GENERAL DESCRIPTION

OB2353 combines a dedicated current mode PWM controller with a high voltage power MOSFET. It is optimized for high performance, low standby power, and cost effective off-line flyback converter applications in sub 8W range.

OB2353 offers complete protection coverage with automatic self-recovery feature including Cycle-by-Cycle current limiting (OCP), over load protection (OLP), over temperature protection (OTP),VDD over voltage clamp and under voltage lockout (UVLO). Excellent EMI performance is achieved with On-Bright proprietary frequency shuffling technique together with soft switching control at the totem pole gate drive output.

The tone energy at below 20KHZ is minimized in the design and audio noise is eliminated during operation. OB2353 is offered in SOP8 package.

FEATURES

- Power on Soft Start Reducing MOSFT Vds Stress
- Frequency shuffling for EMI
- Extended Burst Mode Control For Improved Efficiency and Minimum Standby Power Design
- Audio Noise Free Operation
- Fixed 50KHZ Switching Frequency
- Internal Synchronized Slope Compensation
- Low VDD Startup Current and Low Operating Current
- Leading Edge Blanking on Current Sense Input
- Good Protection Coverage With Auto Self-Recovery
  - VDD Over Voltage Clamp and Under Voltage Lockout with Hysteresis (UVLO)
  - Over Temperature Protection (OTP)
  - On-Bright Proprietary Line Input Compensated Cycle-by-Cycle Over-current Threshold Setting For Constant Output Power Limiting Over Universal Input Voltage Range.
  - Overload Protection (OLP).
  - Over voltage Protection(OVP)

APPLICATIONS

Offline AC/DC flyback converter for

- Cellular Phone Battery Charger
- PDA power supplies
- Digital Camera Adaptor
- Auxiliary Power Supply for PC and Server
- Open-frame SMPS

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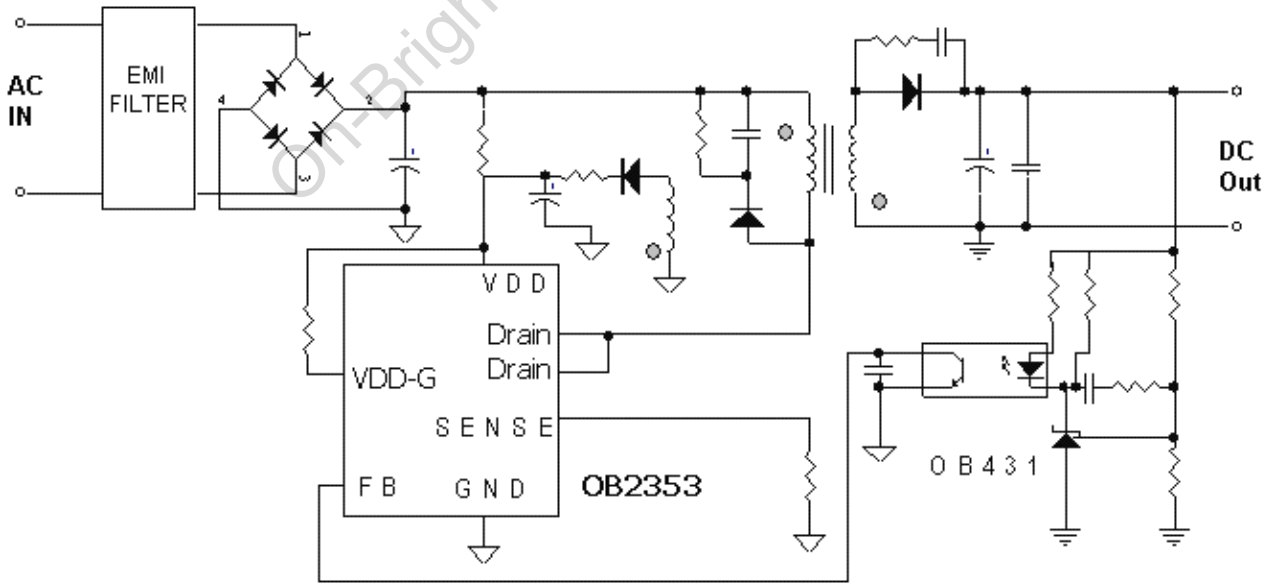
TYPICAL APPLICATION



Output Power Table

Product	230VAC±15%	85-265VAC
	Open Frame[1]	Open Frame[1]
OB2353	8W	6W

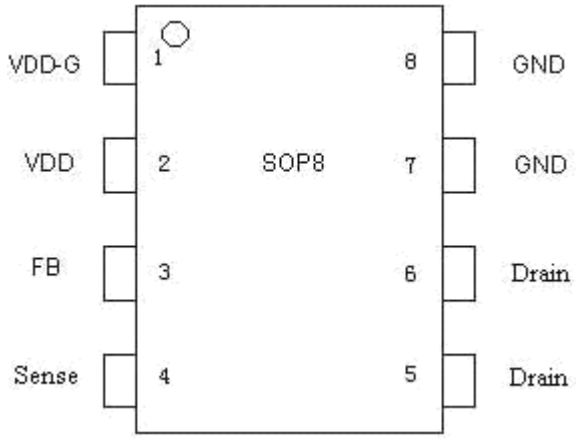
Notes:

Maximum practical continuous power in an open frame design with sufficient drain pattern as a heat sink, at 50℃ ambient. Higher output power is possible with extra added heat sink or air circulation to reduce thermal resistance.

GENERAL INFORMATION

Pin Configuration

The OB2353 is offered in SOP8 package as shown below.



Ordering Information

Part Number	Description
OB2353CP	SOP8, Pb-free in Tube
OB2353CPA	SOP8, PB-free in T&R

Package Dissipation Rating

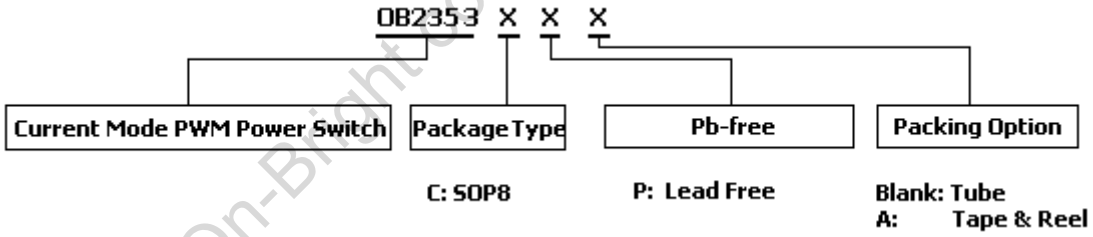
Package	RθJA (℃/W)
SOP8	90

Note: Drain Pin Connected to 100mm2 PCB copper clad.

Absolute Maximum Ratings

Parameter	Value
Drain Voltage (off state)	-0.3V to 600V
VDD Voltage	-0.3V to 30 V
VDD-G Input Voltage	-0.3V to 30 V
VDD Clamp Continuous Current	10mA
FB Input Voltage	-0.3 to 7V
Sense Input Voltage	-0.3 to 7V
Min/Max Operating Junction Temperature T_J	-20 to 150℃
Min/Max Storage Temperature T_{stg}	-55 to 150℃
Lead Temperature (Soldering, 10secs)	260℃

Note: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum-rated conditions for extended periods may affect device reliability.



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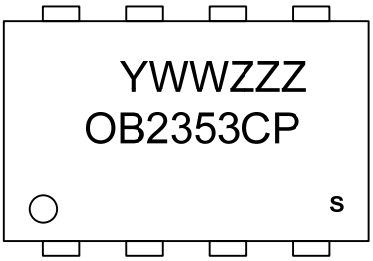
Marking Information

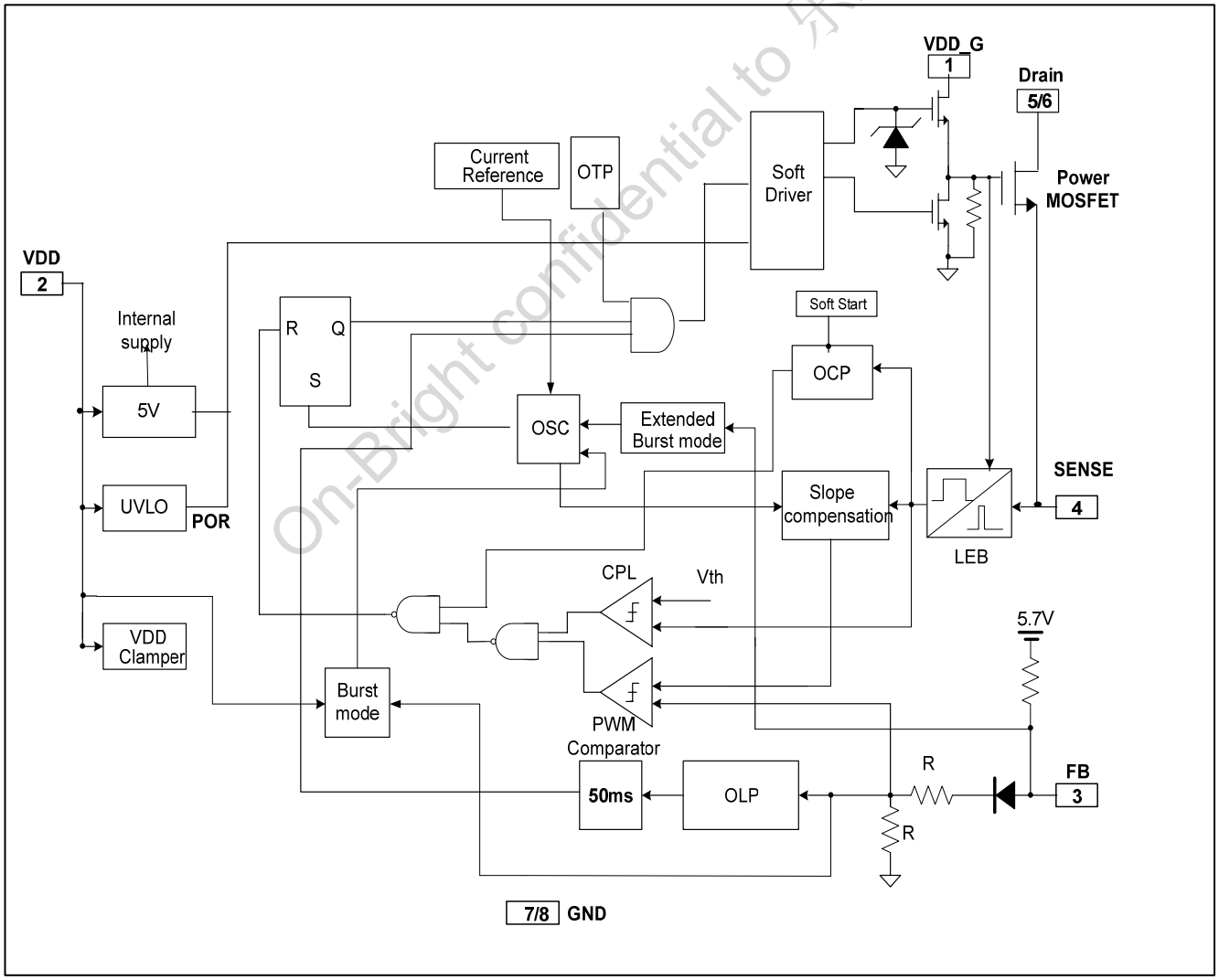


Y:Year Code
WW:Week Code(01-52)
ZZZ:Lot Code
C:SOP8 Package
P:Pb-free Package
S:Internal Code(Optional)

TERMINAL ASSIGNMENTS

Pin Name	I/O	Description
GND	P	Ground
FB	I	Feedback input pin. The PWM duty cycle is determined by voltage level into this pin and the current-sense signal at Pin 4.
VDD-G	P	Internal Gate Driver Power Supply
SENSE	I	Current sense input
VDD	P	IC DC power supply Input
Drain	O	HV MOSFET Drain Pin. The Drain pin is connected to the primary lead of the transformer

BLOCK DIAGRAM



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ELECTRICAL CHARACTERISTICS

(T_A = 25℃, VDD=16V, if not otherwise noted)

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
Supply Voltage (VDD)						
Istartup	VDD Start up Current	VDD=14.5V,Measure Leakage current into VDD		5	20	uA
I_VDD_Operation	Operation Current	V_{FB}=3V		1.6		mA
UVLO(ON)	VDD Under Voltage Lockout Enter		8.7	9.7	10.7	V
UVLO(OFF)	VDD Under Voltage Lockout Exit (Recovery)		14.6	15.8	17.0	V
OVP(ON)	Over voltage protection voltage	CS=0V,FB=3V Ramp up VDD until gate clock is off	27.0	28.5	30.0	V
VDD_Clamp	VDD Zener Clamp Voltage	I_{DD} = 10 mA		30		V
Feedback Input Section(FB Pin)						
V_{FB}_Open	V_{FB} Open Loop Voltage		5.4	5.7	6.0	V
I_{FB}_Short	FB pin short circuit current	Short FB pin to GND and measure current		1.45		mA
V_{TH}_0D	Zero Duty Cycle FB Threshold Voltage			0.8		V
V_{TH}_PL	Power Limiting FB Threshold Voltage			3.7		V
T_D_PL	Power limiting Debounce Time			50		mSec
Z_{FB}_IN	Input Impedance			4		Kohm
Current Sense Input(Sense Pin)						
Soft start time				4		ms
T_blanking	Leading edge blanking time			270		ns
Z_{SENSE}_IN	Input Impedance			40		Kohm
T_D_OC	Over Current Detection and Control Delay	From Over Current Occurs till the Gatedrive output start to turn off		120		nSec
V_{TH}_OC	Internal Current Limiting Threshold Voltage	FB=3.3V	0.72	0.77	0.82	V
Oscillator						
F_{OSC}	Normal Oscillation Frequency		45	50	55	KHZ
△f_Temp	Frequency Temperature Stability			5		%

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△f_VDD	Frequency Voltage Stability			5		%
D_max	Maximum duty cycle	FB=3.3V, CS =0V	56	66	76	%
F_Burst	Burst Mode Base Frequency			22		KHZ
Mosfet Section						
BVdss	MOSFET Drain-Source Breakdown Voltage		600	650		V
RDS(on)	Static Drain to Source On Resistance			12	15	Ω
Frequency Shuffling						
Δf_OSC	Frequency Modulation range /Base frequency		-4		4	%
Over Temperature Protection						
Over temperature protection trip point				150		℃

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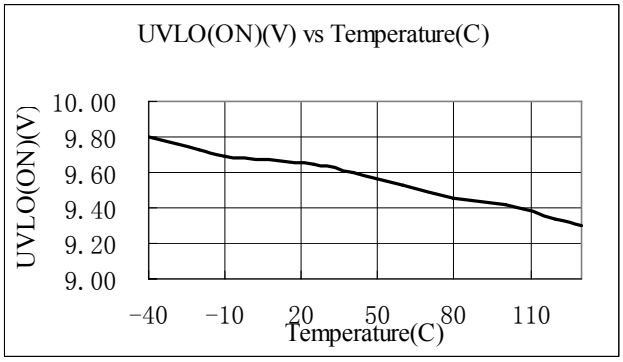
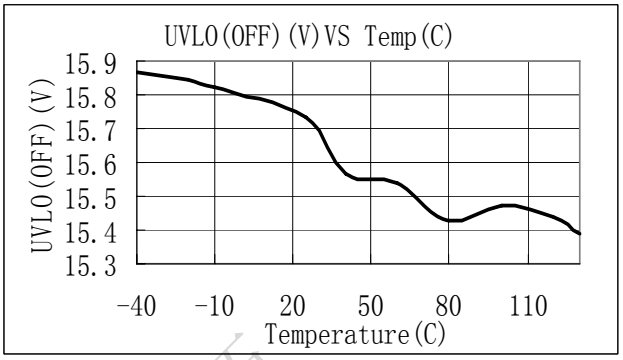
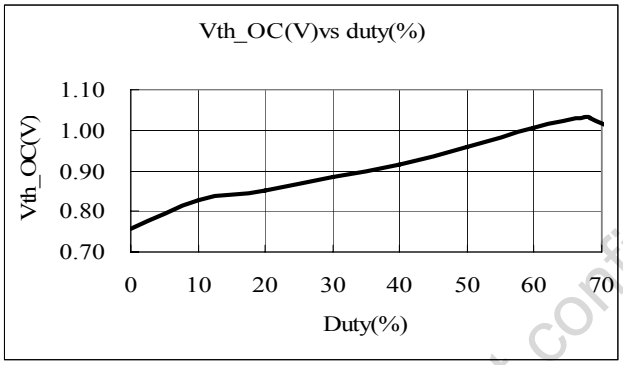
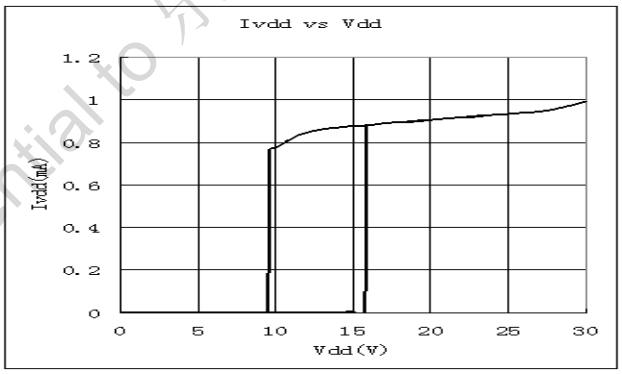
CHARACTERIZATION PLOTS

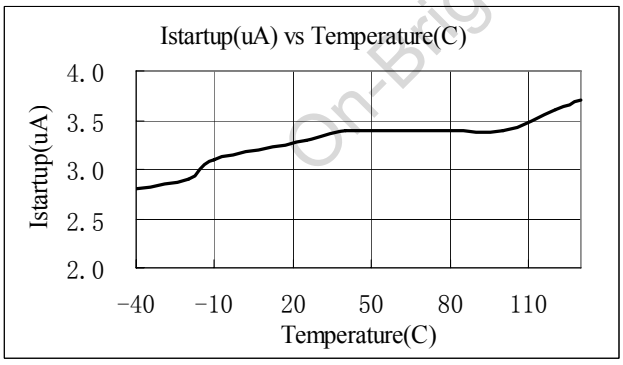
(The characteristic graphs are normalized at Ta=25℃)

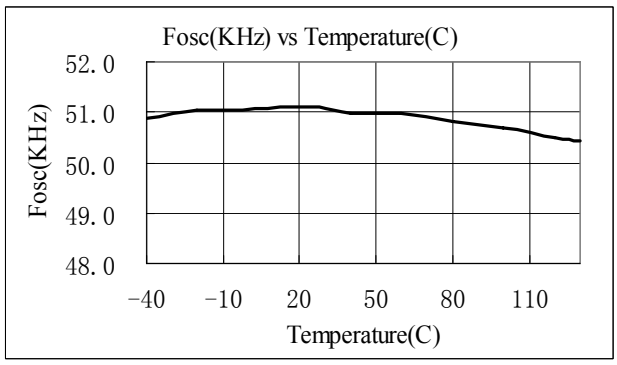


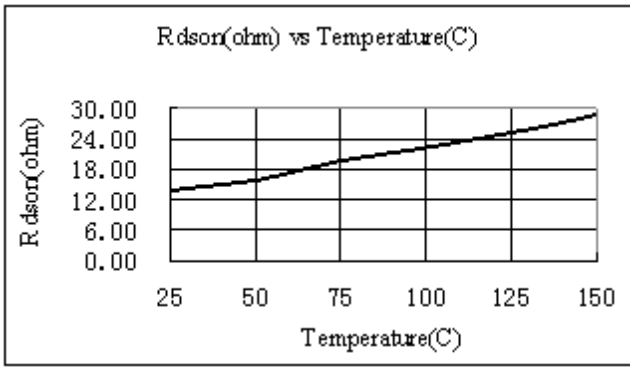












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OPERATION DESCRIPTION

The OB2353 is a low power off-line SMPS Switcher optimized for off-line flyback converter applications in sub 8W power range. The 'Extended burst mode' control greatly reduces the standby power consumption and helps the design easily to meet the international power conservation requirements.

- **Startup Current and Start up Control**

Startup current of OB2353 is designed to be very low so that VDD could be charged up above UVLO threshold level and device starts up quickly. A large value startup resistor can therefore be used to minimize the power loss yet achieve a reliable startup in application. For AC/DC adaptor with universal input range design, a 2 MΩ, 1/8 W startup resistor could be used together with a VDD capacitor to provide a fast startup and yet low power dissipation design solution.

- **Operating Current**

The Operating current of OB2353 is low at 2mA. Good efficiency is achieved with OB2353 low operating current together with the 'Extended burst mode' control features.

- **Soft Start**

OB2353 features an internal 4ms soft start to soften the electrical stress occurring in the power supply during startup. It is activated during the power on sequence. As soon as VDD reaches UVLO(OFF), the peak current is gradually increased from nearly zero to the maximum level of 0.77V. Every restart up is followed by a soft start.

- **Frequency shuffling for EMI improvement**

The frequency Shuffling (switching frequency modulation) is implemented in OB2353. The oscillation frequency is modulated so that the tone energy is spread out. The spread spectrum minimizes the conduction band EMI and therefore eases the system design.

- **Extended Burst Mode Operation**

At light load or zero load condition, most of the power dissipation in a switching mode power supply is from switching loss on the mosfet, the core loss of the transformer and the loss on the snubber circuit. The magnitude of power loss is in proportion to the switching frequency. Lower switching frequency leads to the reduction on the power loss and thus conserves the energy.

The switching frequency is internally adjusted at no load or light load condition. The switch frequency reduces at light/no load condition to improve the conversion efficiency. At light load or no load condition, the FB input drops below burst mode threshold level and device enters Burst Mode control. The Gate drive output switches only when VDD voltage drops below a preset level and FB input is active to output an on state. Otherwise the gate drive remains at off state to minimize the switching loss and reduces the standby power consumption to the greatest extend.
The switching frequency control also eliminates the audio noise at any loading conditions.

- **Oscillator Operation**

The switching frequency of OB2353 is internally fixed at 50KHZ. No external frequency setting components are required for PCB design simplification.

- **Current Sensing and Leading Edge Blanking**

Cycle-by-Cycle current limiting is offered in OB2353 current mode PWM control. The switch current is detected by a sense resistor into the sense pin. An internal leading edge blanking circuit chops off the sensed voltage spike at initial internal power MOSFET on state due to snubber diode reverse recovery and surge gate current of internal power MOSFET so that the external RC filtering on sense input is no longer needed. The current limiting comparator is disabled and cannot turn off the internal power MOSFET during the blanking period. The PWM duty cycle is determined by the current sense input voltage and the FB input voltage.

- **Internal Synchronized Slope Compensation**

Built-in slope compensation circuit adds voltage ramp onto the current sense input voltage for PWM generation. This greatly improves the close loop stability at CCM and prevents the sub-harmonic oscillation and thus reduces the output ripple voltage.

- **Drive**

The internal power MOSFET in OB2353 is driven by a dedicated gate driver for power switch control. Too weak the gate drive strength results in higher conduction and switch loss of MOSFET while too strong gate drive results the compromise of EMI

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A good tradeoff is achieved through the built-in totem pole gate design with right output strength and dead time control. The low idle loss and good EMI system design is easier to achieve with this dedicated control scheme.

In addition to the gate drive control scheme mentioned, the gate drive strength can also be adjusted externally by a resistor connected between VDD and VDDG, the falling edge of the Drain output can be well controlled. It provides great flexibility for system EMI design.

- **Protection Controls**

Good power supply system reliability is achieved with its rich protection features including Cycle-by-Cycle current limiting (OCP), Over Load Protection (OLP) and over voltage clamp, Under Voltage Lockout on VDD (UVLO).

With On-Bright Proprietary technology, the OCP is line voltage compensated to achieve constant output power limit over the universal input voltage range.

At overload condition when FB input voltage exceeds power limit threshold value for more than TD_PL, control circuit reacts to shut down the switcher. Switcher restarts when VDD voltage drops below UVLO limit. Similarly, control circuit shutdowns the power MOSFET when an Over Temperature condition is detected.

VDD is supplied by transformer auxiliary winding output. It is clamped when VDD is higher than 30V. The output of OB2353 is shut down when VDD drops below UVLO_ON limit and Switcher enters power on start-up sequence thereafter.

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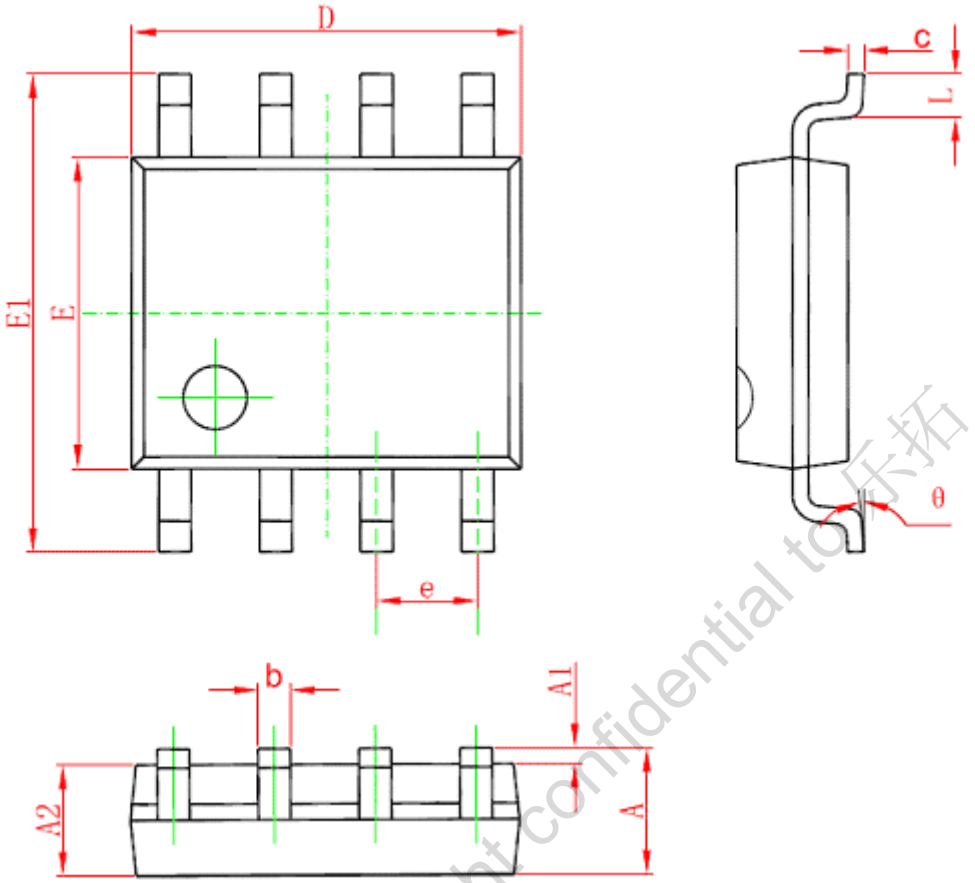
PACKAGE MECHANICAL DATA

8-Pin Plastic SOP



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Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.050	0.250	0.002	0.010
A2	1.250	1.650	0.049	0.065
b	0.310	0.510	0.012	0.020
c	0.100	0.250	0.004	0.010
D	4.700	5.150	0.185	0.203
E	3.800	4.000	0.150	0.157
E1	5.800	6.200	0.228	0.244
e	1.270 (BSC)		0.050 (BSC)	
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°

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