

TPS92610-Q1 汽车单通道线性 LED 驱动器

1 特性

- 符合汽车类应用要求
- 具有符合 AEC-Q100 标准的下列特性：
 - 温度等级 1: -40°C 至 125°C 的环境工作温度范围
 - 器件 HBM ESD 分类等级 H2
 - 器件 CDM ESD 分类等级 C3B
- 提供功能安全
 - 可帮助创建功能安全系统设计的文档
- 具有 PWM 调光功能的单通道恒定电流 LED 驱动器
- 宽输入电压范围: 4.5V - 40V
- 可由检测电阻器进行调节的恒定输出电流
- 精密电流调节, 在 -40°C 至 150°C 的结温范围内具有 $\pm 4.6\%$ 的容差
- 最大电流: 450mA
- 与外部电阻器实现热共享
- 低压降电压 (包含检测电阻器压降)
 - 最大压降: 10mA 时为 150mV
 - 最大压降: 70mA 时为 400mV
 - 最大压降: 150mA 时为 700mV
 - 最大压降: 300mA 时为 1.3V
- 诊断和保护
 - 单 LED 短路检测及自动恢复
 - LED 开路与短路检测及自动恢复
 - 在低压降运行情况下支持诊断并具有可调阈值
 - 多达 15 个器件的故障总线, 可配置为“连带失效”或“仅失效的通道关闭”
 - 较低的静态电流和故障模式电流 (每个器件小于 $250\mu\text{A}$)
- 工作结温范围: -40°C 至 150°C

2 应用

- 汽车便利照明: 座舱顶灯、车门把手、阅读灯和其他灯具
- 汽车尾灯、中央高位刹车灯、侧面标志灯、盲点监测指示灯、充电口指示灯
- 通用 LED 驱动器应用

3 说明

随着 LED 在汽车应用中广泛使用, 简单的 LED 驱动器越来越受欢迎。与分立式解决方案相比, 低成本单片解决方案可降低系统级组件数量, 并显著提高电流精度和可靠性。

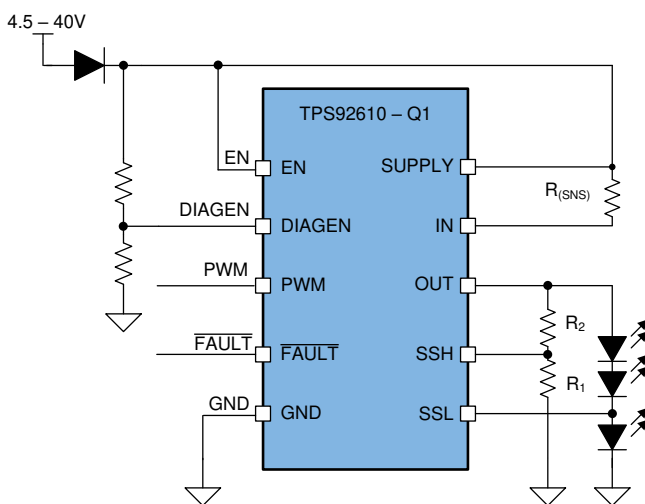
TPS92610-Q1 器件是一款简单的单通道高侧 LED 驱动器, 由汽车蓄电池供电。这是一种简单而巧妙的解决方案, 能够为单个 LED 灯串提供恒定电流, 并具有全面的 LED 诊断功能。此器件的“连带失效”功能可与其他 LED 驱动器 (如 TPS9261x-Q1、TPS9263x-Q1 和 TPS9283x-Q1 器件) 一起工作, 从而满足不同的要求。

器件信息⁽¹⁾

器件型号	封装	封装尺寸 (标称值)
TPS92610-Q1	HTSSOP (14)	5mm x 4.4mm

(1) 如需了解所有可用封装, 请参阅数据表末尾的可订购产品附录。

典型应用图



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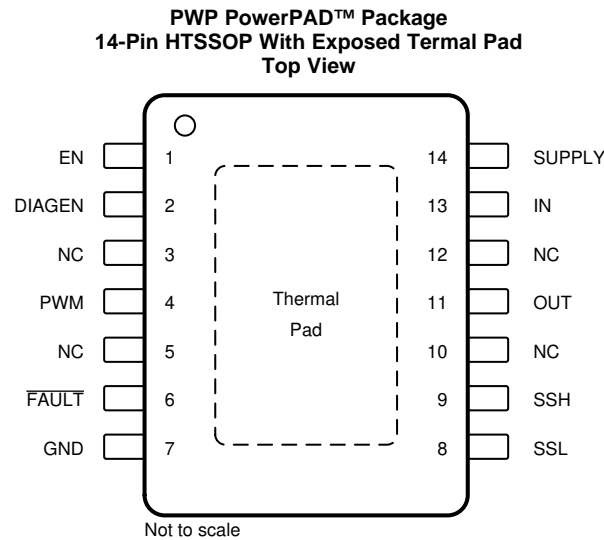
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4 修订历史记录

Changes from Revision A (December 2017) to Revision B	Page
• 向 特性 部分添加了提供功能安全的链接	1

Changes from Original (November 2017) to Revision A	Page
• 产品说明书从“预告信息”更改为“生产数据”	1

5 Pin Configuration and Functions



NC – No internal connection

Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
DIAGEN	2	I	Diagnostics enable, to avoid false open-circuit diagnostics during low-voltage operation
EN	1	I	Device enable
$\overline{\text{FAULT}}$	6	I/O	One-fails–all-fail fault bus
GND	7	—	Ground
IN	13	I	Current input
NC	3, 5, 10, 12	—	Not connected
OUT	11	O	Constant-current output
PWM	4	I	PWM input
SSH	9	I	Single-LED short high-side reference
SSL	8	I	Single-LED short low-side reference
SUPPLY	14	I	Device supply voltage

6 Specifications

6.1 Absolute Maximum Ratings

over operating ambient temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
High-voltage input	DIAGEN, EN, IN, PWM, SSH, SSL, SUPPLY	–0.3	45	V
High-voltage output	OUT	–0.3	45	V
Fault bus	$\overline{\text{FAULT}}$	–0.3	22	V
IN to OUT	$V_{(\text{IN})} - V_{(\text{OUT})}$	–0.3	45	V
SUPPLY to IN	$V_{(\text{SUPPLY})} - V_{(\text{IN})}$	–0.3	1	V
Operating junction temperature, T_J		–40	150	°C
Storage temperature, T_{stg}		–40	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

TPS92610-Q1			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 ⁽¹⁾	All pins	±2000
		Charged-device model (CDM), per AEC Q100-011	All pins	±500
			Corner pins (1, 7, 8, and 14)	±750

(1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

over operating ambient temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
SUPPLY	Device supply voltage	4.5		40	V
IN	Sense voltage	4.4		40	V
PWM	PWM input	0		40	V
DIAGEN	Diagnostics enable pin	0		40	V
OUT	Driver output	0		40	V
SSH	Single LED short high-side reference	0		5	V
SSL	Single LED short low-side reference	0		5	V
EN	Device enable	0		40	V
$\overline{\text{FAULT}}$	Fault bus	0		7	V
T_A	Operating ambient temperature	-40		125	°C

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		TPS92610-Q1	UNIT
		PWP (HTSSOP)	
		14 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	52.4	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	43.5	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	22	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	1.6	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	22.3	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	6.5	°C/W

(1) For more information about traditional and new thermal metrics, see [Semiconductor and IC Package Thermal Metrics](#).

6.5 Electrical Characteristics

$V_{(SUPPLY)} = 5\text{ V} - 40\text{ V}$, $T_J = -40^\circ\text{C} - 150^\circ\text{C}$ unless otherwise noted

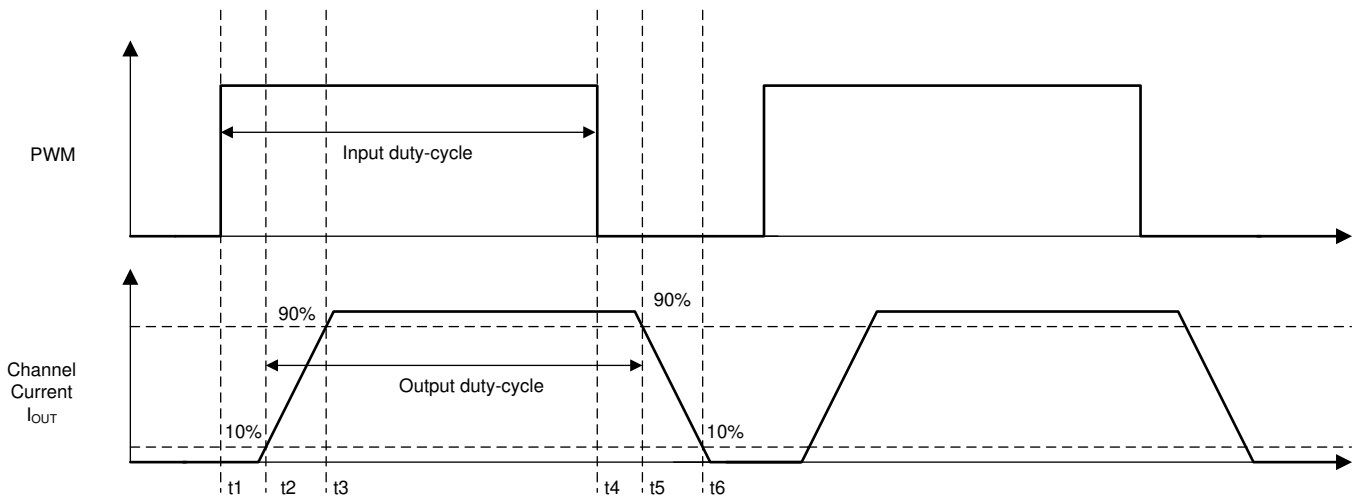
PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
BIAS						
$V_{(POR_rising)}$	Supply voltage POR rising threshold		3.2	4	V	
$V_{(POR_falling)}$	Supply voltage POR falling threshold	2.2	3		V	
$I_{(Shutdown)}$	Device shutdown current	EN = LOW	5	10	μA	
$I_{(Quiescent)}$	Device quiescent current	PWM = HIGH, EN = HIGH	0.1	0.2	0.25	mA
$I_{(FAULT)}$	Device current in fault mode	EN = HIGH, PWM = HIGH, $\overline{\text{FAULT}}$ externally pulled LOW	0.1	0.2	0.25	mA

Electrical Characteristics (continued)
 $V_{(SUPPLY)} = 5\text{ V} - 40\text{ V}$, $T_J = -40^\circ\text{C} - 150^\circ\text{C}$ unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
LOGIC INPUTS (DIAGEN, PWM, EN)						
$V_{IL(DIAGEN)}$	Input logic-low voltage, DIAGEN		1.045	1.1	1.155	V
$V_{IH(DIAGEN)}$	Input logic-high voltage, DIAGEN		1.14	1.2	1.26	V
$V_{IL(PWM)}$	Input logic-low voltage, PWM		1.045	1.1	1.155	V
$V_{IH(PWM)}$	Input logic-high voltage, PWM		1.14	1.2	1.26	V
$V_{IL(EN)}$	Input logic-low voltage, EN				0.7	V
$V_{IH(EN)}$	Input logic-high voltage, EN		2			V
$I_{PD(EN)}$	EN pin pulldown current	$V_{(EN)} = 12\text{ V}$	1.5	3.3	4.5	μA
CONSTANT-CURRENT DRIVER						
$I_{(OUT)}$	Device output-current range	100% duty-cycle	4		450	mA
$V_{(CS_REG)}$	Sense-resistor regulation voltage	$T_A = 25^\circ\text{C}$, $V_{(SUPPLY)} = 4.5\text{ V to }18\text{ V}$	94	98	102	mV
		$T_A = -40^\circ\text{C to }125^\circ\text{C}$, $V_{(SUPPLY)} = 4.5\text{ V to }18\text{ V}$	93.5	98	102.5	
$R_{(SNS)}$	Sense-resistor range				24.5	Ω
$V_{(DROPOUT)}$	Voltage dropout from SUPPLY to OUT	$V_{(CS_REG)}$ voltage included, current setting = 10 mA		120	150	mV
		$V_{(CS_REG)}$ voltage included, current setting = 70 mA		250	400	
		$V_{(CS_REG)}$ voltage included, current setting = 150 mA		430	700	
		$V_{(CS_REG)}$ voltage included, current setting = 300 mA		800	1300	
DIAGNOSTICS						
$V_{(OPEN_th_rising)}$	LED open rising threshold, $V_{(IN)} - V_{(OUT)}$		70	100	135	mV
$V_{(OPEN_th_falling)}$	LED open falling threshold, $V_{(IN)} - V_{(OUT)}$		235	290	335	mV
$V_{(SG_th_falling)}$	Channel output $V_{(OUT)}$ short-to-ground falling threshold		1.14	1.2	1.26	V
$V_{(SG_th_rising)}$	Channel output $V_{(OUT)}$ short-to-ground rising threshold		0.82	0.865	0.91	V
$I_{(Retry)}$	Channel output retry current	$V_{(OUT)} = 0\text{ V}$	0.64	1.08	1.528	mA
$V_{(SSH_th)}$	Single-LED short-detection high-side threshold	$V_{(SSL)} - V_{(SSH)}$	140	190	235	mV
$V_{(SSL_th)}$	Single-LED short-detection low-side threshold		0.8	0.86	0.91	V
FAULT						
$V_{IL(FAULT)}$	Logic-input low threshold				0.7	V
$V_{IH(FAULT)}$	Logic-input high threshold		2			V
$V_{OL(FAULT)}$	Logic-output low voltage	With 500- μA external pullup			0.4	V
$V_{OH(FAULT)}$	Logic-output high voltage	With 1- μA external pulldown, $V_{(SUPPLY)} = 12\text{ V}$	5		7	V
$I_{(FAULT_pulldown)}$	$\overline{\text{FAULT}}$ internal pulldown current		500	750	1000	μA
$I_{(FAULT_pullup)}$	$\overline{\text{FAULT}}$ internal pullup current		5	8	12	μA
THERMAL PROTECTION						
$T_{(TSD)}$	Thermal shutdown junction temperature threshold		167	172	178	$^\circ\text{C}$
$T_{(TSD_HYS)}$	Thermal shutdown junction temperature hysteresis			15		$^\circ\text{C}$

6.6 Timing Requirements

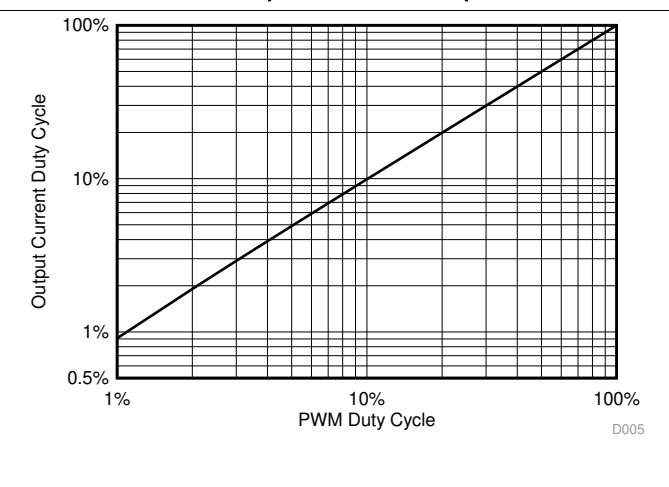
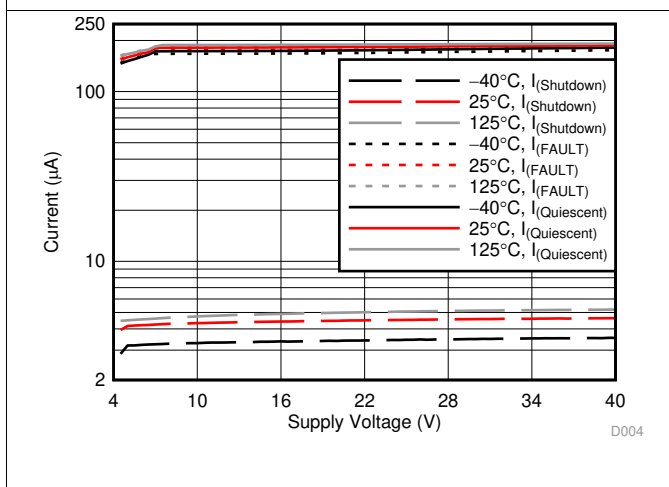
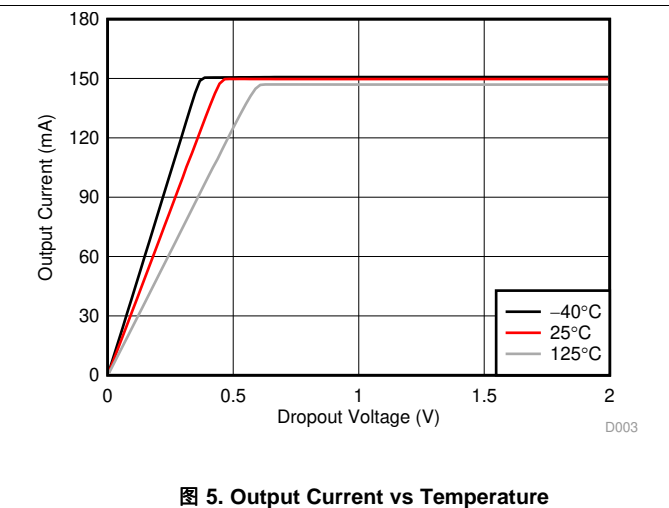
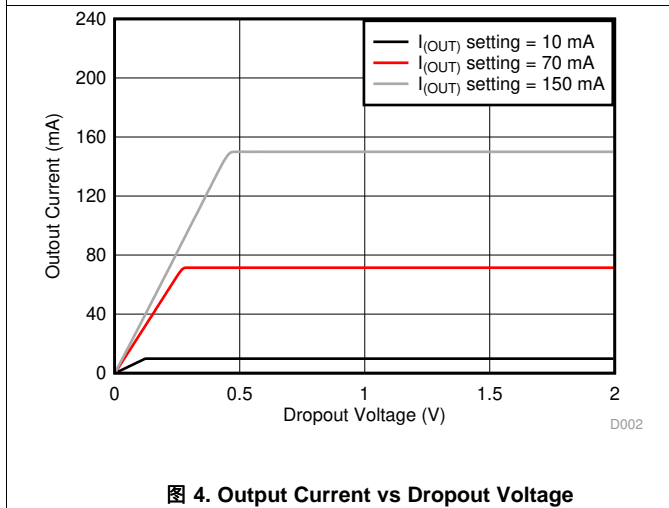
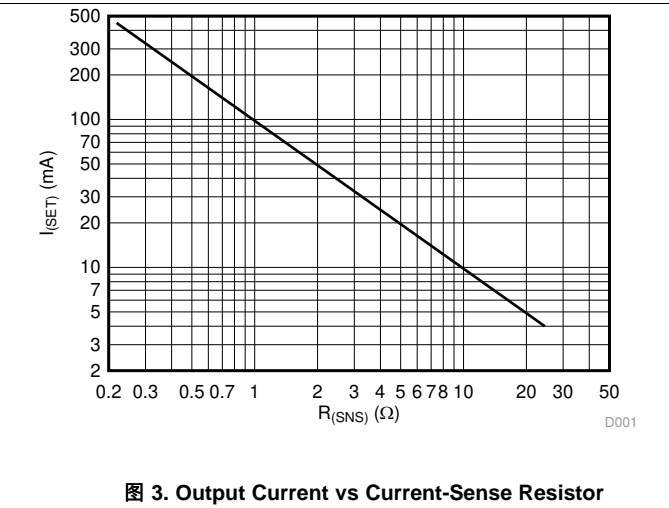
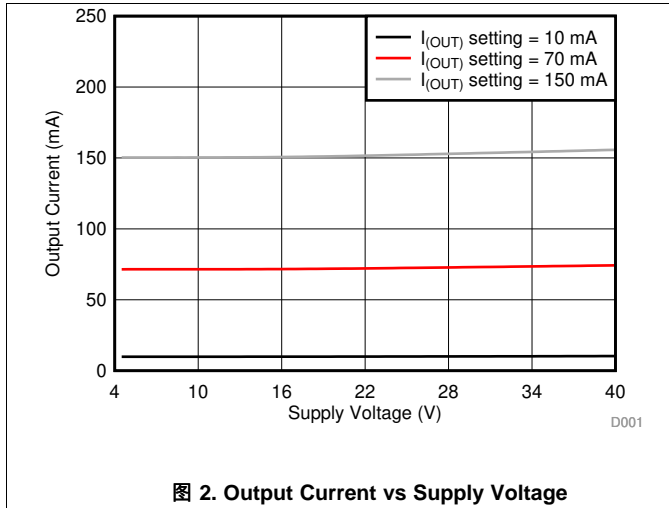
		MIN	NOM	MAX	UNIT
$t_{(PWM_delay_rising)}$	PWM rising edge delay, 50% PWM voltage to 10% of output current, $t_2 - t_1$ as shown in 图 1	10	17	25	μs
$t_{(PWM_delay_falling)}$	PWM falling edge delay, 50% PWM voltage to 90% of output current, $t_5 - t_4$ as shown in 图 1	15	21	30	μs
$t_{(TSD_deg)}$	Thermal overtemperature deglitch time		60		μs
$t_{(DEVICE_STARTUP)}$	EN rising edge to 10% output current at 150-mA set current and 12-V supply voltage		100	150	μs
$t_{(OPEN_deg)}$	LED open-circuit fault-deglitch time	80	125	175	μs
$t_{(SG_deg)}$	Channel-output short-to-ground detection deglitch time	80	125	175	μs
$t_{(SS_deg)}$	Single-LED short-detection deglitch time	80	125	175	μs
$t_{(Recover_deg)}$	Recovery deglitch time		16		μs



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图 1. Output Timing Diagram

6.7 Typical Characteristics



Typical Characteristics (接下页)

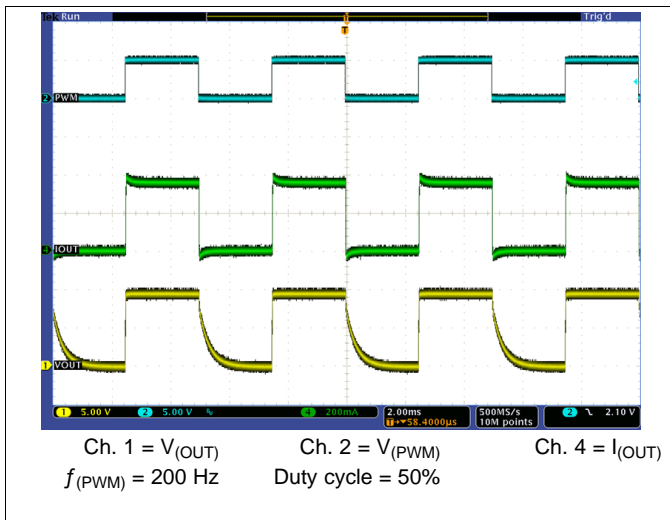


图 8. PWM Dimming via External Input

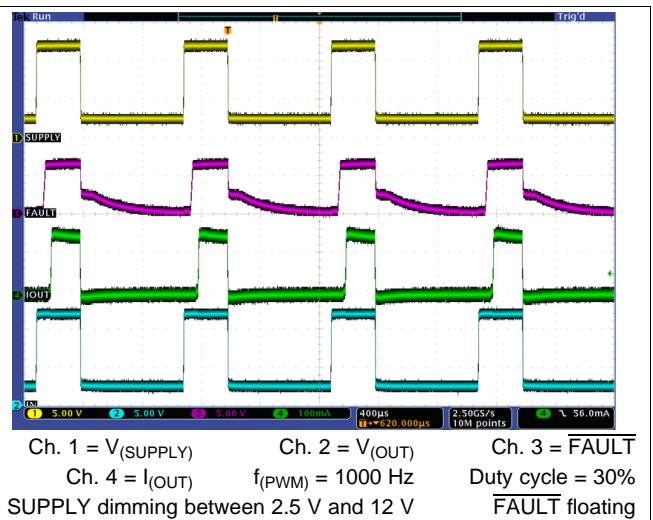


图 9. PWM Dimming via Power Supply

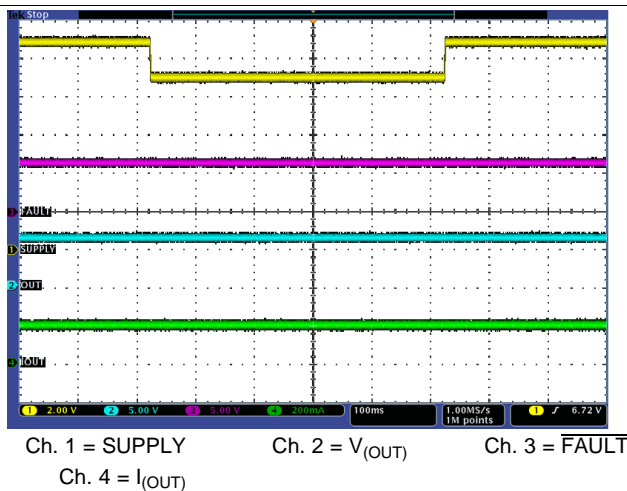


图 10. Transient Undervoltage

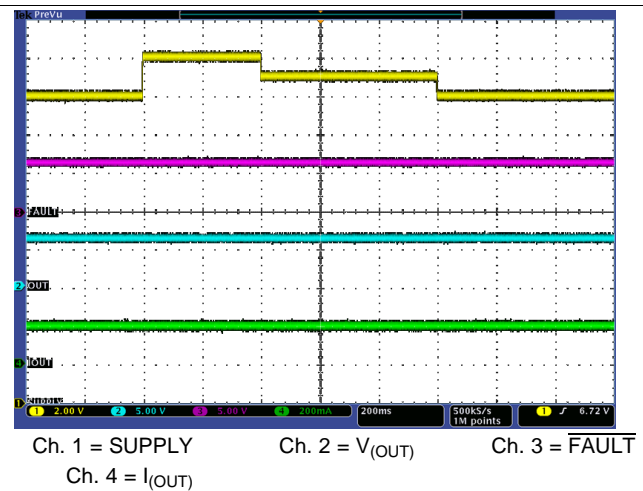


图 11. Transient Overvoltage

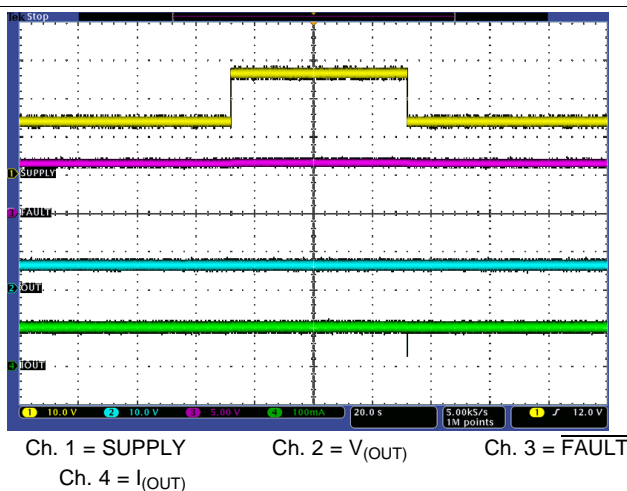


图 12. Jump Start

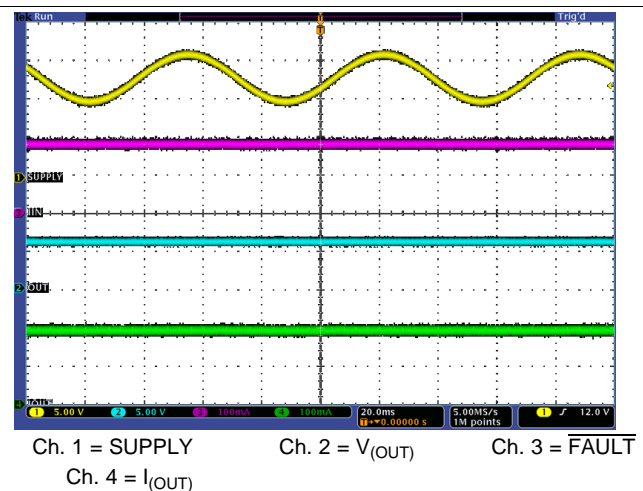


图 13. Superimposed Alternating Voltage, 15 Hz

Typical Characteristics (接下页)

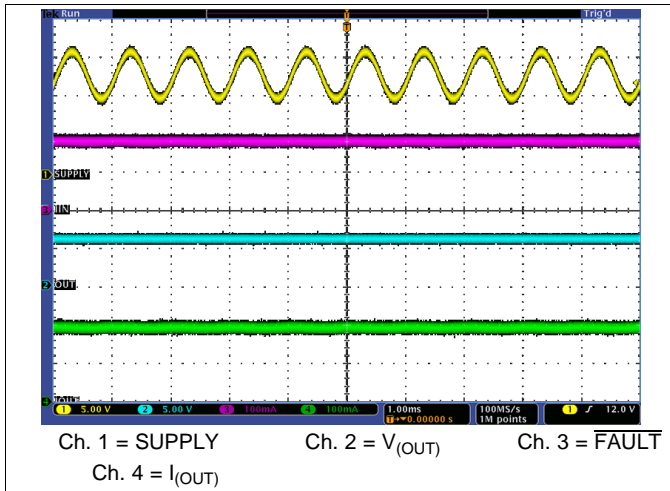


图 14. Superimposed Alternating Voltage, 1 kHz

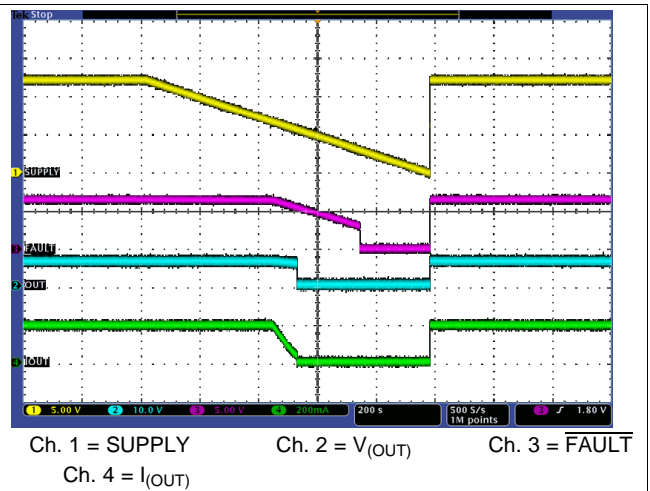


图 15. Slow Decrease, Quick Increase of Supply Voltage

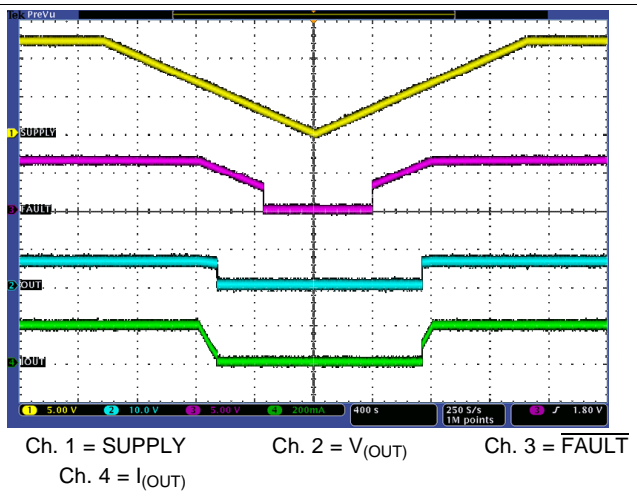


图 16. Slow Decrease and Slow Increase of Supply Voltage

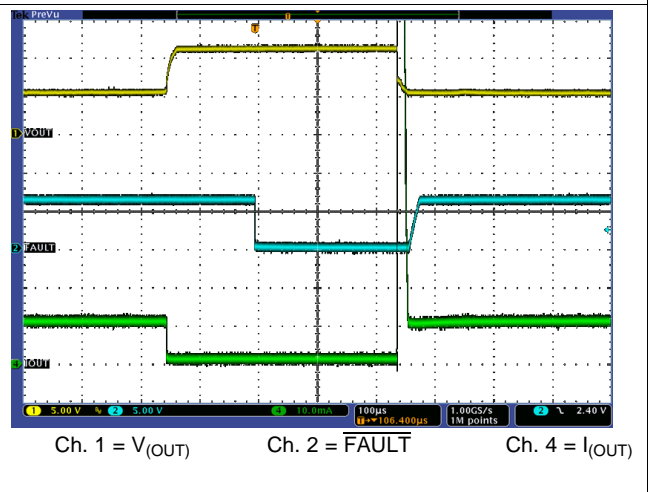


图 17. LED Open-Circuit Protection and Recovery

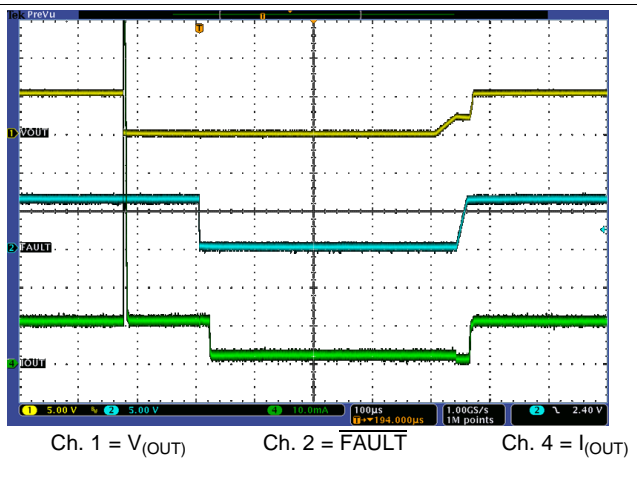


图 18. LED Short-Circuit Protection and Recovery

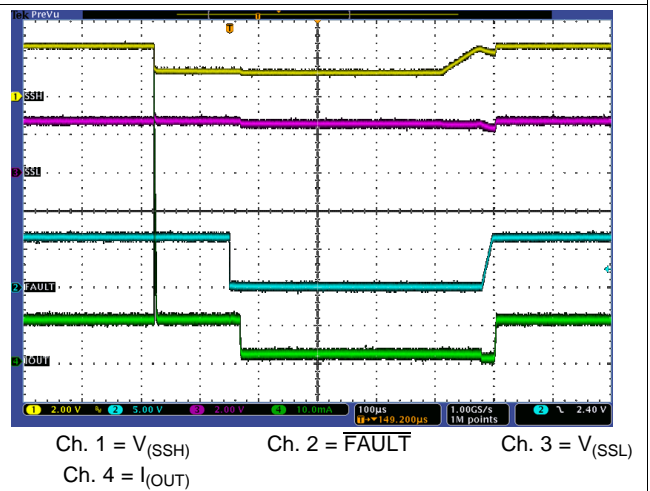


图 19. Single-LED-Short Protection and Recovery

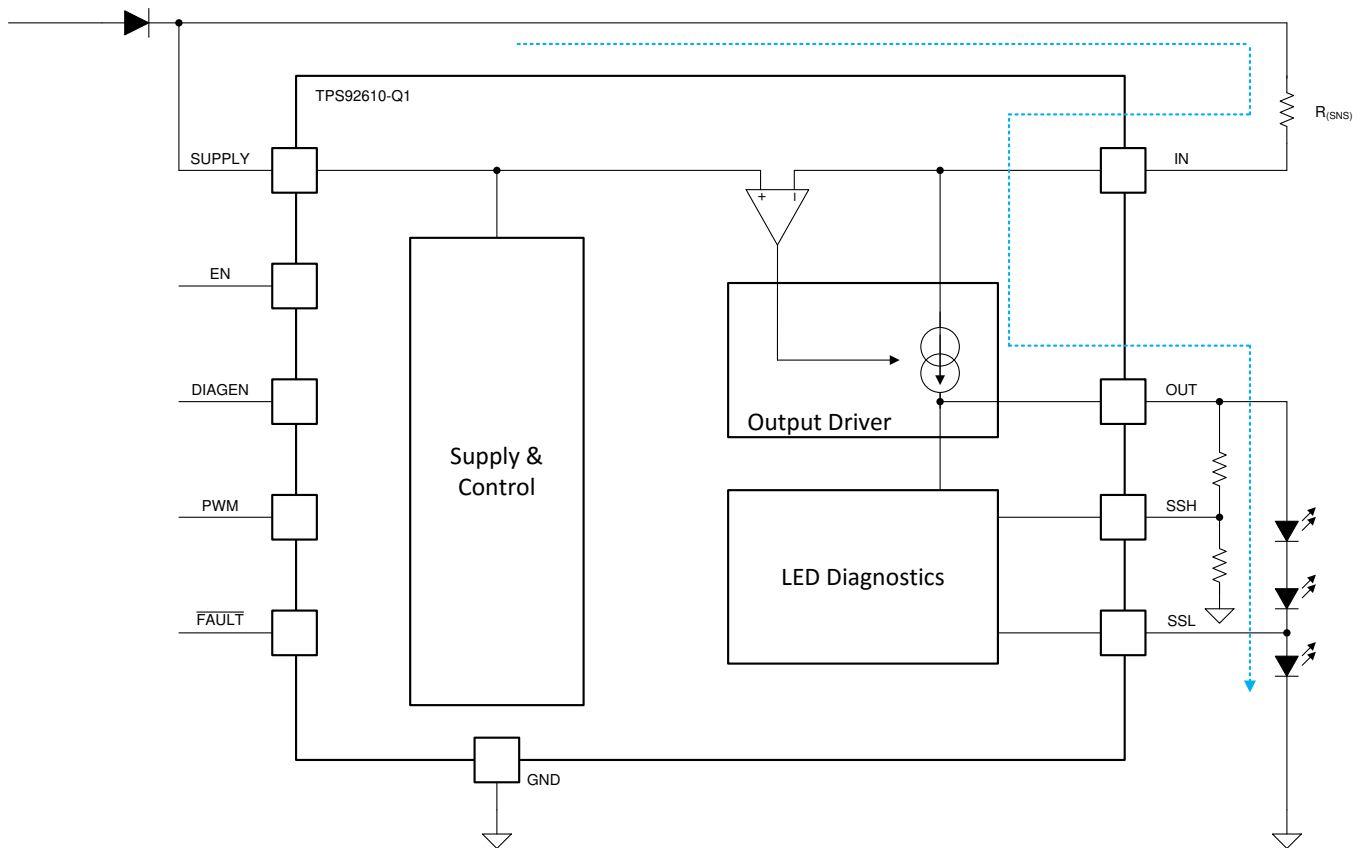
7 Detailed Description

7.1 Overview

The TPS92610-Q1 device is one of a family of single-channel linear LED drivers. The family provides a simple solution for automotive LED applications. Different package options in the family provide a variety of current ranges and diagnostic options. The TPS92610-Q1 device in an HTSSOP-14 package supports LED open-circuit detection and short-to-ground detection. Unique single-LED-short detection in the TPS92610-Q1 device can help diagnose if one LED within a string is shorted. A one-fails–all-fail fault bus allows the TPS92610-Q1 device to be used together with the TPS9261x-Q1, TPS9263x-Q1, and TPS9283x-Q1 families.

The output current can be set by an external $R_{(SNS)}$ resistor. Current flows from the supply through the $R_{(SNS)}$ resistor into the internal current source and to the LEDs.

7.2 Functional Block Diagram



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7.3 Feature Description

7.3.1 Device Bias

7.3.1.1 Power-On Reset (POR)

The TPS92610-Q1 device has an internal power-on-reset (POR) function. When power is applied to SUPPLY, the internal POR holds the device in the reset state until $V_{(SUPPLY)}$ is above $V_{(POR_rising)}$.

7.3.1.2 Low-Quiescent-Current Fault Mode

The TPS92610-Q1 device consumes minimal quiescent current when its \overline{FAULT} pin is externally pulled LOW. At the same time, the device shuts down the output driver.

Feature Description (接下页)

If device detects an internal fault, it pulls the $\overline{\text{FAULT}}$ output LOW with constant current to signal a fault alarm on the one-fails–all-fail fault bus.

7.3.2 Constant-Current Driver

The TPS92610-Q1 device has a high-side constant-current integrated driver. The device senses channel current with an external high-side current-sense resistor, $R_{(\text{SNS})}$. A current regulation loop drives an internal transistor and regulates the current-sense voltage at the current-sense resistor to $V_{(\text{CS_REG})}$. When the output driver is in regulation, the output current can be set using the following equation.

$$I_{(\text{OUT})} = \frac{V_{(\text{CS_REG})}}{R_{(\text{SNS})}} \quad (1)$$

7.3.3 Device Enable

The TPS92610-Q1 device has an enable input, EN. When EN is low, the device is in sleep mode with ultralow quiescent current $I_{(\text{Shutdown})}$. This low current helps to save system-level current consumption in applications where battery voltage directly connects to the device without high-side switches.

7.3.4 PWM Dimming

The TPS92610-Q1 device supports PWM output dimming via PWM input dimming and supply dimming.

The PWM input functions as an enable for the output current. When the PWM input is low, the device also disables the diagnostic features.

Supply dimming applies PWM dimming on the power input. For an accurate PWM threshold, TI recommends using a resistor divider on the PWM input to set the PWM threshold higher than $V_{(\text{POR_rising})}$.

7.3.5 Diagnostics

The TPS92610-Q1 device provides advanced diagnostics and fault protection features for automotive exterior lighting systems. The device is able to detect and protect from LED string short-to-GND, LED string open-circuit, and single-LED-short scenarios. It also supports a one-fails–all-fail fault bus that could flexibly fit different legislative requirements.

7.3.5.1 DIAGEN

The TPS92610-Q1 device supports the DIAGEN pin with an accurate threshold to disable the open-circuit and single-LED-short diagnostic functions. With a resistor divider, the DIAGEN pin can be used to sense SUPPLY voltage with a resistor-programmable threshold. With the DIAGEN feature, the device is able to avoid false error reports due to low-dropout voltage and to drive maximum current in low-dropout mode when the input voltage is not high enough for current regulation.

When $V_{(\text{DIAGEN})}$ is higher than the threshold $V_{\text{IH}(\text{DIAGEN})}$, the device enables LED open-circuit and single-LED-short diagnostics. When $V_{(\text{DIAGEN})}$ is lower than the threshold $V_{\text{IL}(\text{DIAGEN})}$, the device disables LED-open-circuit and single-LED-short diagnostics.

7.3.5.2 Low-Dropout Mode

When the supply voltage drops, the TPS92610-Q1 device tries to regulate current by driving internal transistors in the linear region, also known as low-dropout mode, because the voltage across the sense resistor fails to reach the regulation target.

In low-dropout mode, the open-circuit diagnostic must be disabled. Otherwise, the device treats the low-dropout mode as an open-circuit fault. The DIAGEN pin is used to avoid false diagnostics on the output channel due to low supply voltage.

When the DIAGEN voltage is low, single-LED short- and open-circuit detection is ignored. When the DIAGEN voltage is high, single-LED short- and open-circuit detection return to normal operation.

In dropout mode, a diode in parallel with the sense resistor is recommended to clamp the voltage between SUPPLY and IN (across the sense resistor) in case of a large current pulse during recovery.

Feature Description (接下页)

7.3.5.3 Open-Circuit Detection

The TPS92610-Q1 device has LED open-circuit detection. Open-circuit detection monitors the output voltage when the channel is in the ON state. Open-circuit detection is only enabled when DIAGEN is HIGH. A short-to-battery fault is also detected as an LED open-circuit fault.

The device monitors dropout-voltage differences between the IN and OUT pins when PWM is HIGH. The voltage difference $V_{(IN)} - V_{(OUT)}$ is compared with the internal reference voltage $V_{(OPEN_th_rising)}$ to detect an LED open-circuit failure. If $V_{(IN)} - V_{(OUT)}$ falls below the $V_{(OPEN_th_rising)}$ voltage longer than the deglitch time of $t_{(OPEN_deg)}$, the device asserts an open-circuit fault. Once an LED open-circuit failure is detected, the constant-current source pulls the fault bus down. During the deglitch time period, if $V_{(IN)} - V_{(OUT)}$ rises above $V_{(OPEN_th_falling)}$, the deglitch timer is reset.

When the device is in auto-retry, the device keeps the output ON to retry if the PWM input is HIGH; the device sources a small current $I_{(retry)}$ from IN to OUT when PWM input is LOW. In either scenario, once a faulty channel recovers, the device resumes normal operation and releases the \overline{FAULT} pulldown.

7.3.5.4 Short-to-GND Detection

The TPS92610-Q1 device has LED short-to-GND detection. Short-to-GND detection monitors the output voltage when the channel is in the ON state. Once a short-to-GND LED failure is detected, the device turns off the output channel and retries automatically, ignoring the PWM input. If the retry mechanism detects removal of the LED short-to-GND fault, the device resumes normal operation.

The device monitors the $V_{(OUT)}$ voltage and compares it with the internal reference voltage to detect a short-to-GND failure. If $V_{(OUT)}$ falls below $V_{(SG_th_rising)}$ longer than the deglitch time of $t_{(SG_deg)}$, the device asserts the short-to-GND fault and pulls \overline{FAULT} low. During the deglitching time period, if $V_{(OUT)}$ rises above $V_{(SG_th_falling)}$, the timer is reset.

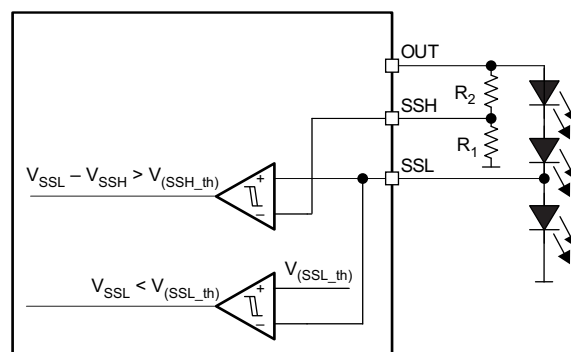
Once the device has asserted a short-to-GND fault, the device turns OFF the output channel and retries automatically with a small current. When retrying, the device sources a small current $I_{(retry)}$ from IN to OUT to pull up the LED loads continuously. Once auto-retry detects output voltage rising above $V_{(SG_th_falling)}$, it clears the short-to-GND fault and resumes normal operation.

7.3.5.5 Single-LED-Short Detection

The TPS92610-Q1 device supports single-LED-short detection by using the SSH and SSL pins. In case there is no need of this feature, SSH and SSL must be tied together to a resistor divider to avoid false alarms as shown in [图 21](#).

The TPS92610-Q1 device has integrated a precision comparator to monitor a single-LED-short failure. The comparator uses the bottom LED forward voltage $V_{(SSL)}$ as a reference and monitors the string voltage $V_{(OUT)}$ with resistor divider R_1 and R_2 at $V_{(SSH)}$.

If a single-LED short is detected, the device turns off the output channel and retries with a small current $I_{(RETRY)}$. Once the fault is removed, the device automatically resumes normal operation.



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图 20. Single-LED Short Detection

Feature Description (接下页)

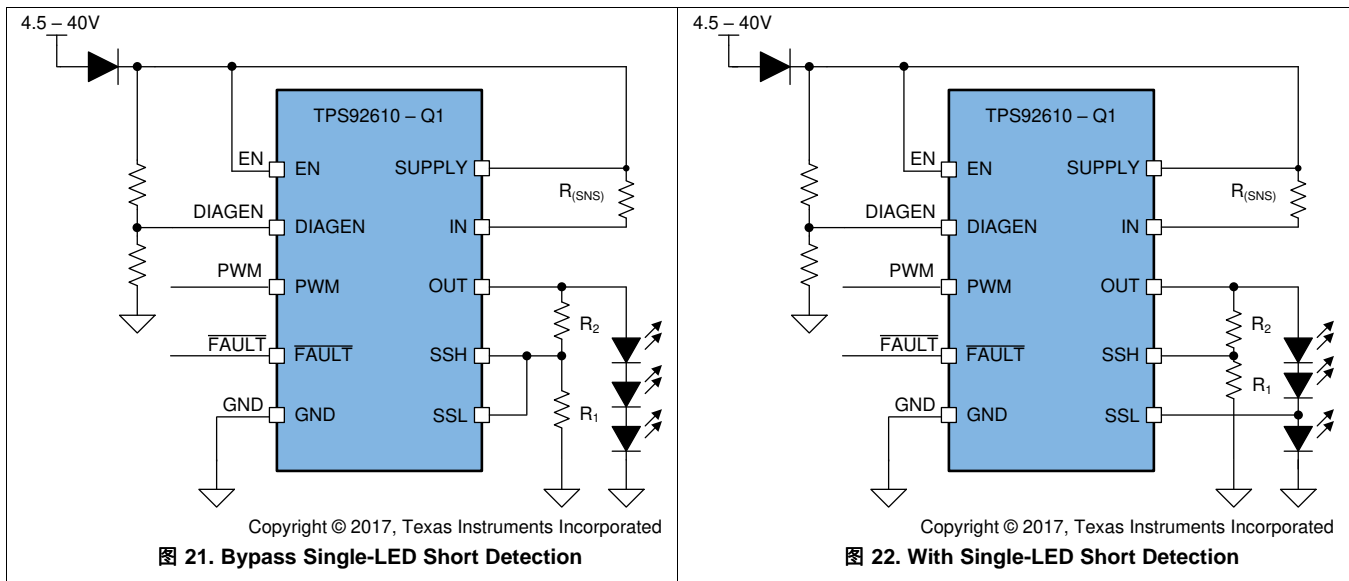
Use the following equation to calculate the ratio of R1 and R2.

$$R_2 = (\text{No. of LEDs} - 1) \times R_1 \quad (2)$$

By using the resistor divider with values calculated in 公式 2, the voltages of SSH and SSL are then equal to the forward voltage of a single LED. With built-in comparators, the device can report failure if any of the LEDs is shorted within this string.

An internal resistor string on SSL and resistors R₁ and R₂ draw current from OUT. TI recommends total resistance of R₁ and R₂ greater than 100-kΩ, so the current has negligible effect on LED luminance.

Even within the same batch of LEDs, the LED forward voltage may vary from one to another. Taking account of forward voltage differences is necessary to avoid any false faults.



7.3.5.6 Overtemperature Protection

The TPS92610-Q1 device monitors device junction temperature. When the junction temperature reaches thermal shutdown threshold $T_{(TSD)}$, the output shuts down. Once the junction temperature falls below $T_{(TSD)} - T_{(TSD_HYS)}$, the device resumes normal operation. During overtemperature protection, the FAULT bus is pulled low.

7.3.6 FAULT Bus Output With One-Fails-All-Fail

The TPS92610-Q1 device has a FAULT bus for diagnostics output. In normal operation, $\overline{\text{FAULT}}$ is weakly pulled up by an internal pullup current source $I_{(\text{FAULT_pullup})}$ higher than $V_{\text{OH}(\text{FAULT})}$. If any fault scenario occurs, the FAULT bus is strongly pulled low by the internal pulldown current source $I_{(\text{FAULT_pulldown})}$. Once $V_{(\text{FAULT})}$ falls below $V_{\text{IL}(\text{FAULT})}$, all outputs shut down for protection. The faulty channel keeps retrying until the fault scenario is removed.

If $\overline{\text{FAULT}}$ is externally pulled up with a current larger than $I_{(\text{FAULT_pulldown})}$, the one-fails-all-fail function is disabled and only the faulty channel is turned off.

The FAULT bus is able to support up to 15 pieces of TPS9261x-Q1, TPS9263x-Q1, or TPS9283x-Q1 devices.

Feature Description (接下页)

表 1. Fault Table With DIAGEN = HIGH

FAULT BUS STATUS	FAULT TYPE	DETECTION MECHANISM	CHANNEL STATE	DEGLITCH TIME	FAULT BUS	FAULT HANDLING ROUTINE	FAULT RECOVERY
FAULT floating or externally pulled up	Open-circuit or short-to-supply	$V_{(IN)} - V_{(OUT)} < V_{(OPEN_th_rising)}$	On	$t_{(OPEN_deg)}$	Constant-current pulldown	Device works normally with FAULT pin pulled low. Device sources $I_{(retry)}$ current when PWM is LOW. Device keeps output normal when PWM is HIGH.	Auto recover
	Short-to-ground	$V_{(OUT)} < V_{(SG_th_rising)}$	On	$t_{(SG_deg)}$	Constant-current pulldown	Device turns output off and retries with constant current $I_{(retry)}$, ignoring the PWM input.	Auto recover
	Single-LED short	$V_{(SSL)} - V_{(SSH)} > V_{(SS_th)}$ or $V_{(SSL)} < V_{(SSL_th)}$	On	$t_{(SS_deg)}$	Constant-current pulldown	Device turns output off and retry with constant current $I_{(retry)}$, ignoring the PWM input.	Auto recover
	Overtemperature	$T_J > T_{(TSD)}$	On or off	$t_{(TSD_deg)}$	Constant-current pulldown	Devices turns output off.	Auto recover
Externally pulled low	Device turns output off						

表 2. Fault Table With DIAGEN = LOW

FAULT BUS STATUS	FAULT TYPE	DETECTION MECHANISM	CHANNEL STATE	DEGLITCH TIME	FAULT BUS	FAULT HANDLING ROUTINE	FAULT RECOVERY
FAULT floating or externally pulled up	Open-circuit or short-to-supply	Ignored					
	Short-to-ground	$V_{OUT} < V_{(SG_th_rising)}$	On	$t_{(SG_deg)}$	Constant-current pulldown	Device turns output off and retries with constant current $I_{(retry)}$, ignoring the PWM input.	Auto recover
	Single-LED short	Ignored					
	Overtemperature	$T_J > T_{(TSD)}$	On or off	$t_{(TSD_deg)}$	Constant-current pulldown	Device turns output off.	Auto recover
Externally pulled low	Device turns output off						

7.4 Device Functional Modes

7.4.1 Undervoltage Lockout, $V_{(SUPPLY)} < V_{(POR_rising)}$

When the device is in undervoltage lockout mode, the TPS92610-Q1 device disables all functions until the supply rises above the UVLO-rising threshold.

Device Functional Modes (接下页)

7.4.2 Normal Operation $V_{(\text{SUPPLY})} \geq 4.5 \text{ V}$

The device drives an LED string in normal operation. With enough voltage drop across SUPPLY and OUT, the device is able to drive the output in constant-current mode.

7.4.3 Low-Voltage Dropout

When the device drives an LED string in low-dropout mode, if the voltage drop is less than open-circuit detection threshold, the device may report a false open-circuit fault. Set the DIAGEN threshold higher than LED string voltage to avoid a false open-circuit detection.

7.4.4 Fault Mode

When the device detects an open circuit or a shorted LED, the device tries to pull down the $\overline{\text{FAULT}}$ pin with a constant current. If the FAULT bus is pulled down, the device switches to fault mode and consumes a fault current of $I_{(\text{FAULT})}$.

8 Application and Implementation

注

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

In automotive lighting applications, thermal performance and LED diagnostics are always design challenges for linear LED drivers.

The TPS92610-Q1 device is capable of detecting LED open-circuit, LED short-circuit and single-LED short failures. To increase current-driving capability, the TPS92610-Q1 device supports using an external a parallel resistor to help dissipate heat as shown in the following application, 图 25. This technique provides the low-cost solution of using external resistors to dissipate heat due to high input voltage, and still keeps high accuracy of the total current output. Note that the one-fails-all-fail feature is not supported by this topology.

8.2 Typical Application

8.2.1 Single-Channel LED Driver With Full Diagnostics

The TPS92610-Q1 device is a potential choice for LED driver for applications with diagnostics requirements. In many cases, single-LED short diagnostics are mandatory for applications such as sequential turn indicators.

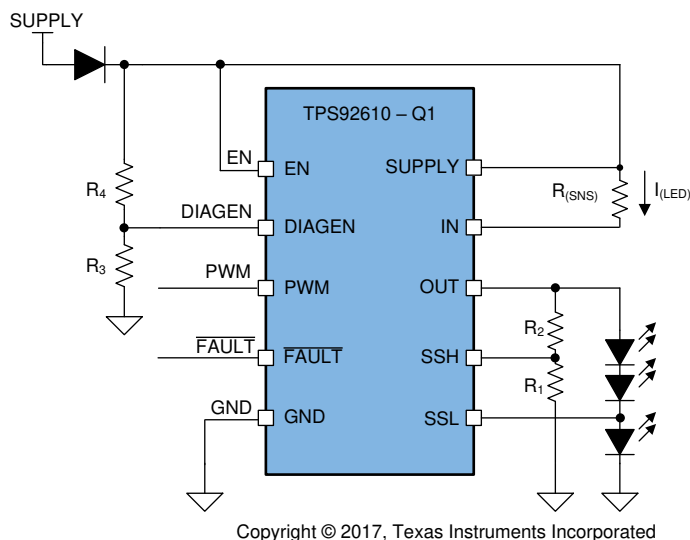


图 23. Typical Application Diagram

8.2.1.1 Design Requirements

Input voltage ranges from 9 V to 16 V, LED maximum forward voltage $V_{fmax} = 2.5$ V, minimum forward voltage $V_{fmin} = 1.9$ V, current $I_{(LED)} = 50$ mA.

8.2.1.2 Detailed Design Procedure

Current setting by sense resistor is as described in 公式 1.

$$R_{(SNS)} = \frac{R_{(CS_REG)}}{I_{(LED)}} = 1.96 \Omega$$

(3)

LED-string maximum forward voltage = 3×2.5 V = 7.5 V.

Typical Application (接下页)

With 400-mV headroom reserved for the TPS92610-Q1 device between SUPPLY and OUT, the TPS92610-Q1 device must disable open-circuit detection when the supply voltage is below 7.9 V by using the DIAGEN feature.

$$V_{IL(DIAG,min)} = \frac{7.9 \times R_3}{R_3 + R_4} \quad (4)$$

Set $R_4 = 10 \text{ k}\Omega$, $R_3 = 65.6 \text{ k}\Omega$.

The single-LED short-detection resistor ratio can be calculated as follows.

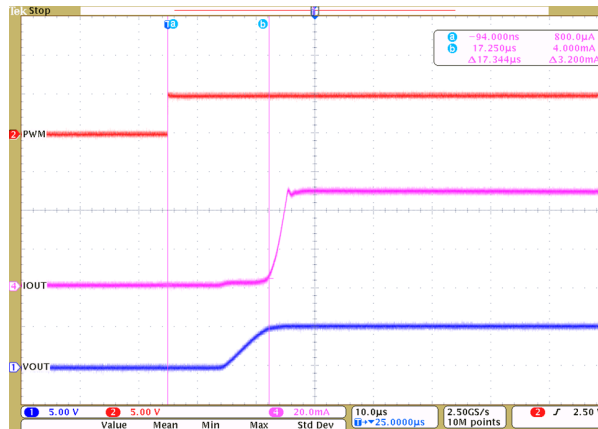
$$\frac{R_2}{R_1} = 2 \quad (5)$$

If $R_1 = 50 \text{ k}\Omega$, $R_2 = 100 \text{ k}\Omega$

Total device power consumption at worst case is with 16-V input and LEDs at minimal forward voltage.

$$\begin{aligned} P_{(Max)} &= (V_{(SUPPLY)} - V_{(CS_REG)} - V_{(OUT)}) \times I_{(LED)} + V_{(SUPPLY)} \times I_{(Quiescent)} \\ &= (16 - 3 \times 1.9 - 0.098) \times 0.05 + 16 \times 0.00025 = 0.5141 \text{ W} \end{aligned} \quad (6)$$

8.2.1.3 Application Curve



Ch. 1 = $V_{(OUT)}$ Ch. 2 = $V_{(PWM)}$ Ch. 4 = $I_{(OUT)}$

图 24. Output Current With PWM Input

Typical Application (接下页)

8.2.2 Single-Channel LED Driver With Heat Sharing

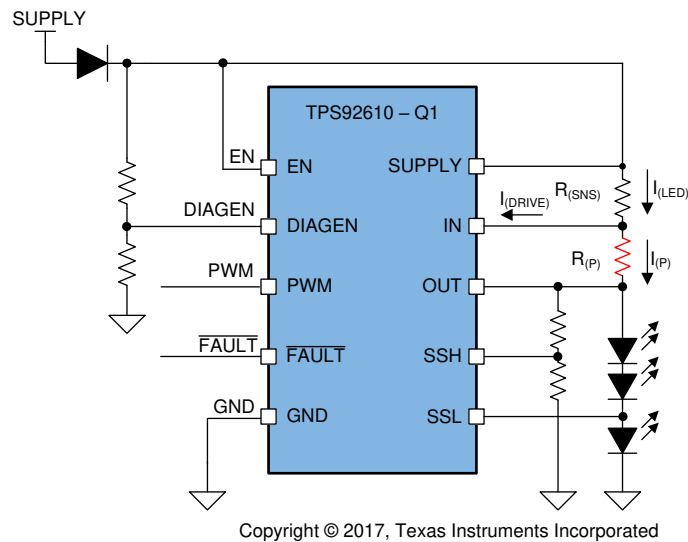


图 25. Heat Sharing With a Parallel Resistor

8.2.2.1 Design Requirements

Input voltage range is 9 V to 16 V, LED maximum forward voltage $V_{fmax} = 2.5$ V, minimum forward voltage $V_{fmin} = 1.9$ V, current $I_{(LED)} = 200$ mA.

8.2.2.2 Detailed Design Procedure

Using parallel resistors, thermal performance can be improved by balancing current between the TPS92610-Q1 device and the external resistors as follows. As the current-sense resistor controls the total LED string current, the LED string current $I_{(LED)}$ is set by $V_{(CS_REG)} / R_{(SNS)}$, while the TPS92610-Q1 current $I_{(DRIVE)}$ and parallel resistor current $I_{(P)}$ combine to the total current.

Note that the parallel resistor path cannot be shut down by PWM or fault protection. If PWM or one-fails-all-fail feature is required, TI recommends an application circuit as described in [Single-Channel LED Driver With Full Diagnostics](#).

In linear LED driver applications, the input voltage variation contributes to most of the thermal concerns. The resistor current, as indicated by Ohm's law, depends on the voltage across the external resistors. The TPS92610-Q1 controls the driver current $I_{(DRIVE)}$ to attain the desired total current. If $I_{(P)}$ increases, the TPS92610-Q1 device decreases $I_{(DRIVE)}$ to compensate, and vice versa.

While in low-dropout mode, the voltage across the $R_{(P)}$ resistor may be close to zero, so that almost no current can flow through the external resistor $R_{(P)}$.

When the input voltage is high, the parallel-resistor current $I_{(P)}$ is proportional to the voltage across the parallel resistor $R_{(P)}$. The parallel resistor $R_{(P)}$ takes the majority of the total string current, generating maximum heat. The device must prevent current from draining out to ensure current regulation capability.

In this case, the parallel resistor value must be carefully calculated to ensure that 1) enough output current is achieved in low-dropout mode, 2) thermal dissipation for both the TPS92610-Q1 device and the resistor is within their thermal dissipation limits, and 3) device current in the high-voltage mode is above the minimal output-current requirement.

Current setting by sense resistor is as described in [公式 7](#).

Typical Application (接下页)

$$R_{(SNS)} = \frac{V_{(CS_REG)}}{I_{(LED)}} = 0.49\Omega$$

(7)

LED-string maximum forward voltage = $3 \times 2.5 \text{ V} = 7.5 \text{ V}$.

Parallel resistor $R_{(P)}$ is recommended to consume 50% of the total current at maximum supply voltage.

$$R_{(P)} = \frac{V_{(SUPPLY)} - V_{(CS_REG)} - V_{(OUT)}}{0.5 \times I_{(LED)}} = \frac{16 - 3 \times 1.9 - 0.098}{0.5 \times 0.2} \approx 100\Omega$$

(8)

Total device power consumption is maximum at 16 V input and LED minimal forward voltage.

$$\begin{aligned} P_{(DEV_MAX)} &= (V_{(SUPPLY)} - V_{(CS_REG)} - V_{(OUT)}) \times \left(I_{(LED)} - \frac{V_{(SUPPLY)} - V_{(CS_REG)} - V_{(OUT)}}{R_{(P)}} \right) + V_{(SUPPLY)} \times I_{(Quiescent)} \\ &= (16 - 3 \times 1.9 - 0.098) \times 0.1 + 16 \times 0.00025 = 1.0242 \text{ W} \end{aligned}$$

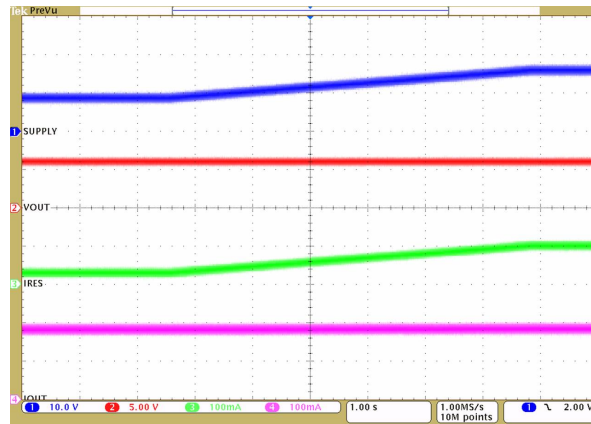
(9)

Resistor $R_{(P)}$ maximum power consumption is at 16-V input.

$$P_{(RP_MAX)} = \frac{(V_{(SUPPLY)} - V_{(CS_REG)} - V_{(OUT)})^2}{R_{(P)}} = \frac{(16 - 3 \times 1.9 - 0.098)^2}{100} = 1.04 \text{ W}$$

(10)

Users must consider the maximum power of both of the device and the parallel resistor.

Typical Application (接下页)
8.2.2.3 Application Curve


Ch. 1 = $V_{(SUPPLY)}$ Ch. 2 = $V_{(OUT)}$ Ch. 3 = $I_{(P)}$

Ch. 4 = $I_{(LED)}$ $V_{(SUPPLY)}$ increases from 9 V to 16 V

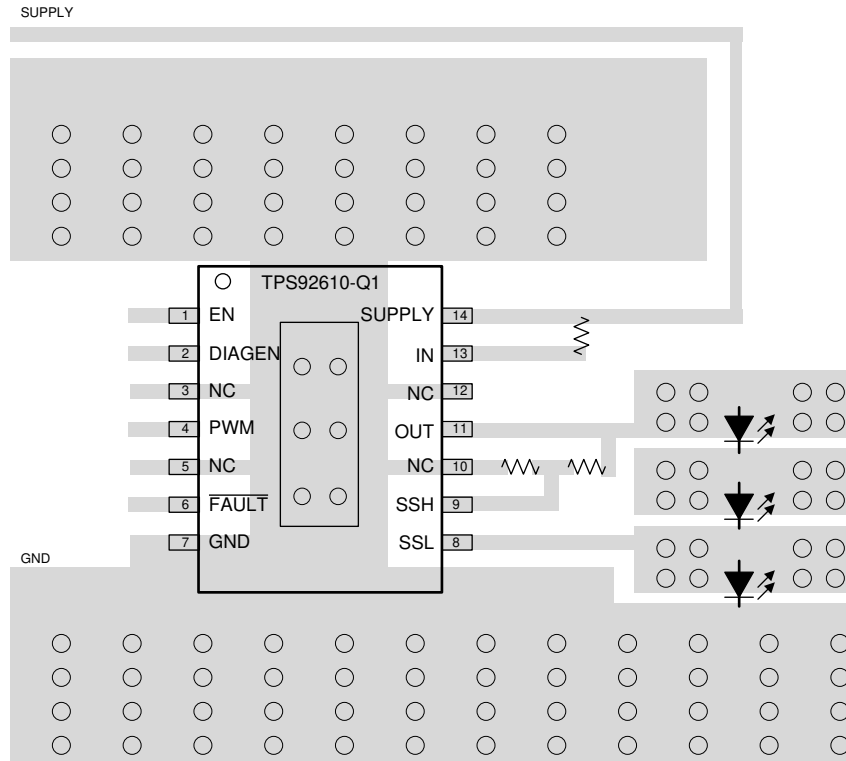
图 26. Constant Output Current With Increasing Supply Voltage

9 Layout

9.1 Layout Guidelines

Thermal dissipation is the primary consideration for TPS92610-Q1 layout. TI recommends good thermal dissipation area connected to thermal pads with thermal vias.

9.2 Layout Example



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图 27. TPS92610-Q1 Example Layout Diagram

10 器件和文档支持

10.1 文档支持

10.1.1 相关文档

请参阅如下相关文档：

- [《TPS92610-Q1 EVM 用户指南》](#)
- [《如何在汽车外部照明应用中计算 TPS92630-Q1 最大输出 电流》](#)
- [《适用于中央高位刹车灯 \(CHMSL\) 的汽车线性 LED 驱动器参考设计》](#)
- [《用户指南：适用于中央高位刹车灯 \(CHMSL\) 的汽车线性 LED 驱动器参考设计》](#)

10.2 接收文档更新通知

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10.3 社区资源

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

10.6 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

11 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是适用于指定器件的最新数据。数据如有变更，恕不另行通知，且不会对此文档进行修订。如需获取此数据表的浏览器版本，请查看左侧的导航面板。

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS92610QPWRQ1	ACTIVE	HTSSOP	PWP	14	2000	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 125	TP92610	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS92610QPWPRQ1	HTSSOP	PWP	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS

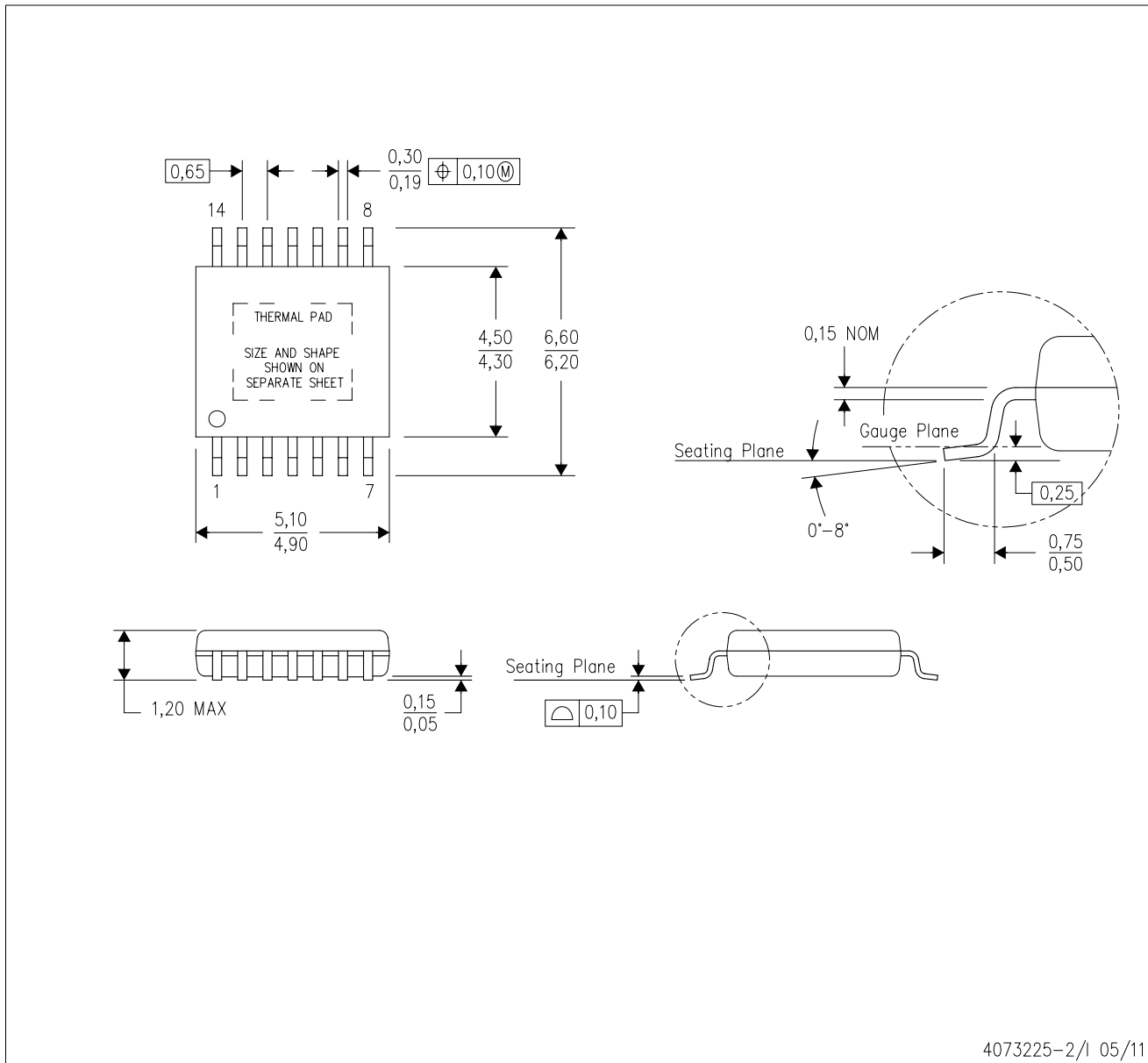


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS92610QPWPRQ1	HTSSOP	PWP	14	2000	350.0	350.0	43.0

PWP (R-PDSO-G14)

PowerPAD™ PLASTIC SMALL OUTLINE



4073225-2/1 05/11

- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Body dimensions do not include mold flash or protrusions. Mold flash and protrusion shall not exceed 0.15 per side.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <<http://www.ti.com>>.
 - See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
 - Falls within JEDEC MO-153

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THERMAL PAD MECHANICAL DATA

PWP (R-PDSO-G14)

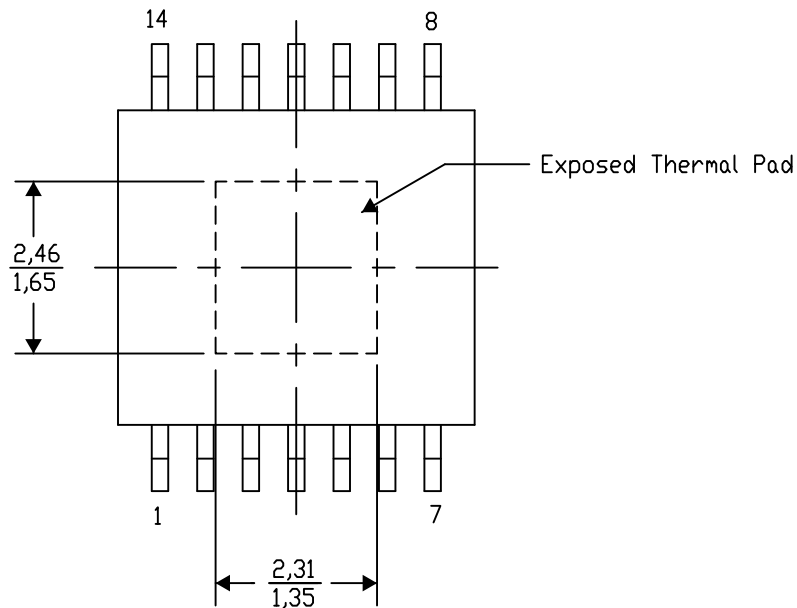
PowerPAD™ SMALL PLASTIC OUTLINE

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Top View

Exposed Thermal Pad Dimensions

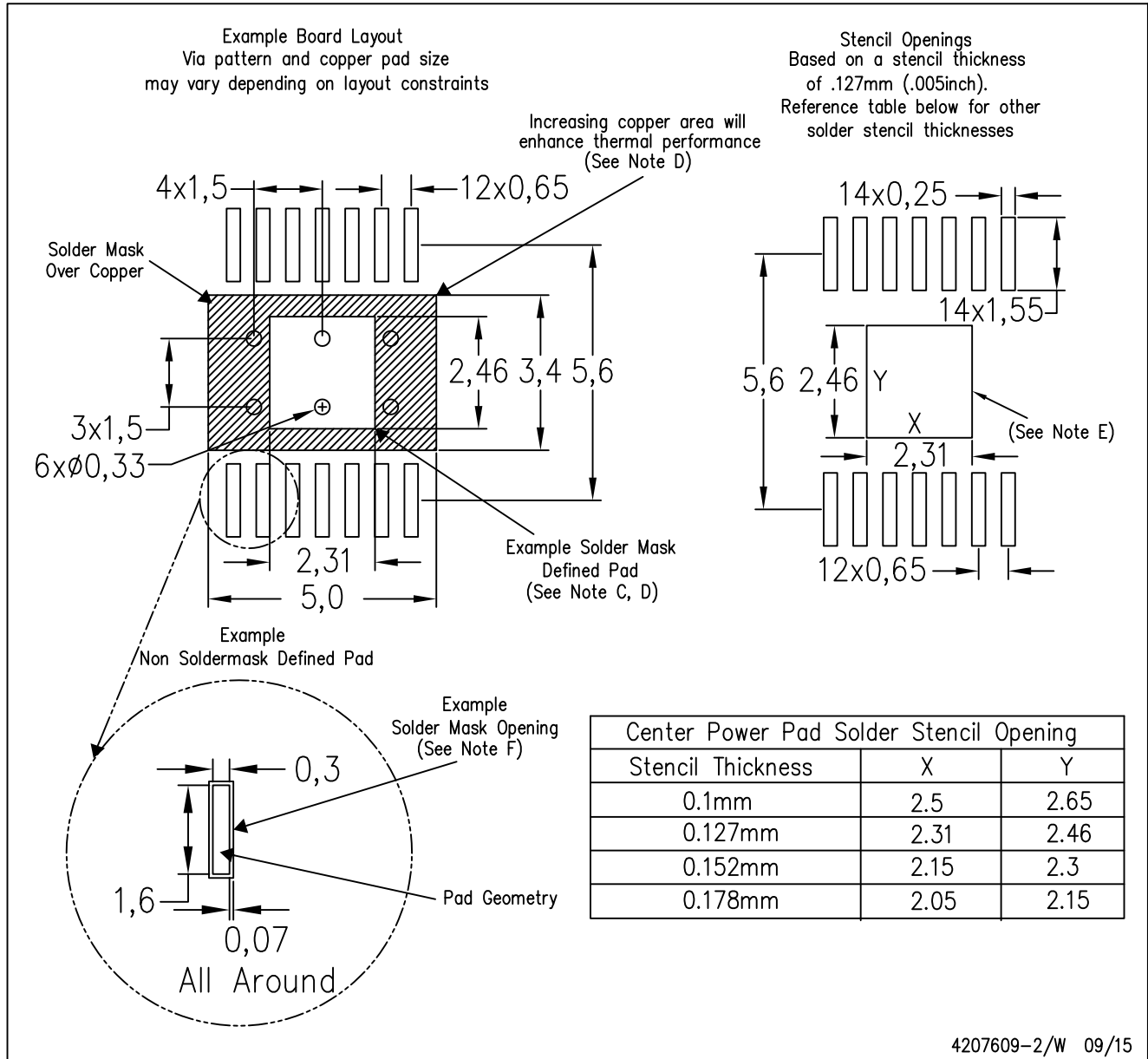
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NOTE: A. All linear dimensions are in millimeters

PowerPAD is a trademark of Texas Instruments

PWP (R-PDSO-G14)

PowerPAD™ PLASTIC SMALL OUTLINE



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- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>. Publication IPC-7351 is recommended for alternate designs.
 - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
 - F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

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