



The Future of Analog IC Technology®

MP2459

0.5A, 55V, 480kHz

Step-Down Converter in a TSOT23-6

DESCRIPTION

The MP2459 is a monolithic, step-down, switch-mode converter with a built-in power MOSFET. It achieves a 0.5A peak-output current over a wide input supply range with excellent load and line regulation. Current-mode operation provides a fast transient response and eases loop stabilization. Fault condition protections include cycle-by-cycle current limiting and thermal shutdown.

The MP2459 requires a minimal number of readily-available external components. The MP2459 is available in a TSOT23-6 package.

FEATURES

- 0.5A Peak Output Current
- 1Ω Internal Power MOSFET
- Stable with Low-ESR Ceramic Output Capacitors
- Up to 90% Efficiency
- 0.1μA Shutdown Mode
- Fixed 480kHz Frequency
- Thermal Shutdown
- Cycle-by-Cycle Over-Current Protection
- Wide 4.5V-to-55V Operating Input Range
- Output Adjustable from 0.81V to 0.95*V_{IN}
- Available in a TSOT23-6 Package

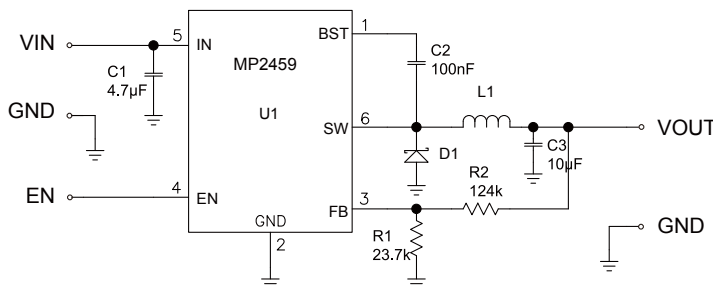
APPLICATIONS

- Power Meters
- Distributed Power Systems
- Battery Chargers
- Pre-Regulator for Linear Regulators
- WLED Drivers

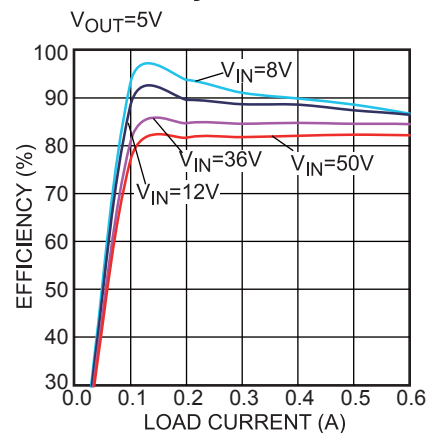
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TYPICAL APPLICATION



Efficiency vs. Load Current

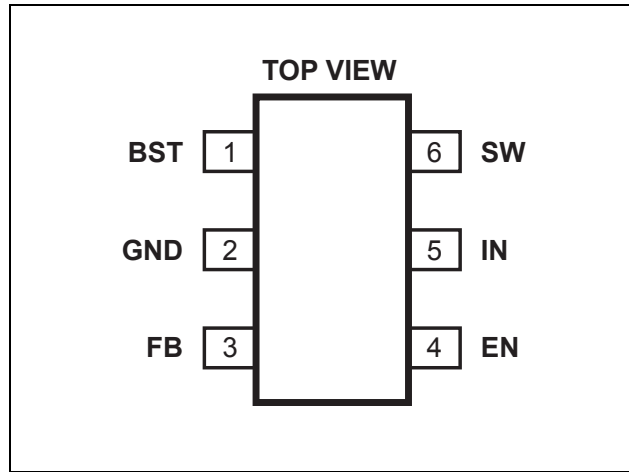


ORDERING INFORMATION

| Part Number* | Package | Top Marking |
|--------------|----------|-------------|
| MP2459GJ | TSOT23-6 | AEQ |

* For Tape & Reel, add suffix -Z (eg. M2459GJ-Z);

PACKAGE REFERENCE



ABSOLUTE MAXIMUM RATINGS ⁽¹⁾

| | |
|---|------------------------|
| Supply Voltage V_{IN} | -0.3V to 60V |
| V_{SW} | -0.3V to $V_{IN}+0.3V$ |
| V_{BS} | $V_{SW} + 6V$ |
| All Other Pins..... | -0.3V to +6V |
| Continuous Power Dissipation ($T_A = +25^\circ C$) ⁽²⁾ | |
| TSOT23-6..... | 0.568W |
| Junction Temperature..... | 150°C |
| Lead Temperature | 260°C |
| Storage Temperature..... | -65°C to +150°C |

Recommended Operating Conditions ⁽³⁾

| | |
|--------------------------------|-------------------------------|
| Supply Voltage V_{IN} | 4.5V to 55V |
| Output Voltage V_{OUT} | 0.81V to $0.95 \times V_{IN}$ |
| Operating Junction Temp..... | -40°C to +125°C |

| Thermal Resistance ⁽⁴⁾ | θ_{JA} | θ_{JC} |
|-----------------------------------|---------------|---------------|
| TSOT23-6 | 220 ... | 110 °C/W |

Notes:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature $T_J(MAX)$, the junction-to-ambient thermal resistance θ_{JA} , and the ambient temperature T_A . The maximum allowable continuous power dissipation at any ambient temperature is calculated by $P_D(MAX) = (T_J(MAX) - T_A) / \theta_{JA}$. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- 3) The device function is not guaranteed outside of the recommended operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB..

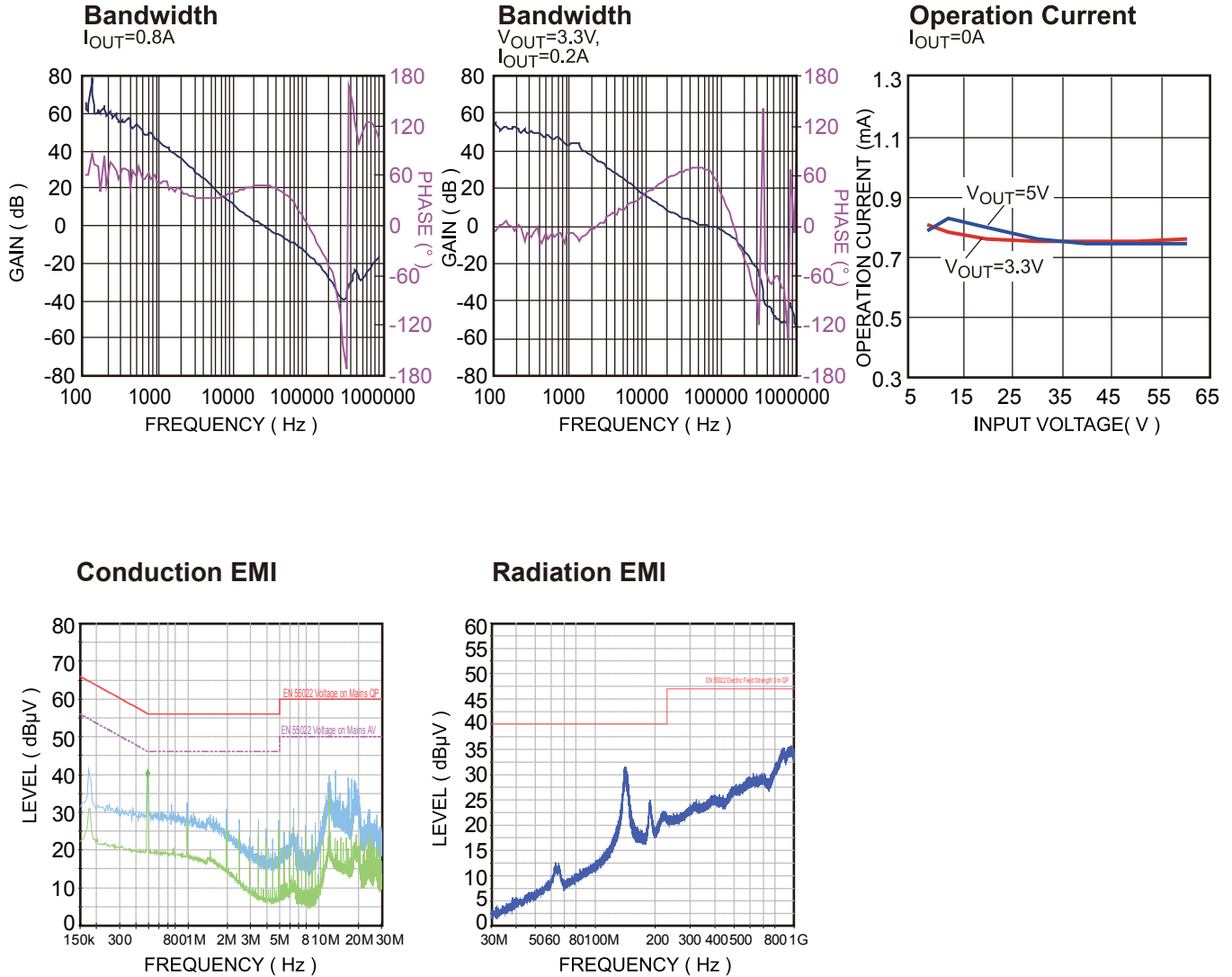
ELECTRICAL CHARACTERISTICS

$V_{IN} = 12V$, $T_A = +25^{\circ}C$, unless otherwise noted.

| Parameters | Symbol | Condition | Min | Typ | Max | Units |
|---|-----------------|-----------------------------|-------|-------|-------|-------------|
| Feedback Voltage | V_{FB} | $4.5V \leq V_{IN} \leq 55V$ | 0.792 | 0.812 | 0.832 | V |
| Feedback Current | I_{FB} | $V_{FB} = 0.85V$ | | | 0.1 | μA |
| Switch-On Resistance | $R_{DS(ON)}$ | | | 1 | | Ω |
| Switch Leakage | I_{SW_LKG} | $V_{EN} = 0V, V_{SW} = 0V$ | | | 1 | μA |
| Current Limit | I_{LIM} | | 1.0 | 1.25 | 1.5 | A |
| Oscillator Frequency | f_{SW} | $V_{FB} = 0.6V$ | 380 | 480 | 580 | kHz |
| Foldback Frequency | f_{SW_F} | $V_{FB} = 0V$ | | 150 | | kHz |
| Maximum Duty Cycle | D_{MAX} | $V_{FB} = 0.6V$ | 90 | 93.5 | | % |
| Minimum ON-Time | τ_{ON} | | | 100 | | ns |
| Under-Voltage Lockout Threshold, Rising | V_{UVLO_R} | | 2.9 | 3.3 | 3.7 | V |
| Under-Voltage Lockout Threshold, Falling | V_{UVLO_F} | | 2.65 | 3.05 | 3.45 | V |
| Under-Voltage Lockout Threshold, Hysteresis | V_{UVLO_HYS} | | | 250 | | mV |
| EN Threshold, Rising | V_{EN_R} | | | 1.35 | | V |
| EN Threshold, Falling | V_{EN_F} | | | 1.17 | | V |
| EN Threshold, Hysteresis | V_{EN_HYS} | | | 180 | | mV |
| EN Input Current | I_{EN} | $V_{EN} = 2V$ | | 3.1 | | μA |
| | | $V_{EN} = 0V$ | | 0.1 | | |
| Supply Current (Shutdown) | I_S | $V_{EN} = 0V$ | | 0.1 | 1.0 | μA |
| Supply Current (Quiescent) | I_Q | $V_{EN} = 2V, V_{FB} = 1V$ | | 0.73 | 0.85 | mA |
| Thermal Shutdown | T_{SD} | | | 165 | | $^{\circ}C$ |
| Thermal Shutdown Hysteresis | T_{SD_HYS} | | | 20 | | $^{\circ}C$ |

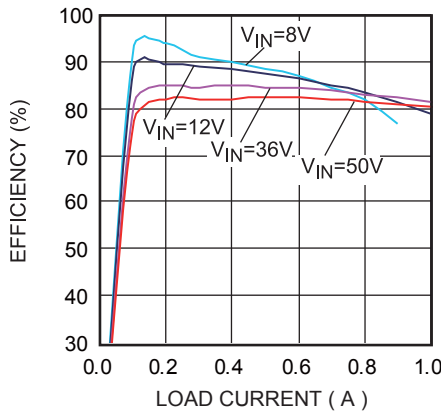
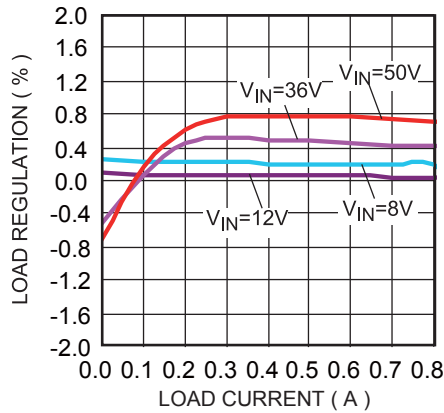
TYPICAL CHARACTERISTICS

$V_{IN}=12V$, $V_{OUT}=5V$, $I_{OUT}=0.5A$, $L=15\mu H$, $T_A=25^\circ C$, unless otherwise noted.

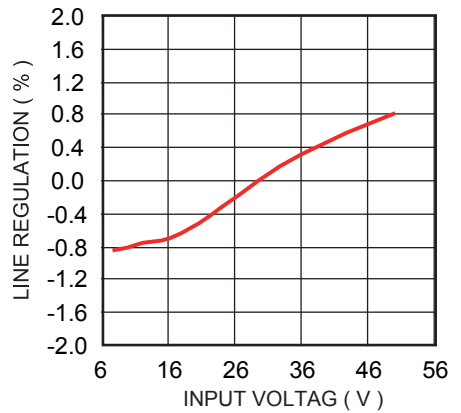


TYPICAL PERFORMANCE CHARACTERISTICS

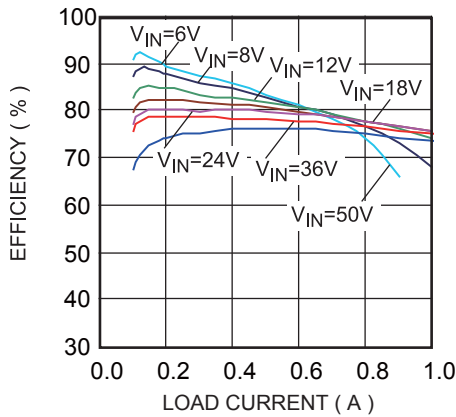
$V_{IN}=12V$, $V_{OUT}=5V$, $L=22\mu H$, $T_A=25^\circ C$, unless otherwise noted.

Efficiency vs. Load

Load Regulation

Line Regulation

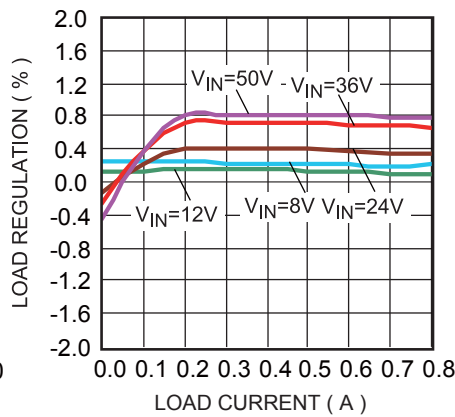
$I_{OUT}=0.5A$


Efficiency vs. Load

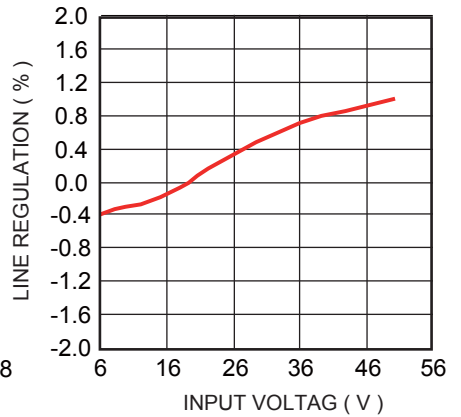
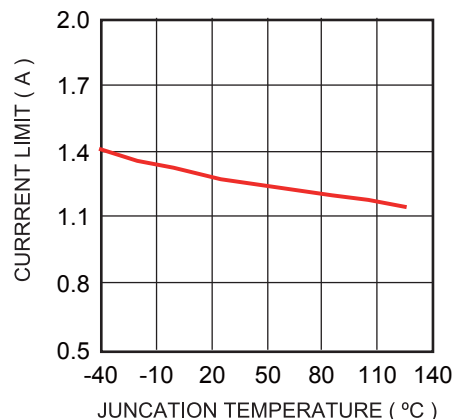
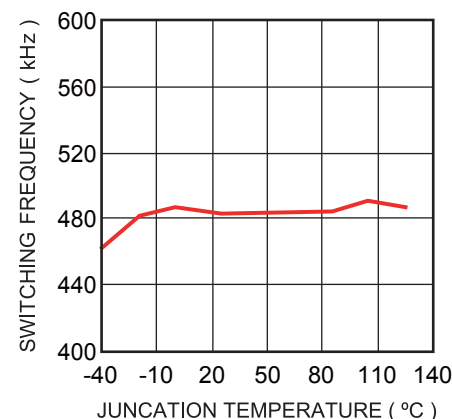
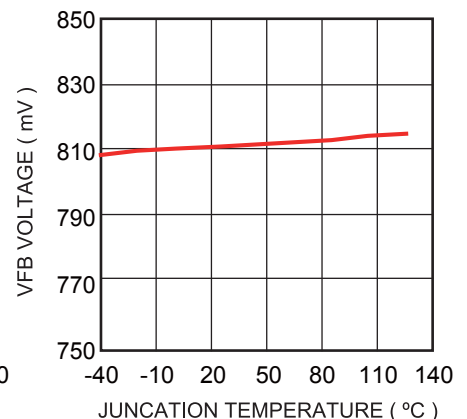
$V_{OUT}=3.3V$


Load Regulation

$V_{OUT}=3.3V$


Line Regulation

$V_{OUT}=3.3V$, $I_{OUT}=0.5A$

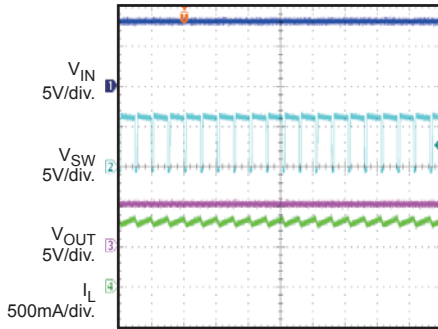

Current Limit vs. T_J

Frequency vs. T_J

Feedback Voltage vs. T_J


TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN}=12V$, $V_{OUT}=5V$, $L=22\mu H$, $T_A=25^\circ C$, unless otherwise noted.

Steady State

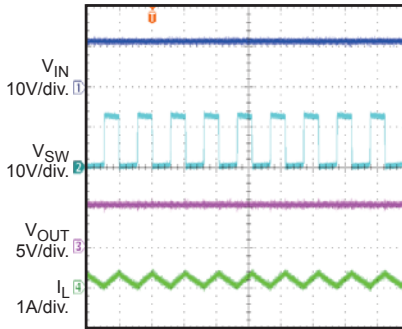
$V_{IN}=8V$, $I_{OUT}=0.8A$



4µs/div.

Steady State

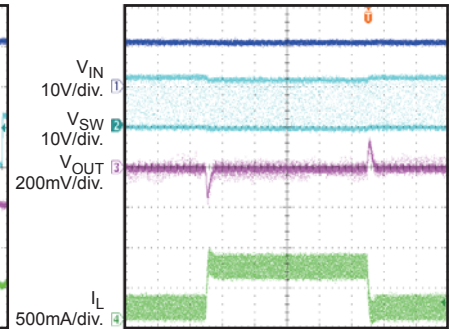
$I_{OUT}=0.15A$



2µs/div.

Load Transient

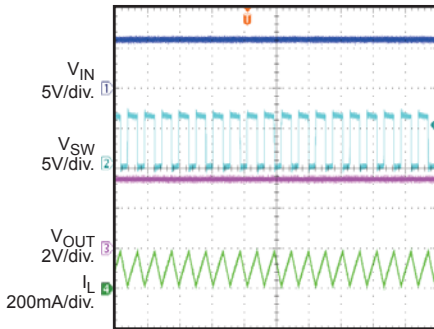
$I_{OUT}=0.2A$ to $0.7A$



200µs/div.

Steady State

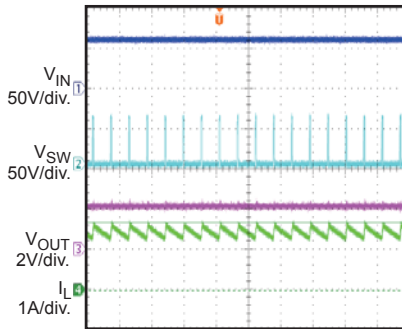
$V_{IN}=6V$, $V_{OUT}=3.3V$, $I_{OUT}=0.1A$



4µs/div.

Steady State

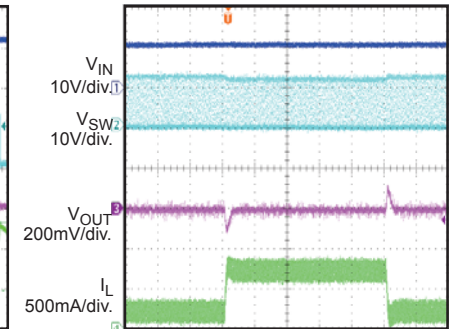
$V_{IN}=60V$, $V_{OUT}=3.3V$, $I_{OUT}=0.9A$



4µs/div.

Load Transient

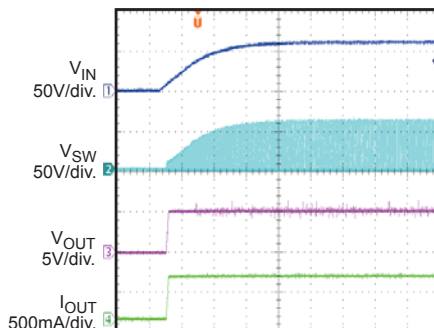
$V_{OUT}=3.3V$, $I_{OUT}=0.2A$ to $0.7A$



200µs/div.

Power Ramp Up

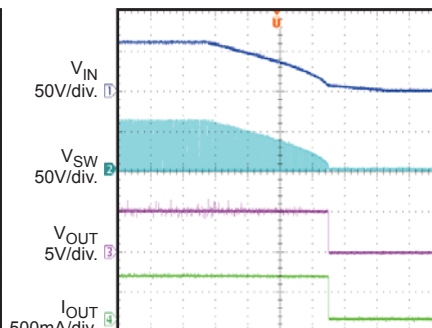
$V_{IN}=60V$, $I_{OUT}=0.5A$



10ms/div.

Power Ramp Down

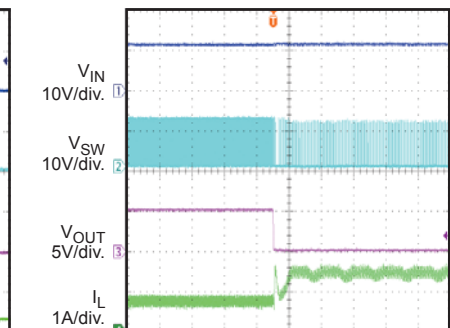
$V_{IN}=60V$, $I_{OUT}=0.5A$



20ms/div.

Short Output

$I_{OUT}=0.65A$

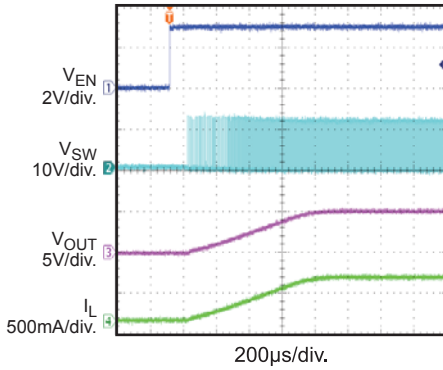


200µs/div.

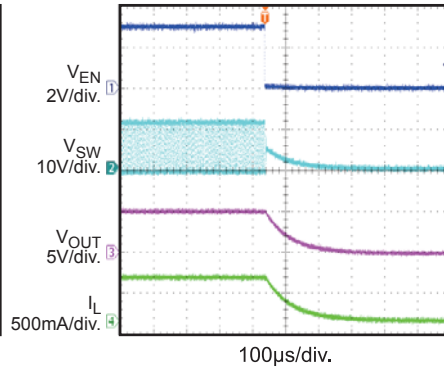
TYPICAL PERFORMANCE CHARACTERISTICS *(continued)*

$V_{IN}=12V$, $V_{OUT}=5V$, $L=22\mu H$, $T_A=25^\circ C$, unless otherwise noted.

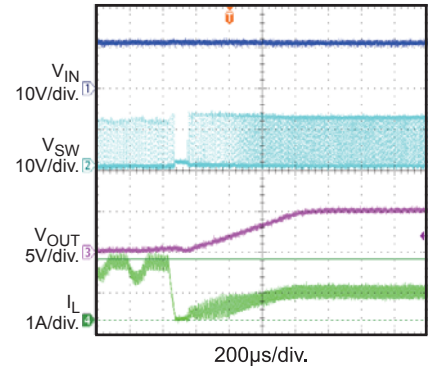
Enable On
 $I_{OUT}=0.5A$



Enable Off
 $I_{OUT}=0.5A$



Short Output Recovery
 $I_{OUT}=0.65A$



PIN FUNCTIONS

| Pin # | Name | Description |
|-------|------|---|
| 1 | BST | Bootstrap. Connect a capacitor between the SW and BS pins to form a floating supply across the power switch driver. This capacitor drives the power switch's gate above the supply voltage. |
| 2 | GND | Ground. Voltage reference for the regulated output voltage. Requires special layout considerations. Isolate this node from the D1 to C1 ground path to prevent switching current spikes from inducing. |
| 3 | FB | Feedback. Sets the output voltage. Connect to the tap of an external resistor divider from the output to GND. The frequency foldback comparator lowers the oscillator frequency when the FB voltage is below 250mV to prevent current-limit runaway during a short-circuit fault. |
| 4 | EN | On/Off. Pull EN above 1.2V to turn the device ON. For automatic enable, connect to V_{IN} using a 100k Ω resistor. |
| 5 | IN | Supply Voltage. The MP2459 operates from a 4.5V-to-55V unregulated input. Requires C1 to prevent large voltage spikes from appearing at the input. |
| 6 | SW | Switch Output. |

OPERATION

The MP2459 is a current mode buck regulator. That is, the EA output voltage is proportional to the peak inductor current.

At the beginning of a cycle, M1 is off. The EA output voltage is higher than the current sense amplifier output, and the current comparator's output is low. The rising edge of the 480kHz CLK signal sets the RS Flip-Flop. Its output turns on M1 thus connecting the SW pin and inductor to the input supply.

The increasing inductor current is sensed and amplified by the Current Sense Amplifier. Ramp compensation is summed to the Current Sense Amplifier output and compared to the Error Amplifier output by the PWM Comparator. When the sum of the Current Sense Amplifier output and the Slope Compensation signal exceeds the EA output voltage, the RS Flip-Flop is reset and M1 is turned off. The external Schottky rectifier diode (D1) conducts the inductor current.

If the sum of the Current Sense Amplifier output and the Slope Compensation signal does not exceed the EA output for a whole cycle, then the falling edge of the CLK resets the Flip-Flop.

The output of the Error Amplifier integrates the voltage difference between the feedback and the 0.81V bandgap reference. The polarity is such that lower than 0.81V FB pin voltage increases the EA output voltage. Since the EA output voltage is proportional to the peak inductor current, an increase in its voltage also increases current delivered to the output.

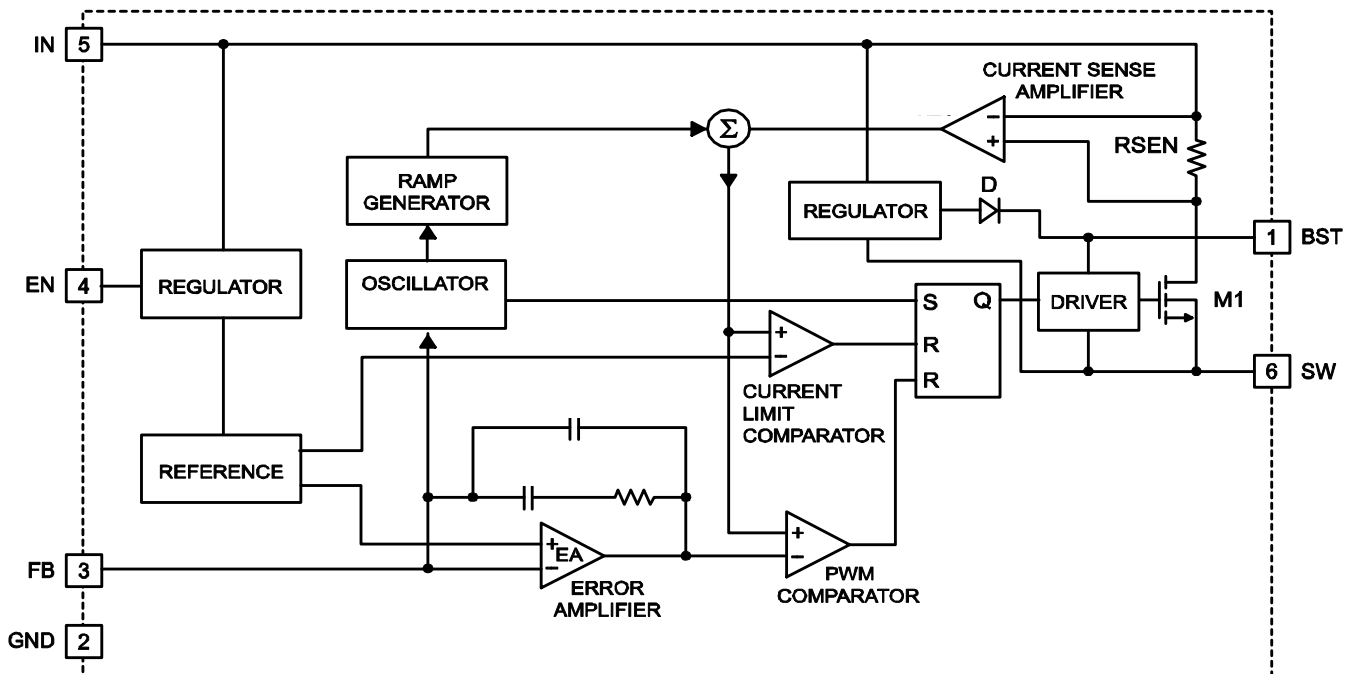


Figure 1: Functional Block Diagram

APPLICATION INFORMATION

Setting Output Voltage

The external resistor divider sets the output voltage (see the Typical Application schematic). Table 1 lists resistors for common output voltages. The feedback resistor (R2) also sets the feedback loop bandwidth with the internal compensation capacitor (see Figure 1). R1 is:

$$R1 = \frac{R2}{\frac{V_{OUT}}{0.812V} - 1}$$

Table 1: Resistor Selection for Common Output Voltages

| V _{OUT} (V) | R1 (kΩ) | R2 (kΩ) |
|----------------------|-----------|----------|
| 1.8 | 102 (1%) | 124 (1%) |
| 2.5 | 59 (1%) | 124 (1%) |
| 3.3 | 40.2 (1%) | 124 (1%) |
| 5 | 23.7 (1%) | 124 (1%) |

Selecting the Inductor

Use an inductor with a DC current rating at least 25% percent higher than the maximum load current for most applications. For best efficiency, the inductor's DC resistance should be less than 200mΩ. Refer to Table 2 for suggested surface-mount inductors.

Table 2: Inductor Selection Guide

| Part Number | Inductance (μH) | Max DCR (Ω) | Current Rating (A) | Dimensions L x W x H (mm ³) |
|---------------------|-----------------|-------------|--------------------|---|
| Würth | | | | |
| 744042100 | 10 | 0.15 | 1.3 | 4.8x4.8x1.8 |
| 744052150 | 15 | 0.19 | 0.95 | 5.8x5.8x1.8 |
| 744053220 | 22 | 0.12 | 1.15 | 5.8x5.8x2.8 |
| ABC | | | | |
| SH4018100YL | 10 | 0.15 | 1.3 | 4.8x4.8x1.8 |
| SH4028150YL | 15 | 0.15 | 1.03 | 4.8x4.8x2.8 |
| SH5028220YL | 22 | 0.12 | 1.15 | 5.8x5.8x2.8 |
| TDK | | | | |
| VLCF5020T-100MR87 | 10 | 0.182 | 1.56 | 5x5x2.0 |
| VLCF5024T-150MR71-2 | 15 | 0.152 | 1.66 | 5x5x2.4 |
| VLCF5028T-220MR71-2 | 22 | 0.14 | 1.66 | 5x5x2.8 |
| Coilcraft | | | | |
| LPS5030-103MRC | 10 | 0.127 | 1.5 | 5x5x3 |
| LPS5030-153MRC | 15 | 0.16 | 1.4 | 5x5x3 |
| LPS5030-223MRC | 22 | 0.19 | 1.25 | 5x5x3 |

For most designs, the required inductance value can be derived from the following equation.

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_L \times f_{SW}}$$

Where ΔI_L is the inductor ripple current.

Choose the inductor ripple current to be 30% of the maximum load current. The maximum inductor peak current is:

$$I_{L(MAX)} = I_{LOAD} + \frac{\Delta I_L}{2}$$

Under light-load conditions (below 100mA), use a larger inductance to improve efficiency.

Selecting the Input Capacitor

The input capacitor reduces the surge current drawn from the input supply and the switching noise from the device. The input capacitor impedance at the switching frequency should be less than the input source impedance to prevent high-frequency-switching current from passing through the input. Use ceramic capacitors with X5R or X7R dielectrics for their low ESRs and small temperature coefficients. For most applications, a 4.7 μ F capacitor will sufficient.

Selecting the Output Capacitor

The output capacitor keeps the output voltage ripple small and ensures feedback loop stability. The output capacitor impedance should be low at the switching frequency. Use ceramic capacitors with X5R or X7R dielectrics for their low ESR characteristics. For most applications, a 22 μ F ceramic capacitor will sufficient.

PCB Layout Guide

PCB layout is very important to stability. Please follow these guidelines and use Figure 2 as reference.

- 1) Keep the path of switching current short and minimize the loop area formed by the input capacitor, high-side MOSFET, and Schottky diode.
- 2) Keep the connection from the power ground \rightarrow Schottky diode \rightarrow SW pin as short and wide as possible.

- 3) Ensure all feedback connections are short and direct. Place the feedback resistors and compensation components as close to the chip as possible.
- 4) Route SW away from sensitive analog areas such as FB.
- 5) Connect IN, SW, and especially GND to large copper areas to cool the chip for improved thermal performance and long-term reliability. For single layer PCBs, avoid soldering the exposed pad

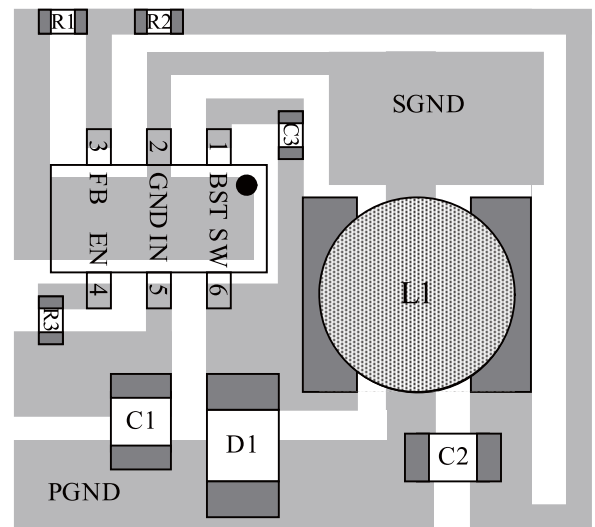


Figure 2: PCB Layout

External Bootstrap Diode

An external bootstrap diode may enhance regulator efficiency under the following conditions:

- $V_{OUT}=5V$ or $3.3V$; and
- High duty cycle: $D = \frac{V_{OUT}}{V_{IN}} > 65\%$

In these cases, add an external BST diode from the output of the voltage regulator to the BST pin, as shown in Figure 3.

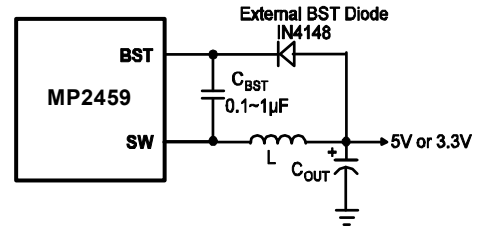
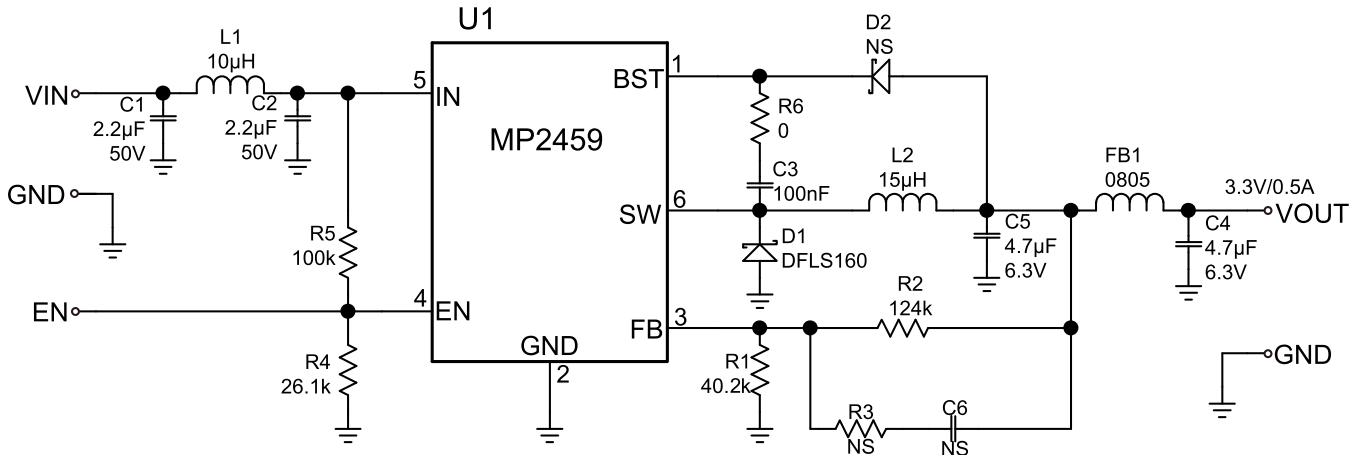


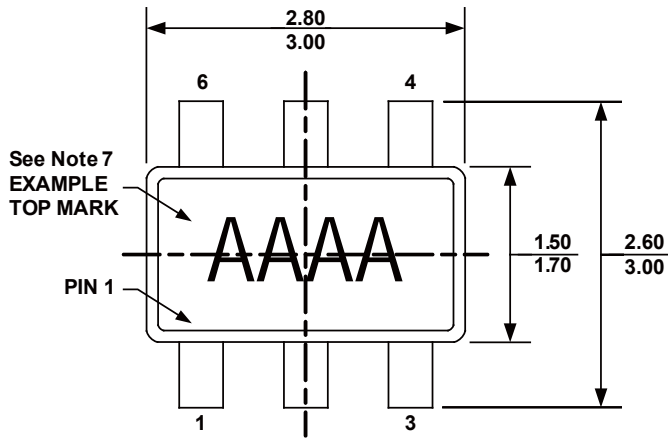
Figure 3: Optional Bootstrap Diode for Enhanced Efficiency

The recommended external BST diode is IN4148, and the BST capacitor is $0.1\mu F-1\mu F$.

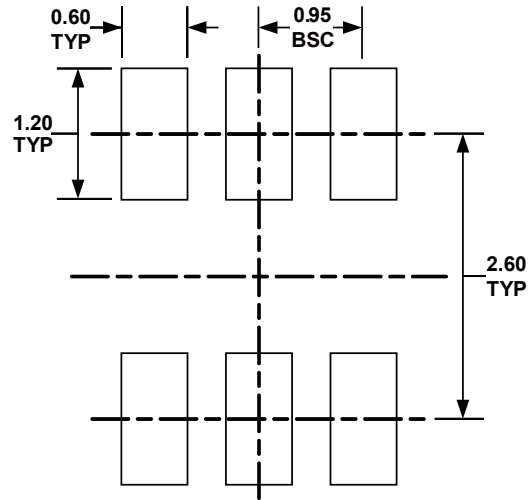
TYPICAL APPLICATION CIRCUIT


PACKAGE INFORMATION

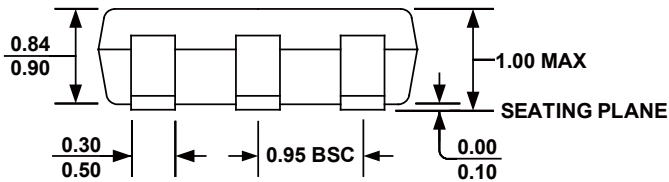
TSOT23-6



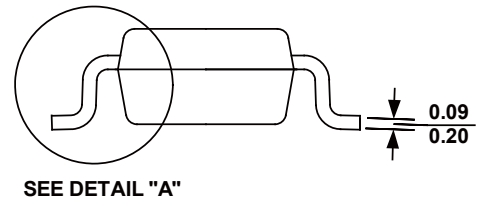
TOP VIEW



RECOMMENDED LAND PATTERN

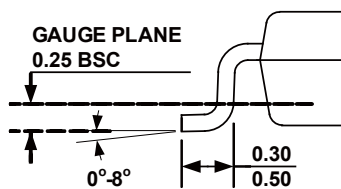


FRONT VIEW



SEE DETAIL "A"

SIDE VIEW



DETAIL "A"

NOTE:

- 1) ALL DIMENSIONS ARE IN MILLIMETERS
- 2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSION OR GATE BURR
- 3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION.
- 4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.10 MILLIMETERS MAX
- 5) DRAWING CONFORMS TO JEDEC MO-193, VARIATION AB
- 6) DRAWING IS NOT TO SCALE
- 7) PIN 1 IS LOWER LEFT PIN WHEN READING TOP MARK FROM LEFT TO RIGHT, (SEE EXAMPLE TOP MARK)

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