcing current while regulating the output voltage equal to  $V_{DDQ} / 2$ . The output stage has been designed to maintain excellent load regulation while preventing shoot through. The LP2996A also incorporates two distinct power rails that separates the analog circuitry from the power output stage. This allows a split rail approach to be utilized to decrease internal power dissipation. It also permits the LP2996A to provide a termination solution for DDR2-SDRAM, DDR3-SDRAM and DDR3L-SDRAM memory. For wide temperature designs, the LP2998/8Q is recommended for all DDR applications.

### 7.4 Device Functional Modes

The LP2996A can also be used to provide a termination voltage for other logic schemes such as SSTL-3 or HSTL. Series Stub Termination Logic (SSTL) was created to improve signal integrity of the data transmission across the memory bus. This termination scheme is essential to prevent data error from signal reflections while transmitting at high frequencies encountered with DDR-SDRAM. The most common form of termination is Class II single parallel termination. This involves one  $R_S$  series resistor from the chipset to the memory and one  $R_T$  termination resistor. Typical values for  $R_S$  and  $R_T$  are 25  $\Omega$ , although these can be changed to scale the current requirements from the LP2996A. This implementation can be seen below in [Figure 16](#).



**Figure 16. SSTL-Termination Scheme**

## 8 Applications and Implementation

### 8.1 Application Information

#### 8.1.1 Input Capacitor

The LP2996A does not require a capacitor for input stability, but it is recommended for improved performance during large load transients to prevent the input rail from dropping. The input capacitor should be located as close as possible to the PVIN pin. Several recommendations exist dependent on the application required. A typical value recommended for AL electrolytic capacitors is 50  $\mu\text{F}$ . Ceramic capacitors can also be used, a value in the range of 10  $\mu\text{F}$  with X5R or better would be an ideal choice. The input capacitance can be reduced if the LP2996A is placed close to the bulk capacitance from the output of the 2.5 V DC-DC converter. If the two supply rails (AVIN and PVIN) are separated then the 47  $\mu\text{F}$  capacitor should be placed as close to possible to the PVIN rail. An additional 0.1  $\mu\text{F}$  ceramic capacitor can be placed on the AVIN rail to prevent excessive noise from coupling into the device.

#### 8.1.2 Output Capacitor

The LP2996A has been designed to be insensitive of output capacitor size or ESR (Equivalent Series Resistance). This allows the flexibility to use any capacitor desired. The choice for output capacitor will be determined solely on the application and the requirements for load transient response of  $V_{\text{TT}}$ . As a general recommendation the output capacitor should be sized above 100  $\mu\text{F}$  with a low ESR for SSTL applications with DDR-SDRAM. The value of ESR should be determined by the maximum current spikes expected and the extent at which the output voltage is allowed to droop. Several capacitor options are available on the market and a few of these are highlighted below:

**AL** - It should be noted that many aluminum electrolytics only specify impedance at a frequency of 120 Hz, which indicates they have poor high frequency performance. Only aluminum electrolytics that have an impedance specified at a higher frequency (between 20 kHz and 100 kHz) should be used for the LP2996A. To improve the ESR several AL electrolytics can be combined in parallel for an overall reduction. An important note to be aware of is the extent at which the ESR will change over temperature. Aluminum electrolytic capacitors can have their ESR rapidly increase at cold temperatures.

**Ceramic** - Ceramic capacitors typically have a low capacitance, in the range of 10 to 100  $\mu\text{F}$  range, but they have excellent AC performance for bypassing noise because of very low ESR (typically less than 10 m $\Omega$ ). However, some dielectric types do not have good capacitance characteristics as a function of voltage and temperature. Because of the typically low value of capacitance it is recommended to use ceramic capacitors in parallel with another capacitor such as an aluminum electrolytic. A dielectric of X5R or better is recommended for all ceramic capacitors.

**Hybrid** - Several hybrid capacitors such as OS-CON and SP are available from several manufacturers. These offer a large capacitance while maintaining a low ESR. These are the best solution when size and performance are critical, although their cost is typically higher than any other capacitor.

#### 8.1.3 Thermal Dissipation

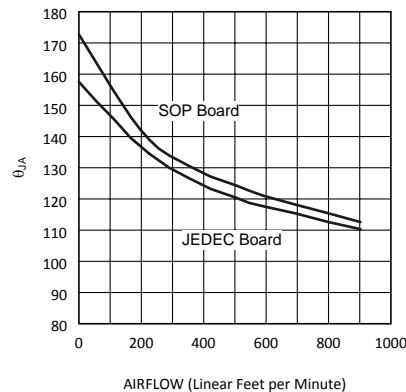
Since the LP2996A is a linear regulator any current flow from  $V_{\text{TT}}$  will result in internal power dissipation generating heat. To prevent damaging the part from exceeding the maximum allowable junction temperature, care should be taken to derate the part dependent on the maximum expected ambient temperature and power dissipation. The maximum allowable internal temperature rise ( $T_{\text{Rmax}}$ ) can be calculated given the maximum ambient temperature ( $T_{\text{Amax}}$ ) of the application and the maximum allowable junction temperature ( $T_{\text{Jmax}}$ ).

$$T_{\text{Rmax}} = T_{\text{Jmax}} - T_{\text{Amax}} \quad (1)$$

From this equation, the maximum power dissipation ( $P_{\text{Dmax}}$ ) of the part can be calculated:

$$P_{\text{Dmax}} = T_{\text{Rmax}} / \theta_{\text{JA}} \quad (2)$$

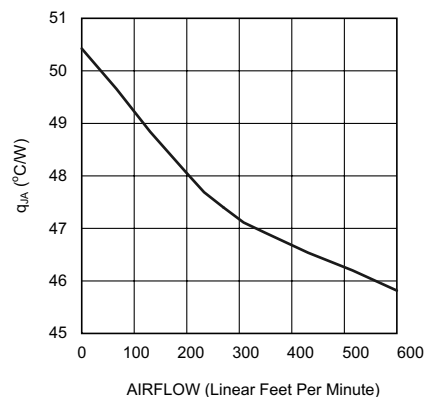
The  $\theta_{\text{JA}}$  of the LP2996A will be dependent on several variables: the package used; the thickness of copper; the number of vias and the airflow.

**Application Information (continued)**

**Figure 17.  $\theta_{JA}$  vs Airflow**

Additional improvements can be made by the judicious use of vias to connect the part and dissipate heat to an internal ground plane. Using larger traces and more copper on the top side of the board can also help. With careful layout it is possible to reduce the  $\theta_{JA}$  further than the nominal values shown in Figure 17.

Layout is also extremely critical to maximize the output current with the SO PowerPAD package. By simply placing vias under the DAP the  $\theta_{JA}$  can be lowered significantly.

Additional improvements in lowering the  $\theta_{JA}$  can also be achieved with a constant airflow across the package. Maintaining the same conditions as above and utilizing the 2x2 via array, Figure 18 shows how the  $\theta_{JA}$  varies with airflow.


**Figure 18.  $\theta_{JA}$  vs Airflow Speed (Jedec Board with 4 Vias)**

Optimizing the  $\theta_{JA}$  and placing the LP2996A in a section of a board exposed to lower ambient temperature allows the part to operate with higher power dissipation. The internal power dissipation can be calculated by summing the three main sources of loss: output current at  $V_{TT}$ , either sinking or sourcing, and quiescent current at AVIN and VDDQ. During the active state (when shutdown is not held low) the total internal power dissipation can be calculated from the following equations:

$$P_D = P_{AVIN} + P_{VDDQ} + P_{VTT} \quad (3)$$

where

$$P_{AVIN} = I_{AVIN} \cdot V_{AVIN} \quad (4)$$

$$P_{VDDQ} = V_{VDDQ} \cdot I_{VDDQ} = V_{VDDQ2} \times R_{VDDQ} \quad (5)$$

To calculate the maximum power dissipation at  $V_{TT}$  both conditions at  $V_{TT}$  need to be examined, sinking and sourcing current. Although only one equation will add into the total,  $V_{TT}$  cannot source and sink current simultaneously.

$$P_{VTT} = V_{VTT} \times I_{LOAD} \text{ (Sinking) or } \quad (6)$$

## Application Information (continued)

$$P_{V_{TT}} = (V_{P_{VIN}} - V_{V_{TT}}) \times I_{LOAD} \text{ (Sourcing)} \quad (7)$$

The power dissipation of the LP2996A can also be calculated during the shutdown state. During this condition the output  $V_{TT}$  will tri-state, therefore that term in the power equation will disappear as it cannot sink or source any current (leakage is negligible). The only losses during shutdown will be the reduced quiescent current at AVIN and the constant impedance that is seen at the VDDQ pin.

$$P_D = P_{AVIN} + P_{VDDQ} \quad (8)$$

$$P_{AVIN} = I_{AVIN} \times V_{AVIN} \quad (9)$$

$$P_{VDDQ} = V_{VDDQ} \times I_{VDDQ} = V_{VDDQ2} \times R_{VDDQ} \quad (10)$$

## 8.2 Typical Application

Several different application circuits are shown below to illustrate some of the options that are possible in configuring the LP2996A. Graphs of the individual circuit performance can be found in the [Typical Performance Characteristics](#) section in the beginning of the datasheet. These curves illustrate how the maximum output current is affected by changes in AVIN and PVIN.

### 8.2.1 Typical Application Circuit

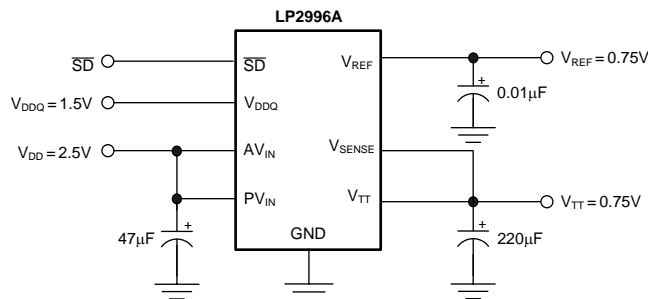


Figure 19. Typical Application Circuit

### 8.2.2 DDR-III Applications

With the separate VDDQ pin and an internal resistor divider it is possible to use the LP2996A in applications utilizing DDR-III memory. The output stage is connected to the 1.5 V rail and the AVIN pin can be connected to a 2.2 V - 5.5 V rail.

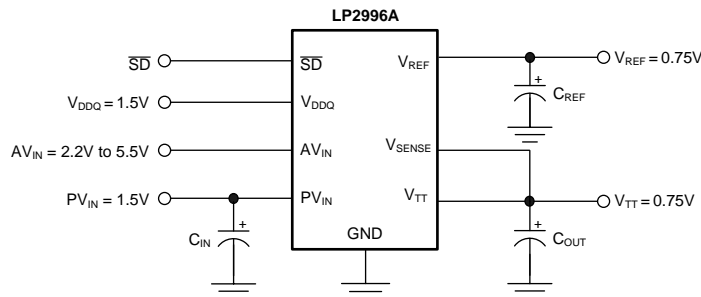


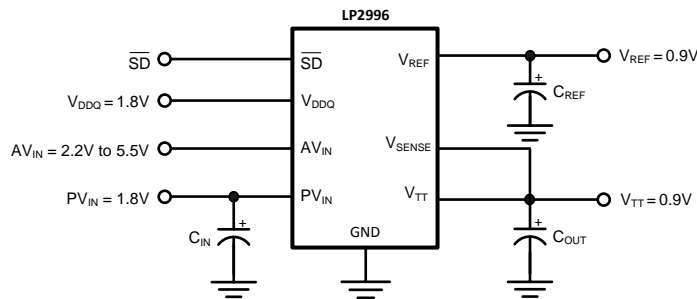
Figure 20. Recommended DDR-III Termination

If it is not desirable to use the 1.5 V - 2.5 V rail it is possible to connect the output stage to a 3.3 V rail. Care should be taken to not exceed the maximum junction temperature as the thermal dissipation increases with lower  $V_{TT}$  output voltages. For this reason it is not recommended to power PVIN off a rail higher than the nominal 3.3 V. The advantage of this configuration is that it has the ability to source and sink a higher maximum continuous current.

## Typical Application (continued)

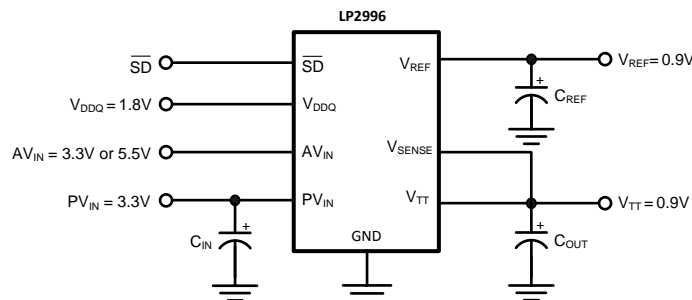
### 8.2.3 DDR-II Applications

With the separate VDDQ pin and an internal resistor divider it is possible to use the LP2996A in applications utilizing DDR-II memory. [Figure 24](#) and [Figure 25](#) show several implementations of recommended circuits with output curves displayed in the [Typical Performance Characteristics](#). [Figure 24](#) shows the recommended circuit configuration for DDR-II applications. The output stage is connected to the 1.8 V rail and the AVIN pin can be connected to either a 3.3 V or 5 V rail. For DDR-III and DDR-III low power designs in wider temperature applications, the LP2998/Q is recommended.



**Figure 21. Recommended DDR-II Termination**

If it is not desirable to use the 1.8 V rail it is possible to connect the output stage to a 3.3 V rail. Care should be taken to not exceed the maximum junction temperature as the thermal dissipation increases with lower  $V_{TT}$  output voltages. For this reason it is not recommended to power PVIN off a rail higher than the nominal 3.3 V. The advantage of this configuration is that it has the ability to source and sink a higher maximum continuous current.



**Figure 22. DDR-II Termination with Higher Voltage Rails**

### 8.2.4 SSTL-2 Applications

For the majority of applications that implement the SSTL-2 termination scheme it is recommended to connect all the input rails to the 2.5 V rail. This provides an optimal trade-off between power dissipation and component count and selection. An example of this circuit can be seen in [Figure 23](#).

Typical Application (continued)

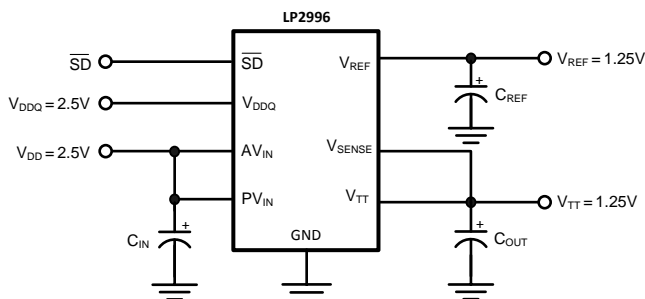


Figure 23. Recommended SSTL-2 Implementation

If power dissipation or efficiency is a major concern then the LP2996A has the ability to operate on split power rails. The output stage (PVIN) can be operated on a lower rail such as 1.8 V and the analog circuitry (AVIN) can be connected to a higher rail such as 2.5 V, 3.3 V or 5 V. This allows the internal power dissipation to be lowered when sourcing current from VTT. The disadvantage of this circuit is that the maximum continuous current is reduced because of the lower rail voltage, although it is adequate for all motherboard SSTL-2 applications. Increasing the output capacitance can also help if periods of large load transients will be encountered.

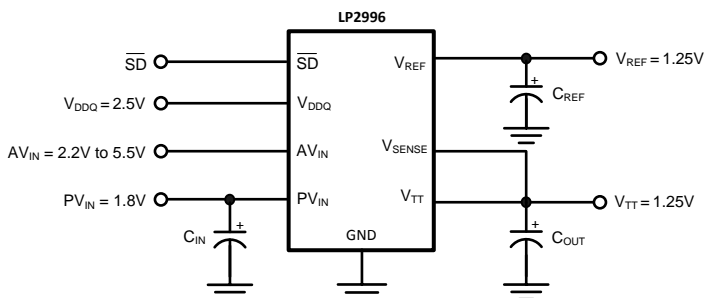


Figure 24. Lower Power Dissipation SSTL-2 Implementation

The third option for SSTL-2 applications in the situation that a 1.8 V rail is not available and it is not desirable to use 2.5 V, is to connect the LP2996A power rail to 3.3 V. In this situation AVIN will be limited to operation on the 3.3 V or 5 V rail as PVIN can never exceed AVIN. This configuration has the ability to provide the maximum continuous output current at the downside of higher thermal dissipation. Care should be taken to prevent the LP2996A from experiencing large current levels which cause the junction temperature to exceed the maximum. Because of this risk it is not recommended to supply the output stage with a voltage higher than a nominal 3.3 V rail.

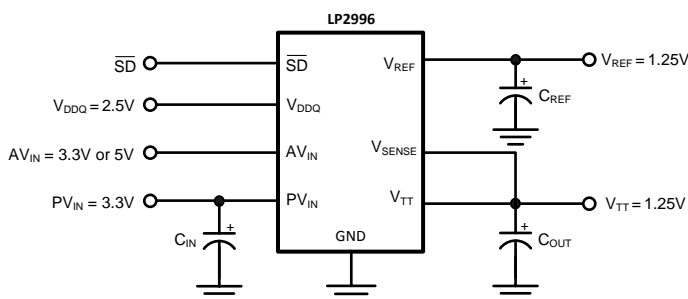


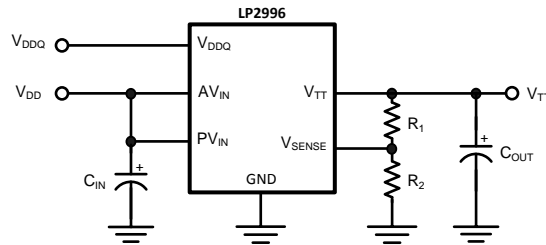
Figure 25. SSTL-2 Implementation with Higher Voltage Rails

## Typical Application (continued)

### 8.2.5 Level Shifting

If standards other than SSTL-2 are required, such as SSTL-3, it may be necessary to use a different scaling factor than 0.5 times  $V_{DDQ}$  for regulating the output voltage. Several options are available to scale the output to any voltage required. One method is to level shift the output by using feedback resistors from  $V_{TT}$  to the  $V_{SENSE}$  pin. This has been illustrated in Figure 26 and Figure 27. Figure 26 shows how to use two resistors to level shift  $V_{TT}$  above the internal reference voltage of  $V_{DDQ}/2$ . To calculate the exact voltage at  $V_{TT}$  the following equation can be used.

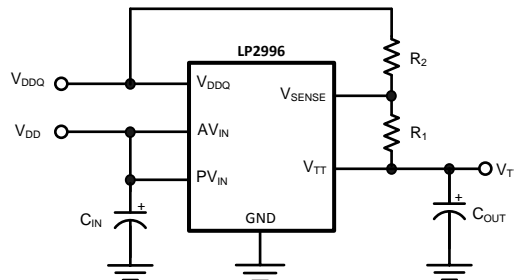
$$V_{TT} = V_{DDQ}/2 (1 + R_1/R_2) \quad (11)$$



**Figure 26. Increasing VTT by Level Shifting**

Conversely, the  $R_2$  resistor can be placed between  $V_{SENSE}$  and  $V_{DDQ}$  to shift the  $V_{TT}$  output lower than the internal reference voltage of  $V_{DDQ}/2$ . The equations relating  $V_{TT}$  and the resistors can be seen below:

$$V_{TT} = V_{DDQ}/2 (1 - R_1/R_2) \quad (12)$$



**Figure 27. Decreasing VTT by Level Shifting**

#### 8.2.5.1 Output Capacitor Selection

For applications utilizing the LP2996A to terminate SSTL-2 I/O signals the typical application circuit shown in Figure 27 can be implemented.

This circuit permits termination in a minimum amount of board space and component count. Capacitor selection can be varied depending on the number of lines terminated and the maximum load transient. However, with motherboards and other applications where  $V_{TT}$  is distributed across a long plane it is advisable to use multiple bulk capacitors and addition to high frequency decoupling. Figure 28 shown below depicts an example circuit where 2 bulk output capacitors could be situated at both ends of the  $V_{TT}$  plane for optimal placement. Large aluminum electrolytic capacitors are used for their low ESR and low cost.

In most PC applications an extensive amount of decoupling is required because of the long interconnects encountered with the DDR-SDRAM DIMMs mounted on modules. As a result bulk aluminum electrolytic capacitors in the range of 1000uF are typically used.



## Typical Application (continued)

### 8.2.6 HSTL Applications

The LP2996A can be easily adapted for HSTL applications by connecting  $V_{DDQ}$  to the 1.5 V rail. This will produce a  $V_{TT}$  and  $V_{REF}$  voltage of approximately 0.75 V for the termination resistors.  $AV_{IN}$  and  $PV_{IN}$  should be connected to a 2.5 V rail for optimal performance.

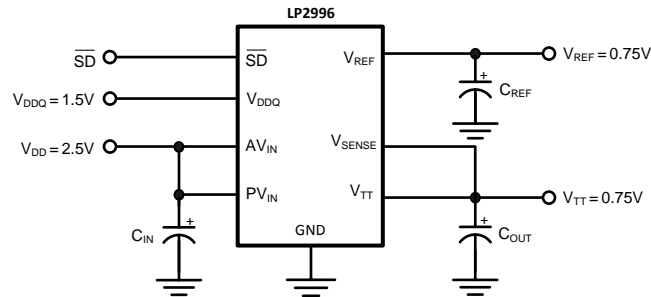


Figure 28. HSTL Application

### 8.2.7 QDR Applications

Quad data rate (QDR) applications utilize multiple channels for improved memory performance. However, this increase in bus lines has the effect of increasing the current levels required for termination. The recommended approach in terminating multiple channels is to use a dedicated LP2996A for each channel. This simplifies layout and reduces the internal power dissipation for each regulator. Separate  $V_{REF}$  signals can be used for each DIMM bank from the corresponding regulator with the chipset reference provided by a local resistor divider or one of the LP2996A signals. Because  $V_{REF}$  and  $V_{TT}$  are expected to track and the part to part variations are minor, there should be little difference between the reference signals of each LP2996A.

## 9 Power Supply Recommendations

There are several recommendations for the LP2996A input power supply. An input capacitor is not required but is recommended for improved performance during large load transients to prevent the input rail from dropping. The input capacitor should be located as close as possible to the PVIN pin. Several recommendations exist dependent on the application required. A typical value recommended for AL electrolytic capacitors is 50  $\mu\text{F}$ . Ceramic capacitors can also be used, a value in the range of 10  $\mu\text{F}$  with X5R or better would be an ideal choice. The input capacitance can be reduced if the LP2996A is placed close to the bulk capacitance from the output of the 2.5 V DC-DC converter. If the two supply rails (AVIN and PVIN) are separated then the 47  $\mu\text{F}$  capacitor should be placed as close to possible to the PVIN rail. An additional 0.1  $\mu\text{F}$  ceramic capacitor can be placed on the AVIN rail to prevent excessive noise from coupling into the device.

## 10 Layout

### 10.1 Layout Guidelines

1. The input capacitor for the power rail should be placed as close as possible to the PVIN pin.
2.  $V_{SENSE}$  should be connected to the  $V_{TT}$  termination bus at the point where regulation is required. For motherboard applications an ideal location would be at the center of the termination bus.
3.  $V_{DDQ}$  can be connected remotely to the  $V_{DDQ}$  rail input at either the DIMM or the Chipset. This provides the most accurate point for creating the reference voltage.
4. For improved thermal performance excessive top side copper should be used to dissipate heat from the package. Numerous vias from the ground connection to the internal ground plane will help. Additionally these can be located underneath the package if manufacturing standards permit.
5. Care should be taken when routing the  $V_{SENSE}$  trace to avoid noise pickup from switching I/O signals. A 0.1uF ceramic capacitor located close to the  $V_{SENSE}$  can also be used to filter any unwanted high frequency signal. This can be an issue especially if long  $V_{SENSE}$  traces are used.
6.  $V_{REF}$  should be bypassed with a 0.01  $\mu$ F or 0.1  $\mu$ F ceramic capacitor for improved performance. This capacitor should be located as close as possible to the  $V_{REF}$  pin.

### 10.2 Layout Examples

The LP2996A layout is very similar to the LP2998/Q layout. This is because the main difference between the two IC's is the wider temperature range,  $-40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ , which the LP2998/Q offers. As such, the below example shows the layout from a LP2998EVM. These layout examples can be used to evaluate the LP2996A.

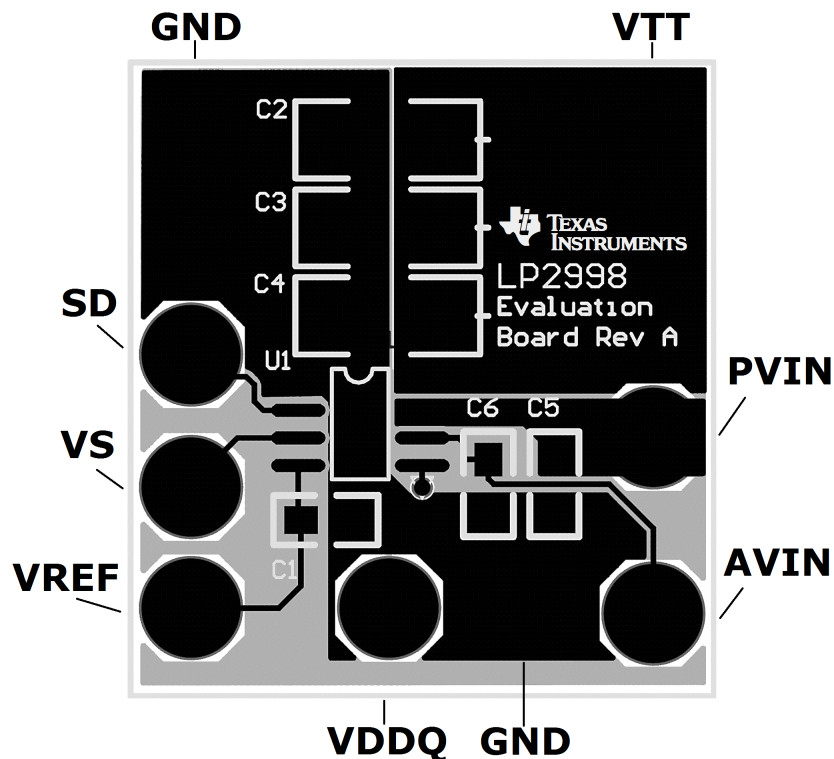


Figure 29. LP2998EVM SO PowerPAD Layout Example (Front)

Layout Examples (continued)

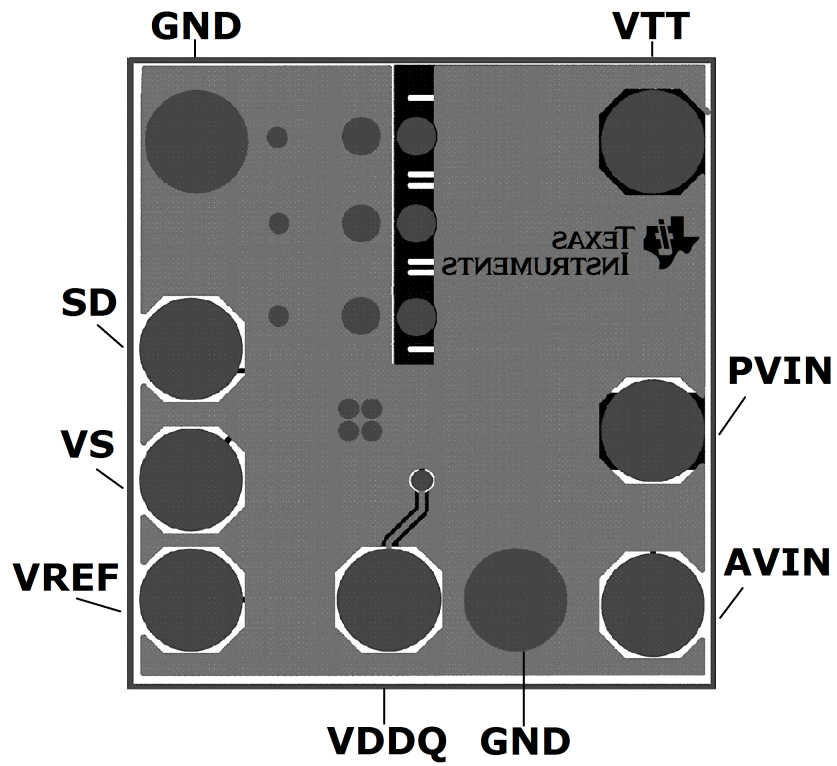


Figure 30. LP2998EVM SO PowerPAD Layout Example (Back)

## 11 Device and Documentation Support

### 11.1 Trademarks

All trademarks are the property of their respective owners.

### 11.2 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### 11.3 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LP2996AMR/NOPB	ACTIVE	SO PowerPAD	DDA	8	95	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	0 to 125	LP2996 AMR	<a href="#">Samples</a>
LP2996AMRE/NOPB	ACTIVE	SO PowerPAD	DDA	8	250	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	0 to 125	LP2996 AMR	<a href="#">Samples</a>
LP2996AMRX/NOPB	ACTIVE	SO PowerPAD	DDA	8	2500	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	0 to 125	LP2996 AMR	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

## TAPE AND REEL INFORMATION



### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LP2996AMRE/NOPB	SO Power PAD	DDA	8	250	178.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LP2996AMRX/NOPB	SO Power PAD	DDA	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1

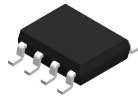


**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LP2996AMRE/NOPB	SO PowerPAD	DDA	8	250	210.0	185.0	35.0
LP2996AMRX/NOPB	SO PowerPAD	DDA	8	2500	367.0	367.0	35.0

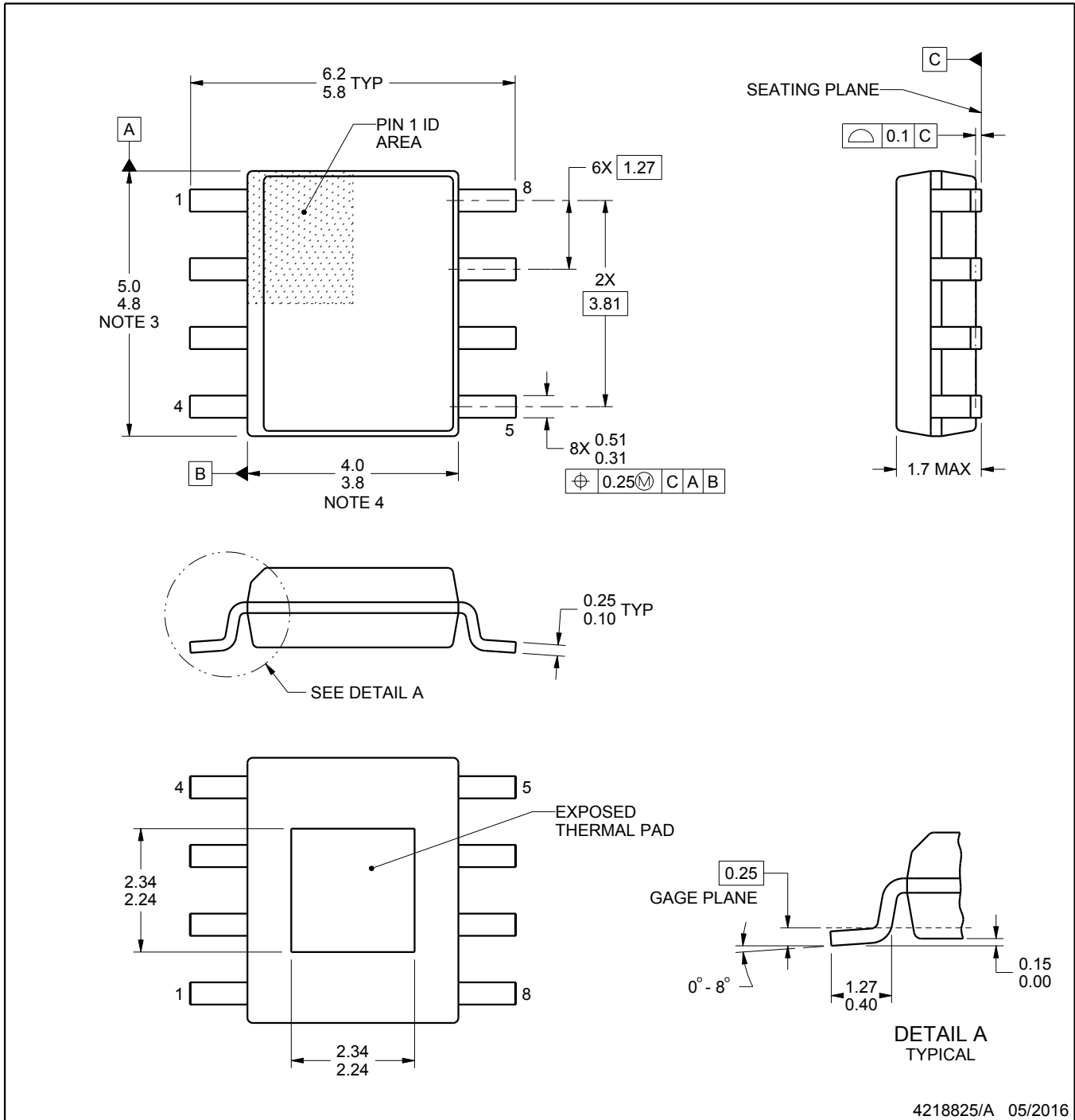
DDA0008A



# PACKAGE OUTLINE

PowerPAD™ SOIC - 1.7 mm max height

PLASTIC SMALL OUTLINE



4218825/A 05/2016

PowerPAD is a trademark of Texas Instruments.

**NOTES:**

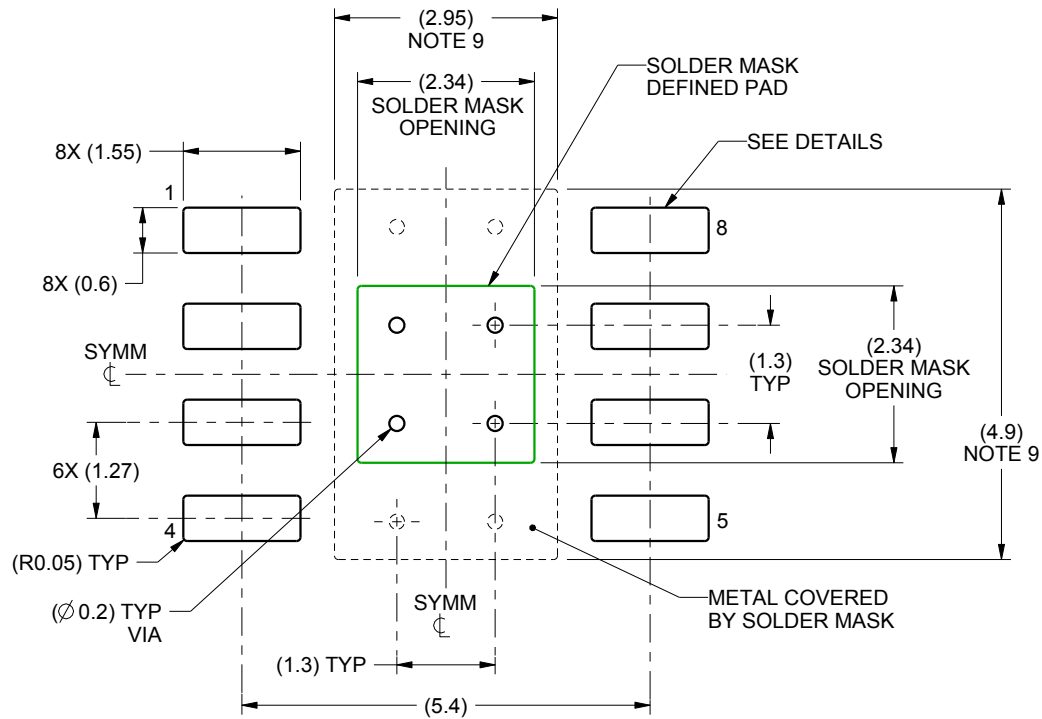
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MS-012.

# EXAMPLE BOARD LAYOUT

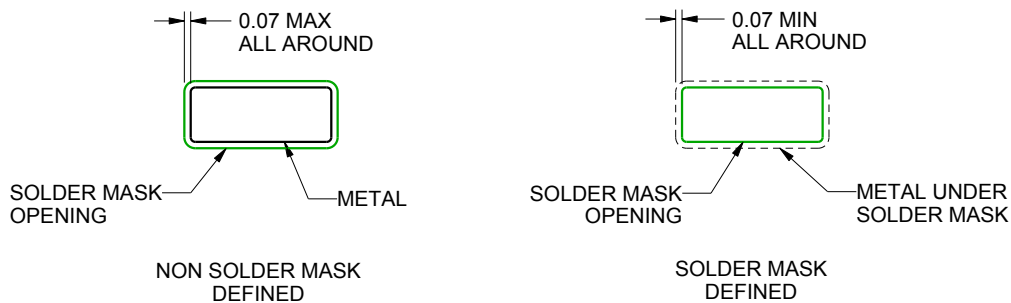
DDA0008A

PowerPAD™ SOIC - 1.7 mm max height

PLASTIC SMALL OUTLINE



LAND PATTERN EXAMPLE  
SCALE:10X



SOLDER MASK DETAILS

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NOTES: (continued)

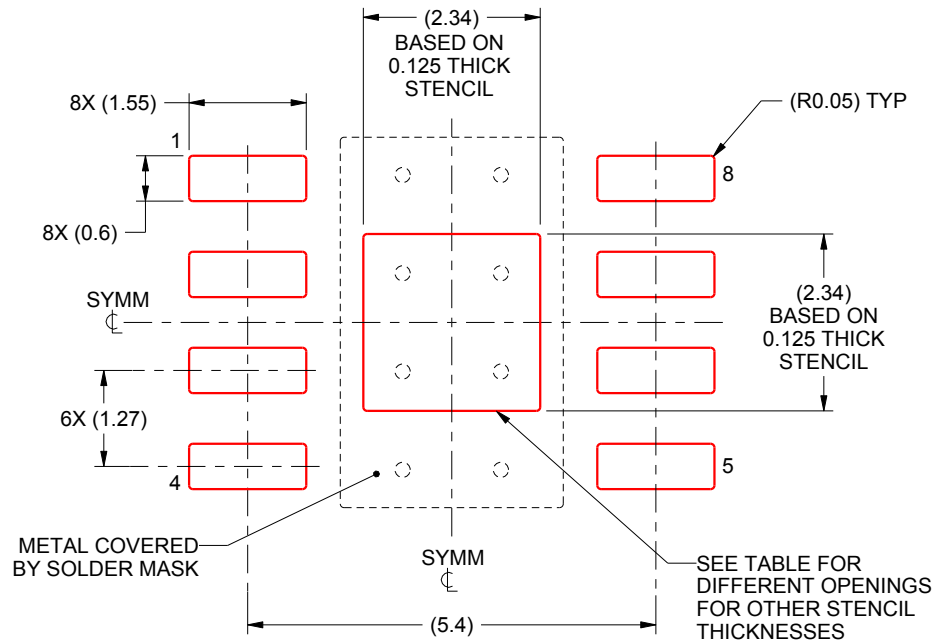
6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature numbers SLMA002 ([www.ti.com/lit/slma002](http://www.ti.com/lit/slma002)) and SLMA004 ([www.ti.com/lit/slma004](http://www.ti.com/lit/slma004)).
9. Size of metal pad may vary due to creepage requirement.
10. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

# EXAMPLE STENCIL DESIGN

DDA0008A

PowerPAD™ SOIC - 1.7 mm max height

PLASTIC SMALL OUTLINE



SOLDER PASTE EXAMPLE  
EXPOSED PAD  
100% PRINTED SOLDER COVERAGE BY AREA  
SCALE:10X

STENCIL THICKNESS	SOLDER STENCIL OPENING
0.1	2.62 X 2.62
0.125	2.34 X 2.34 (SHOWN)
0.150	2.14 X 2.14
0.175	1.98 X 1.98

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NOTES: (continued)

11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
12. Board assembly site may have different recommendations for stencil design.

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